**Electronics Tutorial**

**Introduction**
This document is not an in-depth presentation of the subject of electronics. Instead, it is intended to give you sufficient (empirical) knowledge of the subject to be able to understand, design and build simple circuits such as the control circuits used with the ‘Free Energy’ devices described in the later parts of this eBook.

**Disclaimer**
This material is provided for information purposes only. Should you decide to attempt construction of some device based on information presented here and injure yourself or any other person, I am not liable in any way. To clarify this; should you construct something in a heavy box and drop it on your toe, I am not liable for any injury you may sustain (you should learn to be more careful). If you attempt to construct some electronic circuit and burn yourself with the soldering iron, I am not liable. Also, I strongly recommend that unless you are expert in electronics, you do not construct any device using, or producing more than 30 Volts - high voltage circuits are extremely dangerous and should be avoided until you gain experience or can obtain the help and supervision of a person experienced in constructing high voltage circuits.

**Voltage.**
Voltage is the key to understanding electronics. Without voltage, nothing happens in electronics. What is it? Nobody knows. We know how to generate it. We know what it does. We know how to measure it, but nobody knows what it actually is.

It is also called “Electro Motive Force” or “EMF” which is no help whatsoever in knowing what it is. That, is roughly equivalent to saying “the thing that pushes is the thing that pushes” - very true but absolutely no help whatsoever. OK, having admitted that we really don't know what it is, we can start to say the things we do know about it:

A new battery has a voltage between its terminals. This voltage is said to cause a current to flow through any complete electrical circuit placed across it. The current flowing through the circuit can cause various things to happen such as creating light, creating sound, creating heat, creating magnetism, creating movement, creating sparks, etc., etc.

By using the current caused by a voltage, a device called a ‘Volmeter’ can indicate how big the voltage is. The bigger the voltage, the bigger the current and the bigger the display on the voltmeter. The voltmeter can have a numerical display where you read the voltage directly from the display, or it can be an ‘analogue’ voltmeter where the voltage is shown by the position of a needle on a scale. The size of the voltage is stated in ‘Volts’ which is a unit of measurement named after the man Volta who introduced voltage to the world (it was always there, we just did not know about it).

Voltages add up if they are connected the same way round, i.e. with the + terminals all facing the same way:

The physical size of the battery usually determines the length of time it can supply any given current - the bigger the battery, the longer it can provide any given current. A battery is constructed from a number of ‘cells’. The number of cells in the battery controls the voltage of the battery. For example, an ‘AA’ size battery (what used to be called a ‘penlight’ battery) has a single ‘cell’ and so produces 1.5 Volts when new. The very much larger and heavier ‘D’ battery also has just one cell and so it also produces 1.5 Volts when new. The difference (apart from the higher cost of the ‘D’ cell) is that the larger cell can provide a much higher current if both batteries are discharged over the same period of time.
There are several different types of battery construction. A rechargeable NiCad battery has a single cell but its construction method means that it produces about 1.35 Volts when fully charged. In passing, NiCad batteries have a ‘memory’ characteristic which means that if they are recharged before they are fully discharged, then the next time they are discharged they run out of power at the voltage level it had when the last charging was started. Consequently, it is a good idea to fully discharge a NiCad battery before charging it again.

Car and motorcycle batteries are described as Lead/Acid batteries. This type of construction is not very convenient being large, heavy and potentially corrosive. The big advantages are the ability to provide very high currents and giving 2.0 Volts per cell. These batteries are normally produced as 6 Volt or 12 Volt units. The Amp-Hours for lead/acid car batteries is usually quoted for a 20 hour discharge period, so a fully charged, new, 20 AHR battery can provide 1 Amp for 20 hours of continuous use. That battery loaded to give 5 Amps, will not provide that current for 4 hours but might only last 2 hours, or perhaps a little better. The manufacturer’s literature should give an indication of the performance, but if it is important, run your own test to see how the battery actually works in practice.

“Mains units” are known in the electronics world as “Power Supply Units” or “PSUs” for short. These convert the mains voltage (220 Volts in UK, 110 Volts in USA) to some convenient low voltage; 12 Volts, 9 Volts, 6 Volts, or whatever is needed. A mains unit can provide several different voltages simultaneously.

Resistance.
Being familiar with Voltage and Resistance is the key to understanding electronic circuitry. Resistance is a measure of how difficult it is for current to flow through something. Some materials such as glass, ceramics, wood and most plastics do not easily carry a current and so are considered to be ‘insulators’. That is why you will see power lines hung from their pylons by a series of ceramic discs. Current flows easily through metals, especially along the surface of the metal, so cables are made from metal wires surrounded by a layer of plastic insulation. The higher grade cables have wire cores made up of many small-diameter strands as this increases the surface area of the metal for any given cross-sectional area of the metal core (it also makes the cable more flexible, and generally, more expensive).

There is a very important, third group of materials, silicon and germanium in particular, which fall between conductors and insulators. Not surprisingly, these are called ‘semi-conductors’ and the amount of current they can carry depends on the electrical conditions in which they are placed. Much, much more about this later on.

While a metal wire carries current very well, it is not perfect at the job and so has some ‘resistance’ to current flowing through it. The thicker the wire, the lower the resistance. The shorter the wire, the lower the resistance. The first researchers used this characteristic to control the way circuits operated. Sometimes, as higher resistances were needed, the researcher used to need long lengths of wire which would get tangled up. To control the wire, a board with nails along each side was used and the wire wound backwards and forwards across the board like this:

![Resistor symbol](image)

When drawing a circuit diagram, the researcher would sketch the wire on the board giving a zig-zag line which is still used today to represent a ‘resistor’ although different methods of construction are now used. An alternative symbol for a resistor is a plain rectangle as shown above.

If a resistor is connected across a battery, a circuit is formed and a current flows around the circuit. The current cannot be seen but that does not mean that it is not there. Current is measured in ‘Amps’ and the instrument used to display it is an ‘ammeter’. If we place an ammeter in the circuit, it will show the current flowing around the circuit. In passing, the ammeter itself, has a small resistance and so putting it in the circuit does reduce the current flow around the circuit very slightly. Also shown is a bulb. If the current flowing around the circuit is sufficiently high and the bulb chosen correctly, then the bulb will light up, showing that current is flowing, while the ammeter will indicate exactly how much current is flowing.
Shown on the right, is the way that this circuit would be shown by an electronics expert (the 'Resistor', 'Ammeter' and 'Lamp' labels would almost certainly not be shown). There are several different styles of drawing circuit diagrams, but they are the same in the basic essentials. One important common feature is that unless there is some very unusual and powerful reason not to do so, every standard style circuit diagram will have the positive voltage line horizontally at the top of the diagram and the negative as a horizontal line at the bottom. These are often referred to as the positive and negative ‘rails’. Where possible, the circuit is drawn so that its operation takes place from left to right, i.e. the first action taken by the circuit is on the left and the last action is placed on the right.

Resistors are manufactured in several sizes and varieties. They come in ‘fixed’ and ‘variable’ versions. The most commonly used are the ‘fixed’ carbon ‘E12’ range. This is a range of values which has 12 resistor values which repeat: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and then: 100, 120, 150, 180, 220, 270, 330, 390, 470, 560, 680, 820 and then: 1000, 1200, 1500, 1800, 2200, 2700, 3300, 3900, 4700, 5600, 6800, 8200, etc. etc. Nowadays, circuits often carry very little power and so the resistors can, and are, made in very small physical sizes. The higher the resistance value of a resistor, the less current will flow through it when a voltage is placed across it. As it can be difficult to see printing on small resistors clustered together on a circuit board and surrounded by other larger components, the resistor values are not written on the resistors, instead, the resistors are colour-coded. The unit of measurement for resistors is the ‘ohm’ which has a very small size. Most resistors which you encounter will be in the range 100 ohms to 1,000,000 ohms. The higher the resistance of any resistor, the smaller the current which will flow through it.

The colour code used on resistors is:

0 Black
1 Brown
2 Red
3 Orange
4 Yellow
5 Green
6 Blue
7 Purple (Violet if your colour vision is very good)
8 Grey
9 White

Each resistor has typically, three colour bands to indicate its value. The first two bands are the numbers and the third band is the number of noughts:

Green: 5  Blue: 6  Red: 2 noughts  Value: 5,600 ohms or 5.6K or 5K6

Yellow: 4  Purple: 7  Green: 5 noughts  Value: 4,700,000 ohms or 4.7M or 4M7

The colour bands are read from left to right and the first band is close to one end of the body of the resistor. There is often a fourth band which indicates the manufacturing tolerance: you can ignore that band.
Examples:

Red, Red, Red: 22 00 ohms or 2K2
Yellow, Purple, Orange: 47 000 ohms or 47K
Brown, Black, Brown: 100 ohms or 100R
Orange, Orange, Orange: 33 000 ohms or 33K
Brown, Green, Red: 15 00 ohms or 1K5
Brown, Green, Black: 15 no noughts, or 15 ohms
Blue, Grey, Orange: 68 000 ohms or 68K
Brown, Green, Green: 1,500,000 ohms or 1M5
Yellow, Purple, Brown: 47 0 ohms

As there are only 12 standard resistor values per decade, there are only 12 sets of the first two colour bands:

10: Brown/Black,
12: Brown/Red,
15: Brown/Green,
18: Brown/Grey
22: Red/Red,
27: Red/Purple
33: Orange/Orange,
39: Orange/White
47: Yellow/Purple
56: Green/Blue
68: Blue/Grey
82: Grey/Red

![Resistor Bands Diagram](image-url)
The details above give you all the basic information on resistor colour codes but there are a few additional refinements. There is an extra colour band further down the body of the resistor as shown here:

This extra band is used to indicate the manufacturing tolerance of the construction of the resistor. Resistor values are never exact and this rarely has any significant effect on their use in circuits. If some circuit needs very accurate resistor values in it, then buy several resistors of the same nominal value and use an ohm-meter to measure that actual value of each particular resistor and if none are perfect, then use two or more resistors to give the exact value wanted.

The tolerance band has the following codes:

Silver ± 10% (i.e. a 10K resistor of this type should be between 9K and 11K)
Gold ± 5% (i.e. a 10K resistor of this type should be between 9.5K and 10.5K)
Red ± 2% (i.e. a 10K resistor of this type should be between 9.8K and 10.2K)
Brown ± 1% (i.e. a 10K resistor of this type should be between 9.9K and 10.1K)
Green ± 0.5% (i.e. a 10K resistor of this type should be between 9.95K and 10.05K)
Blue ± 0.25% (i.e. a 10K resistor of this type should be between 9.975K and 10.025K)
Purple ± 0.1% (i.e. a 10K resistor of this type should be between 9.99K and 10.01K)

This type of resistor in the 10% and 5% ranges are the most common as they are the cheapest to buy and so tend to be the most popular. Recently, however, two additions to the coding have been introduced in order to allow for very high specification resistors which the average constructor may never come across. Each of these additions involves one additional colour band. The first additional colour band allows an extra digit in the resistor value, and looks like this:
As before, the colour coding is exactly the same, with the fourth colour band specifying the number of zeros after the digits indicated by the colour bands in front of it. So, in the example shown above, the first band being Red indicates a "2". The second colour band being Purple indicates a "7". The third colour band being Green indicates a "5" and the fourth colour band being Red indicates "2 zeros", so putting those together it produces the value of 27,500 ohms, which can also be written as 27.5 K or more briefly as 27K5.

Another example of this is:

The fourth colour band coding has also been extended to include two other colours:
Gold: meaning "no zeros and divided by 10" so if the band in the example above had been gold, then the value would be 56.4 ohms.
Silver: meaning "no zeros and divided by 100" and if the example band had been silver then the value would have been 5.64 ohms.

So, for example, if the resistor had a fourth colour band which was silver, then the value would be:

Finally, for very high-quality applications (typically military applications), there can be a sixth colour band positioned outside the tolerance band, and that final colour band states how much the resistance value can be expected to alter with changes in temperature. This is not something which is likely to be of any interest to you, but the codes for that final colour band are:

Brown: 0.01% of the resistor value for each degree Centigrade change in temperature.
Red: 0.005% of the resistor value for each degree Centigrade change in temperature.
Yellow: 0.0025% of the resistor value for each degree Centigrade change in temperature.
Orange: 0.0015% of the resistor value for each degree Centigrade change in temperature.

To put this in context, the worst of these represents a change of 1% in the resistor value when moving from the temperature of ice to the temperature of boiling water. Is this something which you really care about? I don't.

Leaving the details of identifying individual resistors, we now come to the interesting part: what happens when there are several resistors in a circuit. The important thing is to keep track of the voltages generated within the circuit. These define the currents flowing, the power used and the way in which the circuit will respond to external events. Take this circuit:
What is the voltage at point 'A'? If you feel like saying “Who cares?” then the answer is “you” if you want to understand how circuits work, because the voltage at point ‘A’ is vital. For the moment, ignore the effect of the voltmeter used to measure the voltage.

If R1 has the same resistance as R2, then the voltage at ‘A’ is half the battery voltage, i.e. 4.5 Volts. Half the battery voltage is dropped across R1 and half across R2. It does not matter what the actual resistance of R1 or R2 is, as long as they have exactly the same resistance. The higher the resistance, the less current flows, the longer the battery lasts and the more difficult it is to measure the voltage accurately.

There is no need to do any calculations to determine the voltage at point “A” as it is the ratio of the resistor values which determines the voltage. If you really want to, you can calculate the voltage although it is not necessary. The method for doing this will be shown you shortly. For example, if R1 and R2 each have a value of 50 ohms, then the current flowing through them will be 9 volts / 100 ohms = 0.09 Amps (or 90 milliamps). The voltage drop across R1 will be 50 ohms = Volts / 0.09 amps or Volts = 4.5 volts. Exactly the same calculation shows that the voltage across R2 is exactly 4.5 volts as well. However, the point to be stressed here is that it is the ratio of R1 to R2 which controls the voltage at point “A”.

If R1 has half as much resistance as R2, then half as much voltage is dropped across it as is dropped across R2, i.e. 3 Volts is dropped across R1, giving point ‘A’ a voltage of 6 Volts and that is what the voltmeter will show. Again, it does not matter what the actual value of R1 is in ohms, so long as R2 has exactly twice the resistance (shown by a higher number on the resistor).

If R1 has twice as much resistance as R2, then twice as much voltage is dropped across it as is dropped across R2, i.e. 6 Volts is dropped across R1, giving point ‘A’ a voltage of 3 Volts. Here are some examples with different resistors:

The same division of the supply voltage can be produced by positioning the slider of a variable resistor at different points by rotating the shaft of the device:
This determination of the voltage levels is the key factor to understanding electronic circuitry. The voltage levels control what currents flow and how every circuit will perform, so it is essential to understand what is happening. Stick with this section until you understand it, and if necessary, ask questions about what you find difficult.

First, please understand that a good battery is an unlimited source of voltage and that voltage does not get “used up” when a resistor or whatever is connected across it:

There can be some difficulty in understanding the “0-volt” connection in a circuit. All this means is that it is the return line for current flowing from the battery. Most conventional circuits are connected to both sides of the battery and that allows a current to flow around a closed “circuit” from one terminal of the battery to the other terminal.

It is normal practice to draw a circuit diagram so that the Plus terminal of the battery is at the top and the minus terminal is at the bottom. Many circuit diagrams show the negative line at the bottom connected to the ground or an “earth” connection, which is literally a metal rod driven into the ground to make a good electrical connection to the ground. This is done because the Earth is literally a vast reservoir of negative electricity. However, in reality, most circuits are not connected directly to the Earth in any way. The standard circuit diagram can be visualised as being like a graph of voltage, the higher up the diagram, the higher the voltage.

Anyway, when there is a circuit connected across the battery, the negative or “0V” line just indicates the return path to the battery for the current flow:
This principle applies immediately to the following circuit:

Here we encounter two new components. The first is ‘VR1’ which is a variable resistor. This device is a resistor which has a slider which can be moved from one end of the resistor to the other. In the circuit above, the variable resistor is connected across the 9 Volt battery so the top of the resistor is at +9 Volts (relative to the battery Minus terminal) and the bottom is at 0 Volts. The voltage on the slider can be adjusted from 0 Volts to 9 Volts by moving it along the resistor by turning the shaft of the component (which normally has a knob attached to it).

The second new device is ‘TR1’ a transistor. This semiconductor has three connections: a Collector, a Base and an Emitter. If the voltage on the base is below 0.7 volts, then the transistor is said to be “OFF” and in that state it has a very high resistance between the collector and the emitter, much higher than the resistance of resistor “R2”. The voltage dividing mechanism just discussed means that the voltage at the collector will therefore, be very near to 9 Volts - caused by the ratio of the transistor’s Collector/Emitter resistance compared to the resistor “R2”.

If the voltage on the base of the transistor is raised to 0.7 volts by moving the slider of the variable resistor slowly upwards, then this will feed a small current to the base which then flows out through the emitter, switching the transistor ON causing the resistance between the collector and the emitter to drop instantly to a very low value, much, much lower than the resistance of resistor ‘R2’. This means that the voltage at the collector will be very close to 0 Volts. The transistor can therefore be switched on and off just by rotating the shaft of the variable resistor:
If a bulb is used instead of R2, then it will light when the transistor switches on. If a relay or opto-isolator is used, then a second circuit can be operated:

If a buzzer is substituted for R2, then an audible warning will be sounded when the transistor switches on. If a light-dependent resistor is substituted for VR1, then the transistor will switch on when the light level increases or decreases, depending on how the sensor is connected. If a thermistor is used instead of VR1, then the transistor can be switched on by a rise or fall in temperature. The same goes for sound, wind speed, water speed, vibration level, etc. etc. - more of this later.

We need to examine the resistor circuit in more detail:

We need to be able to calculate what current is flowing around the circuit. If the circuit contains only resistors, then this can be done using “Ohm’s Law” which states that “Resistance equals Voltage divided by Current” or, if you prefer:

**Ohm’s Law (resistive Circuits only).**

“**Ohms = Volts / Amps**” which indicates the units of measurement.

In the circuit above, if the voltage is 9 Volts and the resistor is 100 ohms, then by using Ohm’s Law we can calculate the current flowing around the circuit as 100 Ohms = 9 Volts / Amps, or Amps = 9 / 100 which equals 0.09 Amps. To avoid decimal places, the unit of 1 milliamp is used. There are 1000 milliamps in 1 Amp. The
current just calculated would commonly be expressed as 90 milliamps which is written as 90 mA.

In the circuit above, if the voltage is 9 Volts and the resistor is 330 ohms, then by using Ohm’s Law we can calculate the current flowing around the circuit as $330 = 9 / \text{Amps}$. Multiplying both sides of the equation by “Amps” gives: $\text{Amps} \times 330 \text{ ohms} = 9 \text{ volts}$. Dividing both sides of the equation by 330 gives: $\text{Amps} = 9 \text{ volts} / 330 \text{ ohms}$ which works out as 0.027 Amps, written as 27 mA.

Using Ohm’s Law we can calculate what resistor to use to give any required current flow. If the voltage is 12 Volts and the required current is 250 mA then as $\text{Ohms} = \text{Volts} / \text{Amps}$, the resistor needed is given by: $\text{Ohms} = 12 / 0.25 \text{ Amps}$ which equals 48 ohms. The closest standard resistor is 47 ohms (Yellow / Purple / Black).

The final thing to do is to check the wattage of the resistor to make sure that the resistor will not burn out when connected in the proposed circuit. The power calculation is given by: $\text{Watts} = \text{Volts} \times \text{Amps}$. In the last example, this gives $\text{Watts} = 12 \times 0.25$, which is 3 Watts. This is much larger than most resistors used in circuitry nowadays.

Taking the earlier example, $\text{Watts} = \text{Volts} \times \text{Amps}$, so $\text{Watts} = 9 \times 0.027$ which gives 0.234 Watts. Again, to avoid decimals, a unit of 1 milliwatt is used, where 1000 milliwatts = 1 Watt. So instead of writing 0.234 Watts, it is common to write it as 234 mW.

This method of working out voltages, resistances and wattages applies to any circuit, no matter how awkward they may appear. For example, take the following circuit containing five resistors:

![Circuit Diagram]

As the current flowing through resistor ‘R1’ has then to pass through resistor ‘R2’, they are said to be ‘in series’ and their resistances are added together when calculating current flows. In the example above, both R1 and R2 are 1K resistors, so together they have a resistance to current flow of 2K (that is, 2,000 ohms).

If two, or more, resistors are connected across each other as shown on the right hand side of the diagram above, they are said to be ‘in parallel’ and their resistances combine differently. If you want to work out the equation above, for yourself, then choose a voltage across Rt, use Ohm’s Law to work out the current through Ra and the current through Rb. Add the currents together (as they are both being drawn from the voltage source) and use Ohm’s Law again to work out the value of Rt to confirm that the $1/Rt = 1/Ra + 1/Rb + \ldots$ equation is correct.

In the example above, R4 is 1K5 (1,500 ohms) and R5 is 2K2 (2,200 ohms) so their combined resistance is given by $1/Rt = 1/1500 + 1/2200$ or $Rt = 892 \text{ ohms}$ (using a simple calculator). Apply a common-sense check to this result: If they had been two 1500 ohm resistors then the combined value would have been 750 ohms. If they had been two 2200 ohm resistors then the combined value would have been 1100 ohms. Our answer must therefore lie between 750 and 1100 ohms. If you came up with an answer of, say, 1620 ohms, then you know straight off that it is wrong and the arithmetic needs to be done again.

So, how about the voltages at points ‘A’ and ‘B’ in the circuit? As R1 and R2 are equal in value, they will have equal voltage drops across them for any given current. So the voltage at point ‘A’ will be half the battery voltage, i.e. 6 Volts.

Now, point ‘B’. Resistors R4 and R5 act the same as a single resistor of 892 ohms, so we can just imagine two resistors in series: R3 at 470 ohms and R4+R5 at 892 ohms. Common-sense rough check: as R3 is only about half the resistance of R4+R5, it will have about half as much voltage drop across it as the voltage drop across R4+R5, i.e. about 4 Volts across R3 and about 8 Volts across R4+R5, so the voltage at point ‘B’ should work out at about 8 Volts.
We can use **Ohm's Law** to calculate the current flowing through point 'B':

\[
\text{Ohms} = \frac{\text{Volts}}{\text{Amps}}, \quad \text{or} \quad \text{Amps} = \frac{\text{Volts}}{\text{Ohms}} \quad \text{or} \quad \text{Volts} = \text{Ohms} \times \text{Amps}
\]

\[
(470 + 892) = 12 / \text{Amps}, \text{ so}
\]

\[
\text{Amps} = 12 / (470 + 892)
\]

\[
\text{Amps} = 12 / 1362 \quad \text{or}
\]

\[
\text{Amps} = 0.00881 \text{ Amps (8.81 milliamps)}.
\]

Now that we know the current passing through \((R4+R5)\) we can calculate the exact voltage across them:

\[
\text{Resistance} = \frac{\text{Volts}}{\text{Amps}} \text{ so}
\]

\[
892 = \frac{\text{Volts}}{0.00881} \quad \text{or}
\]

\[
\text{Volts} = 892 \times 0.00881
\]

\[
\text{Volts} = 7.859 \text{ Volts}.
\]

As our common-sense estimate was 8 Volts, we can accept 7.86 Volts as being the accurate voltage at point 'B'.

**The Potentiometer.**

Just before we leave the subject of resistors and move on to more interesting subjects, we come across the term ‘potentiometer’. This term is often shortened to ‘pot’ and many people use it to describe a variable resistor. I only mention this so that you can understand what they are talking about. A variable resistor is not a potentiometer and really should not be called one. You can skip the rest of this part as it is not at all important, but here is what a potentiometer is:

A fancy name for voltage is ‘potential’, so a circuit powered by a 12 Volt battery can be described as having a ‘potential’ of zero volts at the negative side of the battery and a ‘potential’ of plus twelve volts at the positive side of the battery. Ordinary folks like me would just say ‘voltage’ instead of ‘potential’.

When a voltmeter is used to measure the voltage at any point in a circuit, it alters the circuit by drawing a small amount of current from the circuit. The voltmeter usually has a high internal resistance and so the current is very small, **but** even though it is a small current, it **does** alter the circuit. Consequently, the measurement made is not quite correct. Scientists, in years gone by, overcame the problem with a very neat solution - they measured the voltage without taking any current from the circuit - neat huh? They also did it with a very simple arrangement:

![Diagram of a potentiometer setup](image)

They used a sensitive meter to measure the current. This meter is built so that the needle is in a central position if no current is flowing. With a positive current flowing, the needle deflects to the right. With a negative current flowing, the needle moves to the left. They then connected a variable resistor 'VR1' across the same battery which was powering the circuit. The top end of VR1 is at +12 Volts (they called that 'a potential of +12 Volts') and the bottom end of VR1 is at zero volts or 'a potential of zero volts'.

By moving the slider of VR1, any voltage or 'potential' from zero volts to +12 Volts could be selected. To measure the voltage at point 'A' without drawing any current from the circuit, they would connect the meter as shown and adjust the variable resistor until the meter reading was exactly zero.
Since the meter reading is zero, the current flowing through it is also zero and the current taken from the circuit is zero. As no current is being taken from the circuit, the measurement is not affecting the circuit in any way - very clever. The voltage on the slider of VR1 exactly matches the voltage at point 'A', so with a calibrated scale on the variable resistor, the voltage can be read off.

The slick piece of equipment made up from the battery, the variable resistor and the meter was used to measure the ‘potential’ (voltage) at any point and so was called a ‘potentiometer’. So, please humour me by calling a variable resistor a ‘variable resistor’ and not a ‘potentiometer’. As I said before, this is not at all important, and if you want to, you can call a variable resistor a ‘heffalump’ so long as you know how it works.

Understanding what circuit diagrams mean.

Many people look at a circuit diagram and have no idea what it means, so let’s see if we can make the mystery go away. Take this circuit for example:

```
B R LED
```

This circuit has three components plus some wire. The symbol “B” represents a battery, or more strictly speaking, a battery made up of a number of cells. Batteries come in many different shapes and sizes. Here are some of them:

- 1.5V
- 9V
- 12V

The symbol “R” represents a resistor as described above, and the “LED” is a Light-Emitting Diode which probably looks like this:

```
+ -
```

The longer lead is the Plus. Many LEDs need more than 1.5 volts to light up, and while it is very easy to think of a single AA-size battery as being 1.5 volts, the very common AA-size NiMh batteries are only 1.2 volts. So, let us set up the circuit using a 9V battery and a 330 ohm resistor (Orange, Orange, Brown) to limit the current flowing through the LED. The circuit is:

```
B R LED
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And this indicates that the Plus of the battery gets connected to the resistor. This can be done using some wire, or the resistor can be connected directly to the battery:
Then the LED gets connected to the other end of the resistor:

And finally, the other side of the LED is connected to the Minus of the battery:

If the LED is connected the wrong way round, it will not damage anything but the LED will not light up. Poor quality connections can be made by twisting wires together. Better quality connections can be made using screw connectors:

The spacing of the connectors on the strip varies with the power rating of the connectors and there are four or five sizes commonly available, and so it is sometimes necessary to cut the strip and use individual connectors at times. Another option is to use a plug-in board although they are far from perfect. They used to be very good but then integrated circuits came along with their tiny pin spacing and the boards adapted to them by making the holes and the spacing between the holes small enough to suit the integrated circuits. Now, it is no longer possible to plug in quite ordinary components such as the fast UF5408 diode as the diode wires are too large to plug into the tiny holes:
The most effective method of connection is to solder the components together and that is not particularly difficult to do. Veroboard (stripboard) is convenient and there are several other board styles which can be used. When I was very young and almost no components were available, I used drawing pins and soldered components to them, killing the excessive heat using a wet cloth which is very effective in dropping temperature very rapidly. However, no matter what method of connection is used, you just follow along the connecting lines in any diagram to see what components are connected together.

Semiconductors.
This section deals with discrete semiconductors. A later section deals with ‘Integrated Circuits’ which are large-scale semiconductor devices.

**ORP12** Light-dependent resistor. This device has a high resistance in the dark and a low resistance in bright light. It can be placed in a circuit to create a switch which operates with an increase in light level or a decrease in light level:

In this version, the voltage at point ‘A’ controls the circuit. In darkness, the ORP12 has a resistance ten times greater than that of R1 which is 12,000 ohms. Consequently, the voltage at point ‘A’ will be high. As the light level increases, the resistance of the ORP12 falls, dragging the voltage at point ‘A’ downwards. As the variable resistor ‘VR1’ is connected from point ‘A’ to the ground rail (the -ve of the battery), its slider can be moved to select any voltage between 0 Volts and the voltage of ‘A’. A slider point can be chosen to make the transistor switch off in daylight and on at night. To make the circuit trigger when the light level increases, just swap the positions of R1 and the ORP12.

The transistor shown is a BC109 although most transistors will work in this circuit. The BC109 is a cheap, silicon, NPN transistor. It can handle 100mA and 30V and can switch on and off more than a million times per second. It has three connections: the Collector, marked ‘c’ in the diagram, the Base, marked ‘b’ in the diagram and the Emitter, marked ‘e’ in the diagram.

As mentioned before, it has a very high resistance between the collector and the emitter when no current flows into the base. If a small current is fed into the base, the collector/emitter resistance drops to a very low value. The collector current divided by the base current is called the ‘gain’ of the transistor and is often called ‘hfe’. A transistor such as a BC109 or a BC108 has a gain of about 200, though this varies from actual transistor to actual
transistor. A gain of 200 means that a current of 200m\(A\) passing through the collector requires a current of 1mA through the base to sustain it. Specific information on the characteristics and connections of semiconductors of all kinds can be obtained free from the excellent website [www.alldatasheet.co.kr](http://www.alldatasheet.co.kr) which provides .pdf information files.

The BC109 transistor shown above is an NPN type. This is indicated by the arrow of the symbol pointing outwards. You can also tell by the collector pointing to the positive rail. There are similar silicon transistors constructed as PNP devices. These have the arrow in the transistor symbol pointing inwards and their collectors get connected, directly or indirectly, to the negative rail. This family of transistors are the earliest transistor designs and are called ’bi-polar’ transistors.

These silicon transistors are so efficiently constructed that they can be connected directly together to give greatly increased gain. This arrangement is called a ‘Darlington pair’. If each transistor has a gain of 200, then the pair give a gain of \(200 \times 200 = 40,000\). This has the effect that a very, very small current can be used to power a load. The following diagram shows a Darlington pair used in a water-level detector. This type of alarm could be very useful if you are asleep on a boat which starts taking on water.

Here, (when the circuit is switched on), transistor TR1 has so little leakage current that TR2 is starved of base current and is hard off, giving it a high resistance across its collector/emitter junction. This starves the buzzer of voltage and keeps it powered off. The sensor is just two probes fixed in place above the acceptable water level. If the water level rises, the probes get connected via the water. Pure water has a high electrical resistance but this circuit will still work with pure water.

The odds are that in a practical situation, the water will not be particularly clean. The resistor R1 is included to limit the base current of TR1 should the sensor probes be short-circuited. Silicon bi-polar transistors have a base/emitter voltage of about 0.7V when fully switched on. The Darlington pair will have about 1.4V between the base of TR1 and the emitter of TR2, so if the sensor probes are short-circuited together, resistor R1 will have \(6 - 1.4 = 4.6\text{V}\) across it. Ohms Law gives us the current through it as \(I = \frac{V}{R}\) or \(47,000 = \frac{4.6}{A}\) or \(A = \frac{4.6}{47,000}\) amps. This works out at 0.098mA which with a transistor gain of 40,000 would allow up to 3.9A through the buzzer. As the buzzer takes only 30mA or so, it limits the current passing through it, and TR2 can be considered to be switched hard on with the whole battery voltage across it.

NPN transistors are more common than PNP types but there is almost no practical difference between them. Here is the previous circuit using PNP transistors:

Not a lot of difference. Most of the circuit diagrams shown here use NPN types but not only are these not critical, but there are several ways to design any particular circuit. In general, the semiconductors shown in any circuit are seldom critical. If you can determine the characteristics of any semiconductor shown, any reasonably similar
device can generally be substituted, especially if you have a general understanding of how the circuit works. Either of the two previous circuits can operate as a rain detector. A suitable sensor can easily be made from a piece of strip board with alternate strips connected together to form an interlacing grid:

Here, if a raindrop bridges between any two adjacent strips, the circuit will trigger and sound a warning.

The transistors in the circuit above are connected with their emitter(s) connected to the ground rail (the lower battery line shown in any circuit is considered to be "ground" unless it is specifically shown elsewhere). This connection method is called ‘common emitter’. The following circuit uses the transistor connected in ‘emitter follower’ mode. This is where the emitter is left to follow the base voltage - it is always 0.7V below it unless the base itself is driven below 0.7V:

This is almost the same as the light-operated circuit shown earlier. In this variation, the transistors are wired so that they work as an ‘emitter-follower’ which follows the voltage at point ‘A’ which rises as the light level drops and the resistance of the ORP12 increases. This causes the voltage across the relay to increase until the relay operates and closes its contacts. A relay is a voltage-operated mechanical switch which will be described in more detail later on.

The disadvantage of the above circuit is that as the light level decreases, the current through the relay increases and it may be a significant amount of current for some considerable time. If it was intended to power the unit with a battery then the battery life would be far shorter than it need be. What we would like, is a circuit which switched rapidly from the Off state to the On state even though the triggering input varied only slowly. There are several ways to achieve this, one of them being to modify the circuit to become a ‘Schmitt Trigger’:
Here, an additional transistor (‘TR2’) has changed the circuit operation significantly, with transistor TR3 switching fully on and fully off, rapidly. This results in the current through the relay being very low until the circuit triggers.

The circuit operates as follows. When the voltage at the base of TR1 is high enough, TR1 switches on, which causes the resistance between its collector and emitter to be so low that we can treat it as a short circuit (which is a nearly-zero resistance connection). This effectively connects the 10K and 1K8 resistors in series across the battery. The voltage at their connecting point (both the collector and emitter of TR1) will then be about 1.8 Volts. The two 18K resistors are in series across that voltage so the voltage at their junction will be half that; 0.9 Volts.

This puts the Base of TR2 at about 0.9 Volts and its emitter at 1.8 Volts. The base of TR2 is therefore not 0.7 Volts above its emitter, so no base/emitter current will flow in TR2, which means that TR2 is switched hard off. This means that the TR2 collector/emitter resistance will be very high. The voltage at the base of TR3 is controlled by the 1K8 resistor, the TR2 collector/emitter resistance (very high) and the 3K9 resistor. This pushes the base voltage of TR3 up to near the full battery voltage and as it is wired as an emitter-follower, its emitter voltage will be about 0.7 Volts below that. This means that the relay will have most of the battery voltage across it and so will switch hard on.

Some practical points: The current flowing into the base of TR3 comes via the 3K9 resistor. A 3K9 resistor needs 3.9 Volts across it for every 1 mA which flows through it. If the relay needs 150 mA to operate and TR3 has a gain of 300, then TR3 will need a base current of 0.5 mA to provide 150 mA of current through its collector/emitter junction. If 0.5 mA flows through the 3K9 resistor, there will be a voltage drop across it of some 2 Volts. The TR3 base/emitter voltage will be a further 0.7 Volts, so the voltage across the relay will be about 12.0 - 2.0 - 0.7 = 9.3 Volts, so you need to be sure that the relay will work reliably at 9 Volts.

If you used a Darlington pair of transistors, each with a gain of 300, instead of TR3, then their combined base/emitter voltage drop would be 1.4 Volts, but they would only need a base current of 150 mA / (300 x 300) = 1/600 mA. That current would only drop 0.007 Volts across the 3K9 resistor, so the relay would receive 10.6 Volts.

So, how do you work out the gain of any particular transistor? The main working tool for electronics is a multimeter. This is a digital or analogue meter which can measure a wide range of things: voltage, current, resistance, ... The more expensive the meter, generally, the greater the number of ranges provided. The more expensive meters offer transistor testing. Personally, I prefer the older, passive multimeters. These are looked down on because they draw current from the circuit to which they are attached, but, because they do, they give reliable readings all the time. The more modern battery-operated digital multimeters will happily give incorrect readings as their battery runs down. I wasted two whole days, testing rechargeable batteries which appeared to be giving impossible performances. Eventually, I discovered that it was a failing multimeter battery which was causing false multimeter readings.

Transistor Testers.
For the moment, let us assume that no commercial transistor tester is to hand and we will build our own (or at least, discover how to build our own). The gain of a transistor is defined as the collector/emitter current divided by the base/emitter current. For example, if 1mA is flowing through the collector and 0.01mA is flowing into the base to sustain that collector flow, then the transistor has a gain of 100 times at 1mA. The transistor gain may vary when it is carrying different current loads. For the circuits we have been looking at so far, 1mA is a reasonable current at which to measure the transistor gain. So let’s build a circuit to measure the gain:
With the circuit shown here, the variable resistor is adjusted until a collector current of 1mA is shown on the millimetre and the gain of the transistor is then read off the scale on the variable resistor knob. The circuit is built into a small box containing the battery and with a socket into which the transistor can be plugged. The question then is, what values should be chosen for the resistor R1 and the variable resistor VR1?

Well, we might choose that the minimum gain to be displayed is 10. This would correspond to where the variable resistor slider is taken all the way up to point 'A' in the circuit diagram, effectively taking the variable resistor out of the circuit. If the transistor gain is 10 and the collector current is 1mA, then the base current will be 0.1mA. This current has to flow through the resistor R1 and it has a voltage of (9.0 - 0.7) Volts across it as the base/emitter voltage is 0.7 Volts when the transistor is on. Ohms Law gives us Ohms = Volts / Amps, which for the resistor R1 means Ohms = 8.3 / 0.0001 or 83,000 ohms, or 83K.

Rule of thumb: 1K provides 1mA if it has 1V across it, so 10K will give 0.1mA if it has 1 Volt across it. With 8.3 Volts across it, it needs to be 8.3 times larger to hold the current down to the required 0.1mA so the resistor should be 83K in size.

As 83K is not a standard size, we need to use two or more standard resistors to give that resistance. Nearest standard size below 83K is 82K, so we can used one 82K resistor and one 1K resistor in series to give the required 83K.

Suppose that we say that we would like to have 500 as the highest gain shown on our tester, then when VR1 is at its maximum value, it and R1 should provide 1/500 of the collector current of 1mA, i.e. 0.002mA or 0.000002 Amps. From Ohms Law again we get VR1 + R1 = 4,150,000 ohms or 4M15. Unfortunately, the largest value variable resistor available is 2M2 so the circuit as it stands, will not be able to cope.

Suppose we were to just use a 2M2 variable resistor for VR1, what transistor gain range could we display? Well Ohms Law ... lets us calculate the base current with 8.3 Volts across (83,000 + 2,200,000) ohms and from that the maximum transistor gain which would be 277.77 (at 1mA). You would buy a ‘linear’ standard carbon track variable resistor so that the change in resistance is steady as the shaft is rotated. The scale which you would make up would be in even steps and it would run from 10 at the minimum setting, to 278 at the highest setting.

But that is not what we wanted. We wanted to measure up to 500. But they don’t make variable resistors big enough, so what can we do? Well, if we wanted, we could lower the battery voltage, which in turn would lower the resistor values. As a 9V battery is very convenient for this kind of circuit, lets not go down that route. We could add extra circuitry to drop the 9V battery voltage down to a lower value. The most simple solution is to add an extra resistor and switch to give two ranges. If we switched in an extra 2M2 resistor above VR1 then the circuit would measure transistor gains from 278 to just over 500 and all we would need to do would be to add a second scale for the VR1 pointer knob to move over. We could, provide extra ranges which overlap and which have more convenient scales to mark. The design is up to you.
The design covered above is not the only way to measure the transistor gain. A second way, which accepts that it is not so accurate, picks a set base current and measures the collector current as a guide to the gain. In this simple method, one or more resistor values are chosen to give gain ranges, and the milliammeter used to read the corresponding gain:

Here, resistor R1 might be chosen to give a collector current of 1mA (which is a full-scale deflection on the meter) when the transistor gain is 100. Resistor R2 might be picked to give a full-scale deflection for a gain of 200, R3 for a gain of 400, R4 for a gain of 600, and so on. Generally speaking, it is not essential to know the exact gain but any reasonable approximation to it is sufficient. You are normally selecting a transistor where you need a gain of 180, so it is not important if the transistor you pick has a gain of 210 or 215 - you are only avoiding transistors with gains below 180.

How do you work out the values of the resistors R1 to R4? Well, you probably won’t expect this, but you use Ohms Law. Voltage drop is 8.3 Volts and the base current is given by the full-scale deflection’s 1mA divided by the transistor gain for each range, i.e. 1/100 mA for R1, 1/200 mA for R2,... 1/600 mA for R4,...

**Emitter Followers**

The transistor circuits show so far are known by the technical term “Common Emitter” because the emitters are generally connected to the ‘Negative rail’ or battery minus line. This method of use is very popular because when the transistor is switched on, all of the supply voltage is supplied to the load. Another common and very useful method is known as the ‘Emitter-Follower’ circuit where the load is connected to the negative rail instead of the emitter of the transistor. With this arrangement, the voltage at the emitter remains at 0.7 volts below the voltage of the transistor base and ‘follows’ that voltage no matter how it changes. Generally speaking, the transistor is being used to amplify the current which could be drawn from the point in the circuit where the transistor base is connected.

The circuit arrangement is like this:
If the battery is genuinely 12-volts, then the slider of the variable resistor VR1 can be moved from a voltage of zero volts to a voltage of +12 volts, or any desired value between those two values. That means that the voltage on the base of transistor TR1 can be any of those values. If the voltage on the transistor base is 0.7 volts or higher, then the transistor will conduct current and the voltage across the load will increase until the emitter is 0.7 volts below the base voltage. This means that the voltage across the load can be adjusted to any value from 0 volts to +11.3 volts. This circuit is known as an “Emitter-Follower” circuit.

The actual values encountered in ‘real life’ are that a battery marked as 12-volts is very seldom actually at that voltage and a common value is 12.8 volts. I have called the Base-to-Emitter voltage 0.7 volts but in reality, it may be anything from 0.6 volts to 0.75 volts. A common use for this type of circuit is to pass a constant voltage to a circuit, using a zener diode. The circuit is like this:

This circuit is supposed to have a fixed voltage at point “A” as the zener diode Z1 is supposed to produce a fixed voltage. That can work reasonably well if the battery voltage is fixed, but if the battery voltage alters either upwards or downwards, the voltage at “A” drifts, which means that the voltage across the load also alters. You will sometimes see this in constant-current circuits.

**Constant-Current circuits**

The generally recommended way to arrange a constant current flow through some load or other is to use an integrated circuit designed for the job. The arrangement is generally like this:

Here, the resistor R1 controls how much current will flow in the circuit and the resistor R2 needs to be ten times higher in value than R1. One snag is that the LM334Z drops about 4-volts when stabilising the current through the load. That is a lot of voltage sacrificed. An alternative arrangement is:
With this circuit, two ordinary diodes such as the 1N4007, are used to give a steady voltage due to the current flowing through them, supplied by resistor R1. Each diode has a voltage drop across it roughly equal to the voltage drop across the Base/Emitter junction of the transistor TR1. That means that the resistor R2 will have about the same voltage across it as one of the diodes. It is my experience that the voltage drop across the diodes is not affected much if the battery voltage changes as time goes by. The value of the resistor R2 is chosen to give the desired current flow through the load. The voltage drop across the transistor Collector/Emitter connections adjusts automatically to keep the current through the load at the constant required value.

**Substitute Transistors**

A recent question was how to find a substitute transistor for the T13009 transistor in this chapter 21 circuit, as there seemed to be no local supplier for it, and would a 2N2222 transistor do as a substitute?

That is a very reasonable question. So to answer it, we look at the circuit and we see that the Collector of the transistor is going to be pulled upwards until it exceeds the voltage of the battery chain. There are five 12-volt batteries in a chain going upwards from the transistor Emitter and while those batteries have “12 Volts” written on them, they can charge up to nearly 14 volts each. That means that the transistor collector may be dragged up to a voltage of $5 \times 14 = 70$ volts or more if the batteries are going to be charged. So, common sense says that any successful substitute transistor will have to have a voltage rating of at least 70 volts.

If we want to find out the characteristics of a transistor or diode, we can go to the [http://www.alldatasheet.com/](http://www.alldatasheet.com/) web site, although just Googling the transistor name often gets the needed information very quickly. Anyway, on the web site, the top of the page has an entry section like this:
And if you type in T13009 as the part name:

```
Part Name match □ T13009 □ Search
```

and click on the Search button, then it comes up with this:

**T13009 Datasheet, PDF**

<table>
<thead>
<tr>
<th>Shortcut</th>
<th>T13009(1) recommended result.</th>
</tr>
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<tbody>
<tr>
<td>Match, Like</td>
<td>ST13009(1)</td>
</tr>
</tbody>
</table>

So you click on the blue ST13009 link and it then comes up with a slightly confusing advertisement display which offers information on some totally unrelated component. However, if you scroll down the page a little you reach a link to the datasheet for the transistor:

<table>
<thead>
<tr>
<th>Electronic Manufacturer</th>
<th>Part no</th>
<th>Datasheet</th>
</tr>
</thead>
<tbody>
<tr>
<td>STMicroelectronics</td>
<td>ST13009</td>
<td><a href="#">Link</a></td>
</tr>
</tbody>
</table>

If you then click on the PDF symbol, you get another screen offering the actual link to the pdf file:

```
Part No.        ST13009
Download  ST13009 Click to view
```

Clicking on the link actually provides you with the datasheet which you can store locally to save ever having to go through all that lot again.

**ST13009**

*High voltage fast-switching NPN power transistor*

**Features**

- Low spread of dynamic parameters
- High voltage capability
- Minimum lot-to-lot spread for reliable operation
- Very high switching speed

**Applications**

- Switch mode power supplies
This is not an FET transistor and so our main interest is in the voltage it can withstand, the continuous current which it can carry, the peak current it can manage when fed sudden pulses, how much power overall it can handle, what DC-current gain (that is amplification) you can expect from it and how fast it can operate.

That sounds a lot but it really is quite simple. However, there is a manufacturing spread on transistors and most other electronic components, and so we are looking for just a ball-park number for these things. That is, you can have five identical-looking transistors in your hand but it is most unlikely that any two of them will actually be identical. However, let’s look at this data sheet and see what we find out:

First, the maximum voltage that the transistor can withstand with the base unconnected is 400 volts which is a good deal more that is likely to be reached in our circuit.

Next, the current. The continuous current is stated to be 12 amps and 24 amps if in pulses. That is likely to be more than the circuit needs, as a sustained output of 40 watts from a 12-volt connection is a current of under 4-amps.

Next, the wattage is stated as being 100 watts (a heat sink is definitely needed for that – imagine holding a lit 100 watt light bulb in your hand and think how comfortable that would be). However, in our circuit, the transistor will be off for most of the time and so, wattage is not likely to be a problem.

Next, the switching speed, which is likely to be important in this circuit. The data sheet suggests that about 60 nanoseconds is likely for any T13009 transistor.

And finally, the DC current gain will probably be between 15 and 39 at a current of 5 amps. It is likely to be much better than that at lower currents.

Some people have difficulty visualising how a bipolar transistor works, so let me explain it in a bit more detail. When current is flowing through a bipolar transistor, then the base voltage of that transistor is pretty much fixed. It is a bit like having a large lake with a long horizontal dam wall holding the water in the lake. When the lake water level is below that of the dam, then no water flows over the dam. If the lake level rises, then water spills over the dam. The amount of that water flow is VERY much affected by the depth of water over the dam with even a small increase in depth causing a massive increase in water flow. The same goes for the transistor base and that is why the base current flow is limited by a resistor. Without a resistor, the current flow would very quickly become many amps and burn the transistor out through sheer heating of the base/emitter junction.

The base current flow is like the setting of a valve between the collector and the emitter. If the transistor gain is 200, then 1 mA flowing into the base allows 200 mA to flow between the collector and the emitter, unless there is a load between the collector and the battery – a load which chokes off that current flow, and that is the normal case. For example, if 0.5 mA flows into the base, then a maximum of 100 mA can pass between the collector and emitter. The gain of any transistor depends on the amount of current flowing through the transistor and it varies so much that the only way to specify it properly is to draw a graph of it. Because of that, printed gain figures are given for just one or two currents. Generally, the lower the current, the higher the actual gain, so if a gain is given as 20 at 1 amp and you are only intending to have 100 mA flowing through it, then you can expect a gain much higher than 20. The voltage on the base of a single transistor which is conducting will always be 0.7 volts (or something very close to that depending on how that particular transistor was actually manufactured). That 0.7 volts stays fixed even if the current flowing into the base increases from 0.1 milliamps to 100 milliamps. So back to our T13009 transistor.

Okay, we now know a bit about the T13009 transistor, and the question asked about the 2N2222 transistor, so we look it up on the All Data Sheet web site and we find that the maximum voltage is 40-volts. That rules it out of our circuit where the voltage goes to at least 70-volts and a 2N2222 transistor would die instantly. We then look at the current and see that it has a maximum of 0.8 of an amp which means that it is really not in the ball-park for this circuit.

We know that the TIP3055 (originally packaged as the 2N3055) is very popular with free-energy builders, so we look it up and find out that it can handle voltages up to 60 volts, 90 watts of power and 15 amps of current. While it is a powerful transistor, it looks as if its voltage rating is too low for this circuit.

So, what do we do now? One way is to ask an electronics expert to suggest a suitable alternative. Another way is to look up the transistors offered by your local supplier, which for me is www.esr.co.uk which leads to this table which is one of many and which has far more entries:
We want an NPN transistor and so the MJ11016 looks possible with a 100 volt capacity, 30 amp current and 200 watt dissipation. It is a Darlington pair in a single case and so will switch on around 1.4 volts as opposed to 0.7 volts on the base, but that should not make any difference in our circuit. With a gain of 1000 a simple carbon variable resistor could be used to control the base current. There are many other transistors to choose from.

Another way to find a suitable transistor might be to go on eBay and search on “transistor” and see what transistors are popular and how much they cost. An alternative might be to try the circuit with a FET transistor such as the IRF740 which is high voltage, very powerful and not expensive. However, FET transistors trigger on voltage and draw almost no current through their “grid” connection which is the equivalent to a bipolar “base” connection and so some experimentation with the circuit may be needed.

It might also be worthwhile looking to see what transistors were chosen by Alexkor in his 5-battery circuits in chapter 6. If we do that we find the MJE13009 which has an identical specification and so is almost certainly the same as a T13009 transistor and the MJE version is readily available on eBay. Another of his transistors is the 2SC3552 transistor with 500V capability and 150 watt capacity and described as “fast acting”.

The Diode

One component which has been shown but not described is the diode or ‘rectifier’. This is a device which has a very high resistance to current flowing in one direction and a very low resistance to current flowing in the opposite direction. The base/emitter junction of a transistor is effectively a diode and, at a push, can be used as such. A proper diode is cheap to buy and has far greater voltage and current handling capacities than the base/emitter junction of a transistor.

Diodes are mainly made from one of two materials: germanium and silicon. Germanium diodes are used with very small alternating currents such as radio signals coming from an aerial. This is because a germanium diode needs only 0.2 Volts or so to carry a current while silicon needs 0.6 to 0.7 Volts (same as a silicon transistor base/emitter junction). Germanium diodes (and transistors) are very sensitive to temperature change and so are normally restricted to low power circuits. One very neat application for a silicon diode is as an ‘un-interruptible power supply’ where mains failure is caught instantly:

In this circuit, the mains voltage drives the Power Supply Unit which generates 12 Volts at point ‘A’. This provides current to the Load. The diode has +12 Volts at ‘A’ and +12 Volts at point ‘B’ so there is no voltage drop across it and it will not carry current in either direction. This means that the battery is effectively isolated when the mains is
functioning. If the Power Supply Unit output were to rise above its design level of +12 Volts, then the diode would block it from feeding current into the battery.

If the mains fails, the Power Supply Unit ("PSU") output will fall to zero. If the battery and diode were not there, the voltage at point ‘A’ would fall to zero, which would power-down the Load and possibly cause serious problems. For example, if the load were your computer, a mains failure could cause you to lose important data. With a battery back-up of this type, you would have time to save your data and shut your computer down before the battery ran out.

The circuit operates in a very simple fashion. As soon as the voltage at point ‘A’ drops to 0.7 Volts below the +12 Volts at point ‘B”, the diode starts feeding current from the battery to the Load. This happens in less than a millionth of a second, so the Load does not lose current. It would be worth adding a warning light and/or a buzzer to show that the mains has failed.

Diodes are also supplied packaged as a diode bridge, with four diodes enclosed inside. Usually intended for power supply rectification, they are not particularly fast-acting diodes, but are cheap and can carry a good deal of current. A common size is with the diodes rated at 1000 volts and able to carry 35 amps. Although there are many package types, a very common package looks like this:

![Diode Bridge Diagram]

The alternating signal is connected between two opposite corners and the pulsating DC is taken off from the other two terminals. The symbols shown above are normally marked on the flat face which is not seen in this picture. The package has a hole in the centre so that the metal case can be bolted to a heat-sink in order to keep the device reasonably cool when carrying large currents. The connections inside the package are like this:

![Diode Bridge Connections]

It is possible to connect the bridge in a different way and use it as a higher voltage double diode arrangement as shown here:
By skipping the alternating current ability and connecting to just the Plus and the Minus terminals, the package provides two pairs if diodes in connected in series. This gives twice the voltage handling in both current paths and the rated current handling capacity in both of those two paths which are now connected across each other, which doubles the current handling capacity. The diagram shows how three ordinary, cheap 1000V 35 amp bridges can be connected to give one 70 amp 6000V composite diode. You could, if you wish, raise the specification of a 1000V 35A diode bridge to 2000V 70A by using four of them like this:

Diodes are specified by their voltage handling capacity and their current-carrying capacity and the speed at which they can switch on and off. For power supplies where the frequency is very low, any diode will do, but there are circuits where the switching is needed hundreds of thousand times per second and so the diode specification sheets need to be checked to see what frequency can be handled by any particular diode. Those data sheets can be downloaded free from [http://www.alldatasheet.co.kr/](http://www.alldatasheet.co.kr/).

One other thing which needs to be checked for some circuits is the voltage needed to get the diode to switch on. Two common materials used when making diodes are silicon and germanium. Germanium types have a low forward voltage of around 0.2 volts typically which silicon has about a 0.6 volt threshold generally. These voltage figures vary enormously as the current through the diode increases. Circuits which use very low voltages need germanium diodes such as the 1N34.

**Light-Emitting Diodes.**

There is a widely used variation of the diode which is extremely useful, and that is the Light Emitting Diode or ‘LED’. This is a diode which emits light when carrying current. They are available in red, green, blue, yellow or white light versions. Some versions can display more than one colour of light if current is fed through their different electrical connections.

LEDs give a low light level at a current of about 8 or 10 mA and a bright light for currents of 20 to 30 mA. If they are being used with a 12 Volt system, then a series resistor of 1K to 330 ohms is necessary. LEDs are robust
devices, immune to shock and vibration. They come in various diameters and the larger sizes are very much more visible than the tiny ones.

**Thyristors (“SCR”s) and Triacs.**

Another version of the diode is the Silicon Controlled Rectifier or ‘Thyristor’. This device carries no current until its gate receives an input current. This is just like the operation of a transistor but the SCR once switched on, stays on even though the gate signal is removed. It stays on until the current through the SCR is forced to zero, usually by the voltage across it being removed. SCRs are often used with alternating voltages (described below) and this causes the SCR to switch off if the gate input is removed. SCRs only operate on positive voltages so they miss half of the power available from alternating power supplies. A more advanced version of the SCR is the ‘Triac’ which operates in the same way as an SCR but handles both positive and negative voltages.

**Opto-Isolators.**

Another very useful variation on the LED is the Opto-Isolator. This device is a fully enclosed LED and light-sensitive transistor. When the LED is powered up, it switches the transistor on. The big advantage of this device is that the LED can be in a low voltage, low power sensing circuit, while the transistor can be in a completely separate, high voltage, high power circuit. The opto-isolator isolates the two circuits completely from each other. It is a very useful, and very popular, low-cost device.

**Alternating Current.**

A battery provides a constant voltage. This is called a Direct Current or ‘DC’ source of power. When a circuit is connected to a battery, the positive rail is always positive and the negative rail is always negative.

If you connect a battery to a circuit through a double-pole changeover switch as shown here:

![Diagram of a changeover switch](image)

When the changeover switch is operated, the battery is effectively turned over or inverted. This circuit is called an ‘inverter’ because it repeatedly inverts the supply voltage. If the switch is operated on a regular, rapid basis, the graph of the output voltage is as shown on the right. This is a ‘square wave’ voltage and is used extensively in electronic equipment. It is called alternating current or ‘AC’ for short. SCRs and Triacs can be used conveniently with supply voltages of this type. Mains voltage is also AC but is rather different:

![Diagram of mains voltage](image)

Mains voltage varies continuously in the form of a sine wave. In Britain, the mains voltage is described as ‘240 Volts AC’ and it cycles up and down 50 times per second, i.e. 50 positive peaks and 50 negative peaks in one
second. It would be reasonable to assume that each voltage peak would be 240 Volts but this is not the case. Even though the supply is described as 240 Volts, it peaks at the square root of 2 times greater than that, i.e. 339.4 Volts. The actual supply voltage is not particularly accurate, so any device intended for mains use should be rated to 360 Volts. In America, the supply voltage is 110 Volts AC and it cycles 60 times per second, peaking at plus and minus 155 Volts. Later on, you will see how one or more diodes can be used to convert AC to DC in a unit which is sold as a ‘mains adapter’ intended to allow battery operated equipment be operated from the local mains supply.

**Coils (“Inductors”) and Solenoids.**

If you take a cardboard tube, any size, any length, and wind a length of wire around it, you create a very interesting device. It goes by the name of a ‘coil’ or an ‘inductor’ or a ‘solenoid’.

This is a very interesting device with many uses. It forms the heart of a radio receiver, it used to be the main component of telephone exchanges, and most electric motors use several of them. The reason for this is if a current is passed through the wire, the coil acts in exactly the same way as a bar magnet:
The main difference being that when the current is interrupted, the coil stops acting like a magnet, and that can be very useful indeed. If an iron rod is placed inside the coil and the current switched on, the rod gets pushed to one side. Many doorbells use this mechanism to produce a two-note chime. A ‘relay’ uses this method to close an electrical switch and many circuits use this to switch heavy loads (a thyristor can also be used for this and it has no moving parts).

A coil of wire has one of the most peculiar features of almost any electronic component. When the current through it is altered in any way, the coil opposes the change. Remember the circuit for a light-operated switch using a relay?:

![Circuit Diagram](image)

You will notice that the relay (which is mainly a coil of wire), has a diode across it. Neither the relay nor the diode were mentioned in any great detail at that time as they were not that relevant to the circuit being described. The diode is connected so that no current flows through it from the battery positive to the ‘ground’ line (the battery negative). On the surface, it looks as if it has no use in this circuit. In fact, it is a very important component which protects transistor TR3 from damage.

The relay coil carries current when transistor TR3 is on. The emitter of transistor TR3 is up at about +10 Volts. When TR3 switches off, it does so rapidly, pushing the relay connection from +10 Volts to 0 Volts. The relay coil reacts in a most peculiar way when this happens, and instead of the current through the relay coil just stopping, the voltage on the end of the coil connected to the emitter of TR3 keeps moving downwards. If there is no diode across the relay, the emitter voltage is forced to briefly overshoot the negative line of the circuit and gets dragged down many volts below the battery negative line. The collector of TR3 is wired to +12 Volts, so if the emitter gets dragged down to, say, -30 Volts, TR3 gets 42 Volts placed across it. If the transistor can only handle, say, 30 Volts, then it will be damaged by the 42 Volt peak.

The way in which coils operate is weird. But, knowing what is going to happen at the moment of switch-off, we deal with it by putting a diode across the coil of the relay. At switch-on, and when the relay is powered, the diode has no effect, displaying a very high resistance to current flow. At switch-off, when the relay voltage starts to plummet below the battery line, the diode effectively gets turned over into its conducting mode. When the voltage reaches 0.7 Volts below the battery negative line, the diode starts conducting and pins the voltage to that level until the voltage spike generated by the relay coil has dissipated. The more the coil tries to drag the voltage down, the harder the diode conducts, stifling the downward plunge. This restricts the voltage across transistor TR3 to 0.7 Volts more than the battery voltage and so protects it.

Solenoid coils can be very useful. Here is a design for a powerful electric motor patented by the American, Ben Teal, in June 1978 (US patent number 4,093,880). This is a very simple design which you can build for yourself if you want. Ben’s original motor was built from wood and almost any convenient material can be used. This is the top view:
And this is the side view:

Ben has used eight solenoids to imitate the way that a car engine works. There is a crankshaft and connecting rods, as in any car engine. The connecting rods are connected to a slip-ring on the crankshaft and the solenoids are given a pulse of current at the appropriate moment to pull the crankshaft round. The crankshaft receives four pulls on every revolution. In the arrangement shown here, two solenoids pull at the same moment.

In the side view above, each layer has four solenoids and you can extend the crankshaft to have as many layers of four solenoids as you wish. The engine power increases with every layer added. Two layers should be quite adequate as it is a powerful motor with just two layers.

An interesting point is that as a solenoid pulse is terminated, its pull is briefly changed to a push due to the weird nature of coils. If the timing of the pulses is just right on this motor, that brief push can be used to increase the power of the motor instead of opposing the motor rotation. This feature is also used in the Adams motor described in the ‘Free-Energy’ section of this document.
The strength of the magnetic field produced by the solenoid is affected by the number of turns in the coil, the current flowing through the coil and the nature of what is inside the coil ‘former’ (the tube on which the coil is wound). In passing, there are several fancy ways of winding coils which can also have an effect, but here we will only talk about coils where the turns are wound side by side at right angles to the former.

1. Every turn wound on the coil, increases the magnetic field. The thicker the wire used, the greater the current which will flow in the coil for any voltage placed across the coil. Unfortunately, the thicker the wire, the more space each turn takes up, so the choice of wire is somewhat of a compromise.

2. The power supplied to the coil depends on the voltage placed across it. Watts = Volts x Amps so the greater the Volts, the greater the power supplied. But we also know from Ohm’s Law that Ohms = Volts / Amps which can also be written as Ohms x Amps = Volts. The Ohms in this instance is fixed by the wire chosen and the number of turns, so if we double the Voltage then we double the current.

For example: Suppose the coil resistance is 1 ohm, the Voltage 1 Volt and the Current 1 Amp. Then the power in Watts is Volts x Amps or 1 x 1 which is 1 Watt.

Now, double the voltage to 2 Volts. The coil resistance is still 1 ohm so the Current is now 2 Amps. The power in Watts is Volts x Amps or 2 x 2 which is 4 Watts. Doubling the voltage has quadrupled the power.

If the voltage is increased to 3 Volts. The coil resistance is still 1 ohm so the Current is now 3 Amps. The power in Watts is Volts x Amps or 3 x 3 which is 9 Watts. The power is Ohms x Amps squared, or Watts = Ohms x Amps x Amps. From this we see that the voltage applied to any coil or solenoid is critical to the power developed by the coil.

3. What the coil is wound on is also of considerable importance. If the coil is wound on a rod of soft iron covered with a layer of paper, then the magnetic effect is increased dramatically. If the rod ends are tapered like a flat screwdriver or filed down to a sharp point, then the magnetic lines of force cluster together when they leave the iron and the magnetic effect is increased further.

If the soft iron core is solid, some energy is lost by currents flowing round in the iron. These currents can be minimised by using thin slivers of metal (called ‘laminations’) which are insulated from each other. You see this most often in the construction of transformers, where you have two coils wound on a single core. As it is convenient for mass production, transformers are usually wound as two separate coils which are then placed on a figure-of-eight laminated core.

However, while all that information is a useful, gentle introduction to what an inductor is, it does not convey the most important feature of a coil, which is that every coil stores energy when it is connected to a power source and it returns almost all of that energy when disconnected from the power source. The return of the stored energy happens in a very short period of time and that feature can produce powerful systems if you have the expertise to capture and use that power.

For example, it is not unusual for a simple 12-volt system to generate a rapid series of 400-volt pulses which can be used to recondition and charge car batteries. There are many examples of this in chapter 6.

Paul Babcock (www.paulmariobabcock.com) destroyed more than a thousand transistors when developing his magnetic motor system as the return of coil energy is so fast that it produces high current flows, and if the capacitor into which the current return is being fed is of a low capacity, voltages higher than the supply voltage are produced. For the last hundred years or so, this sort of information has been suppressed, so take what is said in standard textbooks as being a mixture of half truths and downright lies.

As ‘Kone’ has demonstrated, if you short-circuit a powered coil, it causes multiple magnetic pulses as the power in the coil oscillates backwards and forwards through the closed circuit containing the coil:
Magnetism is a field which has not been taught or generally researched for many decades. It is not a simple subject. The magnetic strength produced by any coil increases as the number of turns in the coil increases (if the current flowing through the coil remains the same). That means that a coil with many turns can produce a higher magnetic field at a lower current than a high-current coil with few turns. However, other coil characteristics are also altered. The power loss due to the resistance of the wire in the coil increases with increased turns as they need a longer length of wire. That power loss results in the coil heating up when in use. The speed with which the magnetic field develops and decays is slower for a coil with many turns. Surprisingly, because of this, the best coil for many jobs ends up having relatively few turns.

Transformers.

Transformers are used to alter the voltage of any alternating current power source. If the alteration increases the output voltage, then the transformer is called a ‘step-up’ transformer. If the output voltage is lower than the input voltage then it is called a ‘step-down’ transformer. If the voltages are the same, it is called an ‘isolation’ transformer. A common construction looks like this:

The Coil bobbin sits on the section of the laminations marked ‘A’ above. The coil is wound on its bobbin former, first one winding and then the second winding. The bobbin is then placed on the central part of the ‘E’ shaped laminations and then completely surrounded by the laminations when the crossbar is placed on the top. The mounting strap is used to hold the two sets of laminations together and provide mounting lugs for attaching the transformer to a chassis. There are typically, twenty laminations in each set and every lamination is insulated from the adjoining laminations.

If you want to change the voltage of a battery supply, it is possible to build an electronic circuit to generate an alternating voltage and then use a transformer to change that alternating voltage to whatever voltage you want. The most common form of this, is for generating mains voltage from a 12 Volt car battery, so that mains equipment can be run in remote locations, such as boats, caravans, etc. These circuits are called ‘inverters’ and they are very popular pieces of equipment. The voltage in the secondary coil of any transformer is determined by the ratio of the turns in the primary and secondary windings.

For example; if there is a 10 Volt alternating voltage available and you have a transformer which has 100 turns in the primary coil and 1000 turns in the secondary coil. If you connect the 10 Volts across the primary, there will be 100 Volts generated across the secondary coil.
Instead, if you connect the 10 Volts across the secondary coil, a voltage of 1 Volts will be generated across the primary winding. This is because there is a 10:1 ratio between the two windings. The Law of Conservation of Energy applies to transformers as it does to everything else. The power input to the primary winding will be the same as the power in the secondary winding minus the losses. The losses, in this case, will be a temperature rise of the whole transformer. If the current passed through the transformer is well below its rated capacity, then the losses will be small. The important point is that 10 Volts at 1 Amp into the primary winding will generate 100 Volts in the secondary, but at somewhat less than 0.1 Amps: Power Input is 10 Watts and Power Output is almost 10 Watts. The voltage has been raised to 100 Volts but the potential current draw has been reduced from 1 Amp to 0.1 Amps (100 mA).

In practice, the thickness of the wire used in the windings is very important. If the voltage to be placed across the winding is high, then the wire diameter will be small. Coil windings have fairly low resistances but this is not critical in circuits as coils operate in a peculiar way. Coils have AC ‘impedance’ in addition to their DC ‘resistance’. While Direct Current (from a battery, say) can flow quite easily through a coil with low resistance, Alternating Current may have a hard job getting through the coil due to its high ‘impedance’. Sometimes, coils are used to choke off any AC ripple (interference) coming along a DC power cable. When a coil is used for this purpose it is called a ‘choke’. Each coil has its own resonant frequency and at that frequency it is very difficult for AC to get through the coil. Crystal set radios work on that principle:

![Crystal Set Radio Diagram](image)

Here, the aerial picks up every radio station broadcasting in the area. These are all at different frequencies and they all head down the aerial wire, looking for the easiest path to the earth connection. Most of them run through the coil with no problem whatsoever. If the resonant frequency of the coil matches the frequency of one of the radio stations, then that radio signal (and only that signal) finds it very hard to get through the coil and looks for an easier path to earth. The next easiest path is through the diode and the headphones, so the signal goes that way. The diode blocks part of the signal which generates the sound of the radio broadcast in the headphones.

This system works very well indeed if there is a good radio signal. A germanium diode is used as the radio signal voltage is very small and a germanium diode operates on 0.2 Volts while a silicon diode needs 0.7 Volts to operate. That difference is significant at these very low voltages. The resonant frequency of the coil depends on the number of turns in the coil. In this design, the coil has a slider which allows the number of turns to be altered and so, different radio stations to be tuned in.

**Rectification and Power Supplies.**

We now have the question of how do we turn an alternating voltage into a constant ‘direct’ voltage. The crystal radio set operates by chopping off half of the alternating radio signal. If we were to do this to the output from a mains transformer with an output of say, 12 Volts AC, the result is not very satisfactory:
Here, we have the situation shown in the upper diagram. The output consists of isolated pulses at 50 per second. You will notice that there is no output power for half of the time. The negative part of the waveform is blocked by the high resistance of the diode while the positive part of the waveform is allowed through by the low resistance of the ‘forward-biased’ diode. It should be remembered that the diode drops 0.7 Volts when conducting so the output of the half-wave rectified transformer will be 0.7 Volts lower than the transformer’s actual output voltage.

If four diodes are used instead of one, they can be arranged as shown in the lower diagram. This arrangement of diodes is called a ‘bridge’. Here the positive part of the waveform flows through the upper blue diode, the load ‘L’ and on through the lower blue diode. The negative part flows through the left hand red diode, the load and then the right hand red diode. This gives a much better output waveform with twice the power available. The output voltage will be 1.4 Volts less than the transformer output voltage as there are two silicon diodes in the supply chain.

The output from even the full-wave rectifier is still unsatisfactory as there is a voltage drop to zero volts 100 times per second. Only a few devices operate well with a power supply like that, an incandescent bulb as used in a car can use this output, but then, it could use the original AC supply without any rectification. We need to improve the output by using a reservoir device to supply current during those moments when the voltage drops to zero. The device we need is a **Capacitor** which used to be called a ‘condenser’. The circuit of a mains unit using a capacitor is shown here:
This produces a much better result as the capacitor stores some of the peak energy and gives it out when the voltage drops. If the load on the unit is light with not very much current taken from it, the output voltage is quite good. However, if the current drain is increased, the output voltage gets dragged down 100 times per second. This voltage variation is called ‘ripple’ and if the unit is supplying an audio system or a radio, the ripple may well be heard as an annoying hum. The larger the capacitor for any given current draw, the smaller the ripple.

To improve the situation, it is normal to insert an electronic control circuit to oppose the ripple:

This circuit uses one new component, a new variety of diode called a ‘Zener’ diode. This device has an almost constant voltage drop across it when its current-blocking direction breaks down. The diode is designed to operate in this state to provide a reference voltage. The circuit merely uses a tiny current from the top of the zener diode to drive the Darlington pair emitter-follower transistors used to provide the output current.

With this circuit, when the output current is increased, the resistance of the transistor pair automatically reduces to provide more current without varying the output voltage. The 1K resistor is included to give the transistors a completed circuit if no external equipment is connected across the output terminals. The zener diode is chosen to give 1.4 Volts more than the required output voltage as the two transistors drop 1.4 Volts when conducting.

You should note that the output transistor is dropping 6 Volts at the full supply current. Watts = Volts x Amps so the power dissipated by the transistor may be quite high. It may well be necessary to mount the transistor on an aluminium plate called a ‘heat sink’ to keep it from overheating. Some power transistors, such as the 2N3055, do
not have the case isolated from the active parts of the transistor. It is good practice to use a mica gasket between the transistor and the heat-sink as it conducts then heat without making an electrical connection to the metal heat-sink.

A capacitor, being an electrical reservoir, can be used as part of a timer circuit. If the current flow into it is restricted by passing it through a resistor. The length of time between starting the flow on an empty capacitor, and the voltage across the capacitor reaching some chosen level, will be constant for a high-quality capacitor.

![Graph showing voltage over time]

As the voltage increase tails off, it becomes more difficult to measure the difference accurately, so if the capacitor is to be used for generating a time interval, it is normal to use the early part of the graph area where the line is fairly straight and rising fast.

**The Voltage Doubler.**

It is possible to increase the output voltage of a transformer although this does reduce its ability to supply current at that voltage. The way that this is done is to feed the positive cycles into one storage capacitor and the negative cycles into a second reservoir capacitor. This may sound a little complicated, but in reality, it isn't. A circuit for doing this is shown here:

![Circuit diagram for voltage doubler]

With this circuit, the transformer output is some voltage, say "V" volts of AC current. This output waveform is fed to capacitor "C1" through diode "D1" which lops off the negative part of the cycle. This produces a series of positive half-cycles which charge up capacitor "C1" with a positive voltage of "V".

The other half of the output is fed to capacitor "C2" through diode "D2" which cuts off the positive part of the cycle, causing capacitor "C2" to develop a voltage of -V across it. As the two capacitors are 'in series' and not placed across each other, their voltages add up and produce twice the transformer output voltage.

A word of warning here. The transformer is producing an AC waveform and these are marked with the average voltage of the waveform, which is usually a sine wave. The peak voltage of a sinewave is 41% greater than this, so if your transformer has an AC output of 10 volts, then the peaks fed to the capacitors will be about 14.1 volts. If there is no current draw from the capacitors (that is, with the load switched off), then each capacitor will charge to this 14.1 volts and the overall output voltage will be 28.2 volts and not the 20 volts which you might expect. You need to understand that as this is only a half-wave supply, there will be considerable ripple on the output voltage if the current draw is high.

Using one additional smoothing capacitor and paying attention to the voltage ratings of the capacitors, the 28 volts supply circuit might be like this:
Multivibrators: The Bistable.
The number of electronic circuits which can be built with basic components such as resistors, capacitors, transistors, coils, etc. is limited only by your imagination and needs. Here is a circuit where two transistors operate as a pair:

This circuit has two stable states and so it is called a “bi” “stable” or “bistable” circuit. It is important to understand the operation of this simple and useful circuit.

If press-button switch ‘A’ is pressed, it short-circuits the base/emitter junction of transistor TR1. This prevents any current flowing in the base/emitter junction and so switches TR1 hard off. This makes the voltage at point ‘C’ rise as high as it can. This leaves transistor TR2 powered by R1 and R2 which have 11.3 Volts across them and switches TR2 hard on.

This pulls point ‘D’ down to about 0.1 Volts. This happens in less than a millionth of a second. When the press-button switch ‘A’ is released, transistor TR1 does not switch on again because its base current flows through resistor R3 which is connected to point ‘D’ which is far, far below the 0.7 Volts needed to make TR1 start conducting.

The result is that when press-button ‘A’ is pressed, transistor TR2 switches on and stays on even when press-button ‘A’ is released. This switches transistor TR3 off and starves the Load of current. This is the first ‘stable state’.

The same thing happens when press-button ‘B’ is pressed. This forces transistor TR2 into its ‘off’ state, raising point ‘D’ to a high voltage, switching transistor TR3 hard on, powering the Load and holding transistor TR1 hard off. This is the second of the two ‘stable states’.

In effect, this circuit ‘remembers’ which press-button was pressed last, so millions of these circuits are used in computers as Random Access Memory (‘RAM’). The voltage at point ‘C’ is the inverse of the voltage at point ‘D’, so if ‘D’ goes high then ‘C’ goes low and if ‘D’ goes low, then ‘C’ goes high. In passing, the output at ‘D’ is often called ‘Q’ and the output at ‘C’ is called ‘Q-bar’ which is shown as the letter Q with a horizontal line drawn above it. This is shown on the next circuit diagram.

A minor variation of this circuit allows a load to be energised when the circuit is powered up:
When powered down, the capacitor 'C1' in this circuit is fully discharged through resistor 'R6'. When the 12 Volts supply is connected to the circuit, capacitor C1 does not charge instantly and so holds the base of TR2 down below 0.7 Volts for much longer than it takes for transistor TR1 to switch on (which, in turn, holds TR2 hard off). Mind you, if it is not necessary to have the Load held powered on indefinitely, then an even more simple circuit can do this:

Here, when the switch is closed, both sides of the capacitor C1 are at +12 Volts and this causes the 1K8 resistor to conduct heavily, driving the transistor and powering the load. The capacitor charges rapidly through the transistor and reaches the point at which it can no longer keep the transistor switched on. When the battery is switched off, the 1M resistor discharges the capacitor, ready for the next time the battery is connected.

the Monostable Multivibrator.

The monostable has one stable state and one unstable state. It can be flipped out of its stable state but it will 'flop' back into its stable state. For that reason, it is also known as a 'flip-flop' circuit. It is similar to a bistable circuit, but one of the cross-link resistors has been replaced by a capacitor which can pass current like a resistor, but only for a limited amount of time, after which, the capacitor becomes fully charged and the current flow stops, causing the 'flop' back to the stable state once more.

In this circuit, the 'R' resistor and the 'C' capacitor values determine how long the monostable will be in its unstable state. The circuit operates like this:

1. In the stable state, transistor TR1 is off. Its collector voltage is high, pushing the left hand side of capacitor 'C' to near +12 Volts. As the right hand side of capacitor 'C' is connected to the base of TR2 which is at 0.7 Volts, the capacitor gets charged to about 11.3 Volts.
2. The press-button switch is operated briefly. This feeds current through its 10K resistor to the base of transistor TR1, switching it hard on. This drops the collector voltage of TR1 to near 0 Volts, taking the left hand side of the capacitor with it.

3. As the voltage across a capacitor can’t change instantly, the right hand side of the capacitor drives the base of transistor TR2 down below 0.7 Volts, causing TR2 to switch off.

4. The circuit can’t hold TR2 in its ‘off’ state for ever. The resistor ‘R’ feeds current into the capacitor, forcing the voltage at the base of TR2 steadily upwards until the voltage reaches 0.7 Volts and transistor TR2 switches on again, forcing TR1 off again (provided that the press-button switch has been released). This is the stable state again. If the press-button switch is held on, then both transistors will be on and the output voltage will still be low. Another output pulse will not be generated until the press-button is let up and pressed again.

This circuit could be used to switch a microwave oven on for any chosen number of seconds, create a delay on your home-built burglar alarm, to give you time to switch it off after walking through your front door, operate a solenoid valve to feed a pre-determined quantity of beverage into a bottle on a production line, or whatever...

The Astable Multivibrator.
The astable circuit is the monostable with a second capacitor added so that neither state is stable. This results in the circuit flopping backwards and forwards continuously:

![Astable Multivibrator Circuit Diagram]

The rate of switching is controlled by the R1/C1 and R2/C2 combinations. The load’s ON time to its OFF time is called the ‘mark-space’ ratio, where the ON period is the ‘mark’ and the OFF period is the ‘space’. If you choose to use electrolytic capacitors which have their own polarity, then the +ve end of each capacitor is connected to the transistor collector.

While it is good to understand how these multivibrator circuits operate and can be built, nowadays there are pre-built circuits encased in a single package which you are much more likely to choose to use. These are called Integrated Circuits or ‘ICs’ for short. We will be discussing these shortly. Before we do, notice that in the circuit above, transistor TR3 has been changed to a new variety called a Field Effect Transistor (‘FET’). This type of transistor is newer than the ‘bipolar’ transistors shown in the earlier circuits. FETs come in two varieties: ‘n-channel’ which are like NPN transistors and ‘p-channel’ which are like PNP transistors.

FETs are more difficult to make but have now reached a level of cost and reliability which makes them very useful indeed. They require almost no base current (called ‘gate’ current with this type of transistor) which means that they have almost no effect on any circuit to which they are attached. Also, many of them can handle large currents and boast major power handling capabilities. Because of this, it is usual to see them packaged with a metal plate mounting, ready to be bolted to an aluminium heat-sink plate to help dissipate the heat generated by the large amount of power flowing through them. The ‘RFP50N06’ shown above can handle up to 50 Volts and carry up to 60 Amps, which is serious power handling.

Inverters and Truth Tables.
Consider the following circuit:
If neither of the press-button switches are operated, the transistor has no base/emitter current flow and so it is off. This places the collector voltage at ‘C’ near the positive rail (+5 Volts).

If press-button switch ‘A’ is operated, the base voltage tries to rise to half of the battery voltage but doesn’t make it because the transistor base pins it down to 0.7 Volts. This feeds base current to the transistor, switching it hard on and causing the output at ‘C’ to drop to nearly 0 Volts.

If press-button switch ‘B’ is operated (don’t do this when switch ‘A’ is closed or you will get a very high ‘short-circuit’ current flowing directly through the two switches) it has no effect on the output voltage which will stay high.

If we re-draw the circuit like this:

We can see that if the voltage at the input ‘A’ is taken high, then the output voltage at ‘C’ will be low. If the voltage at the input ‘A’ is taken low, then the output voltage at ‘C’ will be high. A circuit which does this is called an ‘Inverter’ because it ‘inverts’ (or ‘turns upside down’) the input voltage.

We can summarise this operation in a table. Personally, I would call the table an ‘Input/Output’ table, but for no obvious reason, the standard name is a ‘Truth’ table. The purpose of this table is to list all of the possible inputs and show the corresponding output for each input.

Another standard, is to substitute ‘1’ for ‘High Voltage’ and ‘0’ for ‘Low Voltage’. You will notice that many items of electrical and electronic equipment have these symbols on the ON / OFF switch. In computer circuitry (hah! you didn’t notice that we had moved to computer circuits, did you?), the ‘0’ represents any voltage below 0.5 Volts and the ‘1’ represents any voltage above 3.5 Volts. Many, if not most, computers operate their logic circuits on 5 Volts. This Inverter circuit is a ‘logic’ circuit.

A criticism of the above circuit is that its input resistance or ‘impedance’ is not particularly high, and its output impedance is not particularly low. We would like our logic circuits to be able to operate the inputs of eight other logic circuits. The jargon for this is that our circuit should have a ‘fan-out’ of eight.

Let’s go for a simple modification which will improve the situation:
Here, the input impedance has been increased by a factor of 100 by using a Darlington pair of transistors which need far less base current, and so can have a much higher input resistor.

Unfortunately, the output impedance is still rather high when the transistors are in their OFF state as any current taken from the positive line has to flow through the 1K8 (1800 ohm) resistor. But we need this resistor for when the transistors are in their ON state. We really need to change the 1K8 resistor for some device which has a high resistance at some times and a low resistance at other times. You probably have not heard of these devices, but they are called ‘transistors’.

There are several ways to do this. We might choose to use PNP transistors (we normally use NPN types) and connect these in place of the 1K8 resistor. Perhaps we might use a circuit like this:

This circuit is starting to look complicated and I don’t like complicated circuits. It is not as bad as it looks. The NPN transistors at the bottom are almost the same as the previous circuit. The only difference is that the collector load is now two 100 ohm resistors plus the resistance of the two transistors. If the PNP transistors are OFF when the NPN transistors are ON, then the circuit loading on the NPN transistors will be negligible and the whole of the NPN transistors output will be available for driving external circuits through the lower 100 ohm resistor (a large ‘fan-out’ for the ‘0’ logic state). To make sure that the PNP transistors are hard off before the NPN transistors start to switch on, the resistor ‘R1’ needs to be selected carefully.

The PNP transistors are an exact mirror image of the NPN side, so resistor R2 needs to be selected carefully to ensure that the NPN transistors are switched hard OFF before the PNP transistors start to switch ON.

You need not concern yourself unduly with that circuit, because you will almost certainly use an Integrated Circuit rather than building your own circuit from ‘discrete’ components. An Integrated Circuit containing six complete inverters is the 7414 which is shown above. This comes in a small black case with two rows of 7 pins which make it look a bit like a caterpillar. Because there are two row of pins, the packaging is called “Dual In-Line” or “DIL” for short.

Now, consider the following circuit:
This circuit operates the same way as the Inverter circuit, except that it has two inputs ('A' and 'B'). The output voltage at 'C' will be low if either, 'A' OR 'B' or both, of the inputs is high. The only time that the output is high, is when both Input 'A' and Input 'B' are low. Consequently, the circuit is called an “OR” gate. Strictly speaking, because the output voltage goes Down when the input voltage goes Up, it is called a “Not OR” gate, which gets shortened to a “NOR” gate. In this context, the word “not” means “inverted”. If you fed the output 'C' into an inverter circuit, the resulting circuit would be a genuine “OR” gate. The digital circuit symbols for an AND gate, a NAND gate, an OR gate and a NOR gate are:

Gating.
These common chips are usually supplied with 2, 4 or 8 inputs. So, why is it called a “Gate” - isn't it just a double inverter? Well, yes, it is a double inverter, but a double inverter acts as a gate which can pass or block an electronic signal. Consider this circuit:

Here, transistors 'TR1' and 'TR2' are connected to form an astable (multivibrator). The astable runs freely, producing the square wave voltage pattern shown in red. Transistor 'TR3' passes this voltage signal on. TR3 inverts the square wave, but this has no practical effect, the output being the same frequency square wave as the signal taken from the collector of TR2.

If the press-button switch at point ‘A’ is operated, a current is fed to the base of TR3 which holds it hard on. The voltage at point ‘C’ drops to zero and stays there. The square wave signal coming from the collector of TR2 is blocked and does not reach the output point ‘C’. It is as if a physical ‘gate’ has been closed, blocking the signal from reaching point ‘C’. As long as the voltage at point ‘A’ is low, the gate is open. If the voltage at point ‘A’ goes high, the gate is closed and the output is blocked.

There is no need for a manual switch at point ‘A’. Any electronic switching circuit will do:
Here, a slow-running astable is substituted for the manual switch. When the output voltage of ‘Astable 2’ goes high, it switches the gate transistor ‘TR3’, holding it hard on and blocking the square-wave signal from ‘Astable 1’. When the output voltage of ‘Astable 2’ goes low, it frees transistor ‘TR3’ and it then passes the ‘Astable 1’ signal through again. The resulting gated waveform is shown in red at point ‘C’ and it is bursts of signal, controlled by the running rate of ‘Astable 2’. This is the sort of waveform which Stan Meyer found very effective in splitting water into Hydrogen and Oxygen (see Chapter 10).

This circuit could also be drawn as:

The small circle on the output side of logic devices is to show that they are inverting circuits, in other words, when the input goes up, the output goes down. The two logic devices we have encountered so far have had this circle: the Inverter and the NAND gate.

If you wish, you can use a NAND gate chip which has the circuitry also built as a Schmitt trigger, which as you will recall, has a fast-switching output even with a slowly moving input. With a chip like that, you can get three different functions from the one device:
If the two inputs of a NAND gate are connected together, then the output will always be the opposite of the input, i.e. the gate acts as an inverter. This arrangement also works as a Schmitt Trigger due to the way the NAND gate circuitry is built. There are several packages built with this type of circuitry, the one shown here is the “74132” chip which contains four “dual-input” NAND gates. Gates can have almost any number of inputs but it is rare to need more than two in any given circuit. Another chip with identical pin connections is the 4011 chip (which is not a Schmitt circuit). This ‘quad dual-input’ NAND gate package uses a construction method called “CMOS” which is very easily damaged by static electricity until actually connected into a circuit. CMOS chips can use a wide range of voltages and take very little current. They are cheap and very popular.

The number of devices built into an Integrated Circuit is usually limited by the number of pins in the package and one pin is needed for one connection to the ‘outside world’. Packages are made with 6 pins (typically for opto-isolators), 8 pins (many general circuits), 14 pins (many general circuits, mostly computer logic circuits), 16 pins (ditto, but not as common) and then a jump to large numbers of pins for Large Scale devices such as microprocessors, memory chips, etc. The standard IC package is small:

Prototype circuits are often built on ‘strip board’ which is a stiff board with strips of copper running along one face, and punched with a matrix of holes. The strips are used to make the electrical connections and are broken where necessary. This strip board is usually called “Veroboard”:

Nowadays, the strip board holes are spaced 2.5 mm (1/10”) apart which means that the gaps between the copper strips is very small indeed. I personally, find it quite difficult to make good solder joints on the strips without the solder bridging between two adjacent strips. Probably, a smaller soldering iron is needed. I need to use an 8x magnifying glass to be sure that no solder bridging remains in place before a new circuit is powered up for the first time.
time. Small fingers and good eyesight are a decided advantage for circuit board construction. The narrow spacing of the holes is so that the standard IC DIL package will fit directly on the board.

Circuits built using computer circuitry, can experience problems with mechanical switches. An ordinary light switch turns the light on and off. You switch it on and the light comes on. You switch it off and the light goes off. The reason it works so well is that the light bulb takes maybe, a tenth of a second to come on. Computer circuits can switch on and off 100,000 times in that tenth of a second, so some circuits will not work reliably with a mechanical switch. This is because the switch contact bounces when it closes. It may bounce once, twice or several times depending on how the switch is operated. If the switch is being used as an input to a counting circuit, the circuit may count 1, 2 or several switch inputs for one operation of the switch. It is normal to “de-bounce” any mechanical switch. This could be done using a couple of NAND gates connected like this:

The NAND Latch.

Here, the mechanical switch is buffered by a ‘latch’. When the ‘Set’ switch is operated, the output goes low. The unconnected input of gate ‘1’ acts as if it has a High voltage on it (due to the way the NAND gate circuit was built). The other input is held low by the output of gate ‘2’. This pushes the output of gate ‘1’ high, which in turn, holds the output of gate ‘2’ low. This is the first stable state.

When the ‘Set’ switch is operated, the output of gate ‘2’ is driven high. Now, both inputs of gate ‘1’ are high which causes its output to go low. This in turn, drives one input of gate ‘2’ low, which holds the output of gate ‘2’ high. This is the second stable state.

To summarise: pressing the ‘Set’ switch any number of times, causes the output to go low, once and only once. The output will stay low until the ‘Reset’ switch is operated once, twice or any number of times, at which point the output will go high and stay there.

This circuit uses just half of one cheap NAND gate chip to create a bistable multivibrator which is physically very small and light.

Gate Circuits.
NAND Gates can be used as the heart of many electronic circuits apart from the logic circuits for which the package was designed. Here is a NAND gate version of the rain alarm described earlier. The ‘4011B’ chip is a CMOS device which has a very high input impedance and can operate at convenient battery voltages (3 to 15 Volts):
This circuit is comprised of a rain sensor, two astable multivibrators and a power-driver feeding a loudspeaker:

1. The rain sensor is a wired-up strip board or similar grid of interlaced conductors, forming a voltage-divider across the battery rails.

2. The output voltage from this, at point ‘A’ in the circuit diagram, is normally low as the strip board is open-circuit when dry. This holds the first NAND gate locked in the OFF state, preventing the first astable from oscillating. This first astable is colour-coded blue in the diagram. Its frequency (the pitch of the note it produces) is governed by the values of the 47K resistor and the 1 microfarad capacitor. Reducing the value of either of these will raise the frequency (note pitch). If rain falls on the sensor, the voltage at point ‘A’ goes high letting the astable run freely. If the voltage at ‘A’ does not rise sufficiently when it rains, increase the value of the 1M resistor.

3. The output of the first astable is a low voltage when the sensor is dry. It is taken from point ‘B’ and passed to the gating input of the second astable, holding it in its OFF state. The speed of the second astable is controlled by the value of the 470K resistor and the 0.001 microfarad capacitor. Reducing the value of either of these will raise the pitch of the note produced by the astable. The rate at which this astable operates is very much higher than the first astable.

When it rains, the voltage at point ‘A’ rises, letting the first astable oscillate. As it does so, it turns the second astable on and off in a steady rhythmic pattern. This feeds repeated bursts of high speed oscillations from the second astable to point ‘C’ in the diagram.

4. The Darlington-pair emitter-follower transistors cause the voltage at point ‘D’ to follow the voltage pattern at point ‘C’ (but 1.4 Volts lower voltage due to the 0.7 Volts base/emitter voltage drop for each transistor). The high gain of the two transistors ensures that the output of the second oscillator is not loaded unduly. These power-driver transistors place the output voltage across an eighty ohm loudspeaker, padded with a resistor to raise the overall resistance of the combination. The voltage pattern produced is shown at point ‘D’ and is an attention-grabbing sound.

So, why does this circuit oscillate?:

The circuit will not oscillate if the gating input is low, so assume it to be high. Take the moment when the output of gate 2 is low. For this to happen, the inputs of gate 2 have to be high. As the output of gate 1 is wired directly to the inputs of gate 2, it must be high, and for that to be true, at least one of its inputs must be low. This situation is
There is now a full voltage drop between point ‘A’ and point ‘B’. The 47K resistor and the capacitor are in series across this voltage drop, so the capacitor starts to charge up, progressively raising the voltage at point ‘C’. The lower the value of the resistor, the faster the voltage rises. The larger the value of the capacitor, the slower the voltage rises.

When the voltage at point ‘C’ rises sufficiently, the 100K resistor raises the input voltage of gate 1 far enough to cause it to change state. This creates the following situation:

Now, the voltage across ‘A’ to ‘B’ is reversed and the voltage at point ‘C’ starts to fall, its rate governed by the size of the 47K resistor and the 1 microfarad capacitor. When the voltage at point ‘C’ falls low enough, it takes the input of gate 1 low enough (via the 100K resistor) to cause gate 1 to switch state again. This takes the circuit to the initial state discussed. This is why the circuit oscillates continuously until the gating input of gate 1 is taken low to block the oscillation.

Now, here is a NAND gate circuit for a sequential on/off switch:

This circuit turns the Light Emitting Diode on and off repeatedly with each operation of the press-button switch. When the on/off switch is closed, capacitor ‘C1’ holds the voltage at point ‘A’ low. This drives the output of gate 1 high, which moves the inputs of gate 2 high via the 100K resistor ‘R1’. This drives the voltage at point ‘B’ low, turning the transistor off, which makes the LED stay in its off state. The low voltage at point ‘B’ is fed back via the 100K resistor ‘R2’ to point ‘A’, keeping it low. This is the first stable state.

As the output of gate 1 is high, capacitor ‘C2’ charges up to that voltage via the 2M2 resistor. If the press-button switch is operated briefly, the high voltage of ‘C2’ raises the voltage of point ‘A’, causing gate 1 to change state, and consequently, gate 2 to change state also. Again, the high voltage at point ‘B’ is fed back to point ‘A’ via the 100K resistor ‘R2’, keeping it high, maintaining the situation. This is the second stable state. In this state, point ‘B’ has a high voltage and this feeds the base of the transistor via the 4.7K resistor, turning it on and lighting the LED.

In this second state, the output of gate 1 is low, so capacitor ‘C2’ discharges rapidly to a low voltage. If the press-button switch is operated again, the low voltage of ‘C2’ drives point ‘A’ low again, causing the circuit to revert to its original stable state.

We could, if we wished, modify the circuit so that it would operate for three or four minutes after switch-on but then stop operating until the circuit was turned off and on again. This is accomplished by gating one of the gates instead of just using both as inverters. If we gated the second gate, then the LED would be left permanently on,
so we will modify the first gate circuit:

This circuit operates exactly the same way as the previous circuit if, and only if, the voltage at point ‘C’ is high. With the voltage at point ‘C’ high, gate 1 is free to react to the voltage at point ‘A’ as before. If the voltage at point ‘C’ is low, it locks the output of gate 1 at the high level, forcing the output of gate 2 to the low level and holding the LED off.

When the circuit is first powered up, the new 100 microfarad capacitor ‘C3’ is fully discharged, which pulls the voltage at point ‘C’ to nearly + 9 Volts. This allows gate 1 to operate freely, and the LED can be toggled on and off as before. As time passes, the charge on capacitor ‘C3’ builds up, fed by the 2M2 resistor. This causes the voltage at point ‘C’ to fall steadily. The rate of fall is governed by the size of the capacitor and the size of the resistor. The larger the resistor, the slower the fall. The larger the capacitor, the slower the fall. The values shown are about as large as are practical, due to the current ‘leakage’ of ‘C3’.

After three or four minutes, the voltage at point ‘C’ gets driven low enough to operate gate 1 and prevent further operation of the circuit. This type of circuit could be part of a competitive game where the contestants have a limited time to complete some task.

The NAND Gate as an Amplifier.
Gates can also be used as amplifiers although they are not intended to be used that way and there are far better integrated circuits from which to build amplifiers. The following circuit shows how this can be done:

This circuit operates when there is a sudden change in light level. The previous light-level switching circuit was designed to trigger at some particular level of increasing or decreasing level of lighting. This is a shadow-detecting circuit which could be used to detect somebody walking past a light in a corridor or some similar situation.

The voltage level at point ‘A’ takes up some value depending on the light level. We are not particularly interested
in this voltage level since it is blocked from the following circuitry by capacitor ‘C1’. Point ‘B’ does not get a voltage pulse unless there is a sudden change of voltage at point ‘A’, i.e. there is a sudden change in light level reaching the light-dependent resistor ORP12.

The first gate amplifies this pulse by some fifty times. The gate is effectively abused, and forced to operate as an amplifier by the 10M resistor connecting its output to its input. At switch-on, the output of gate 1 tries to go low. As its voltage drops, it starts to take its own inputs down via the resistor. Pushing the voltage on the inputs down, starts to raise the output voltage, which starts to raise the input voltage, which starts to lower the output voltage, which ...... The result is that both the inputs and the output take up some intermediate voltage (which the chip designers did not intend). This intermediate voltage level is easily upset by an external pulse such as that produced by the ORP12 through capacitor ‘C1’. When this pulse arrives, an amplified version of the pulse causes a voltage fluctuation at the output of gate 1.

This voltage change is passed through the diode and variable resistor to the input of gate 2. Gates 2 and 3 are wired together as a makeshift Schmitt trigger in that the output voltage at point ‘D’ is fed back to point ‘C’ via a high value resistor. This helps to make their change of state more rapid and decisive. These two gates are used to pass a full change of state to the output stage transistor. The variable resistor is adjusted so that gate 2 is just about to change state and is easily triggered by the pulse from amplifier gate 1. The output is shown as an LED but it can be anything you choose. It could be a relay used to switch on some electrical device, a solenoid used to open a door, a counter to keep track of the number of people using a passageway, etc. etc. Please note that an operational amplifier chip (which will be described later) is a far better choice of IC for a circuit of this type. A gate amplifier is shown here only to show another way that a gate can be utilised.

The NE555 Timer Chip.
There is an exceptionally useful chip designated by the number 555. This chip is designed to be used in oscillator and timer circuits. Its use is so widespread that the chip price is very low for its capability. It can operate with voltages from 5 Volts to 18 Volts and its output can handle 200 mA. It takes 1 mA when its output is low and 10 mA when its output is high. It comes in an 8-pin Dual-In-Line package and there is a 14-pin package version which contains two separate 555 circuits. The pin connections are:

```
0 Volts Trigger Output Reset
1 8
2 7
3 6
4 5
```

This device can operate as a monostable or astable multivibrator, a Schmitt trigger or an inverting buffer (low current input, high current output).

Here it is wired as a Schmitt trigger, and for variation, it is shown triggering a triac which will then stay on until the circuit is powered down (an SCR could be used just as well with this DC circuit):
And here, a monostable:

![Monostable multivibrator diagram](image)

And here are two astables, the second of which has fixed, equal mark/space ratio and the first a high output voltage time determined by \( Ra + Rb \) and a low voltage output time determined by \( Rb \) (2:1 in this case):

![Astable multivibrators diagram](image)

Frequency: \[
\text{Frequency} = \frac{0.72}{C \times R}
\]
Astable Frequencies

<table>
<thead>
<tr>
<th>Capacitor Value (μF)</th>
<th>100 Hz</th>
<th>470 Hz</th>
<th>1K Hz</th>
<th>4.7K Hz</th>
<th>10K Hz</th>
<th>47K Hz</th>
<th>100K Hz</th>
<th>470K Hz</th>
<th>1M Hz</th>
</tr>
</thead>
<tbody>
<tr>
<td>0.1 μF</td>
<td>72,000 Hz</td>
<td>15,319 Hz</td>
<td>7,200 Hz</td>
<td>1,532 Hz</td>
<td>720 Hz</td>
<td>153 Hz</td>
<td>72 Hz</td>
<td>15 Hz</td>
<td>7.2 Hz</td>
</tr>
<tr>
<td>0.47 μF</td>
<td>15,319 Hz</td>
<td>3,259 Hz</td>
<td>1,532 Hz</td>
<td>326 Hz</td>
<td>153 Hz</td>
<td>33 Hz</td>
<td>15 Hz</td>
<td>3.3 Hz</td>
<td>1.5 Hz</td>
</tr>
<tr>
<td>1.0 μF</td>
<td>7,200 Hz</td>
<td>1,532 Hz</td>
<td>720 Hz</td>
<td>153 Hz</td>
<td>72 Hz</td>
<td>15 Hz</td>
<td>7.2 Hz</td>
<td>1.5 Hz</td>
<td>1.4 Hz</td>
</tr>
<tr>
<td>2.2 μF</td>
<td>3,272 Hz</td>
<td>696 Hz</td>
<td>327 Hz</td>
<td>70 Hz</td>
<td>33 Hz</td>
<td>7 Hz</td>
<td>3.3 Hz</td>
<td>1.4 Hz</td>
<td>3 Hz</td>
</tr>
<tr>
<td>4.7 μF</td>
<td>1,532 Hz</td>
<td>326 Hz</td>
<td>153 Hz</td>
<td>33 Hz</td>
<td>15 Hz</td>
<td>3.3 Hz</td>
<td>1.5 Hz</td>
<td>3 Hz</td>
<td>6.7 Hz</td>
</tr>
<tr>
<td>10 μF</td>
<td>720 Hz</td>
<td>153 Hz</td>
<td>72 Hz</td>
<td>15 Hz</td>
<td>7.2 Hz</td>
<td>1.5 Hz</td>
<td>1.4 Hz</td>
<td>6.7 Hz</td>
<td>14 Hz</td>
</tr>
<tr>
<td>22 μF</td>
<td>327 Hz</td>
<td>70 Hz</td>
<td>33 Hz</td>
<td>7 Hz</td>
<td>3.3 Hz</td>
<td>1.4 Hz</td>
<td>3 Hz</td>
<td>14 Hz</td>
<td>30 Hz</td>
</tr>
<tr>
<td>47 μF</td>
<td>153 Hz</td>
<td>33 Hz</td>
<td>15 Hz</td>
<td>3.3 Hz</td>
<td>1.5 Hz</td>
<td>3 Hz</td>
<td>6.7 Hz</td>
<td>30 Hz</td>
<td>65 Hz</td>
</tr>
<tr>
<td>100 μF</td>
<td>72 Hz</td>
<td>15 Hz</td>
<td>7.2 Hz</td>
<td>1.5 Hz</td>
<td>1.4 Hz</td>
<td>6.7 Hz</td>
<td>14 Hz</td>
<td>65 Hz</td>
<td>139 Hz</td>
</tr>
<tr>
<td>220 μF</td>
<td>33 Hz</td>
<td>7 Hz</td>
<td>3.3 Hz</td>
<td>1.4 Hz</td>
<td>3 Hz</td>
<td>14 Hz</td>
<td>30 Hz</td>
<td>139 Hz</td>
<td>307 Hz</td>
</tr>
<tr>
<td>470 μF</td>
<td>15 Hz</td>
<td>3.3 Hz</td>
<td>1.5 Hz</td>
<td>3 Hz</td>
<td>6.7 Hz</td>
<td>30 Hz</td>
<td>65 Hz</td>
<td>307 Hz</td>
<td>614 Hz</td>
</tr>
<tr>
<td>1,000 μF</td>
<td>7.2 Hz</td>
<td>1.5 Hz</td>
<td>1.4 Hz</td>
<td>6.7 Hz</td>
<td>14 Hz</td>
<td>65 Hz</td>
<td>139 Hz</td>
<td>614 Hz</td>
<td>12 - 52</td>
</tr>
<tr>
<td>2,200 μF</td>
<td>3.3 Hz</td>
<td>1.4 Hz</td>
<td>3 Hz</td>
<td>14 Hz</td>
<td>30 Hz</td>
<td>65 Hz</td>
<td>307 Hz</td>
<td>614 Hz</td>
<td></td>
</tr>
<tr>
<td>4,700 μF</td>
<td>1.5 Hz</td>
<td>3.3 Hz</td>
<td>6.7 Hz</td>
<td>30 Hz</td>
<td>65 Hz</td>
<td>307 Hz</td>
<td>614 Hz</td>
<td></td>
<td></td>
</tr>
<tr>
<td>10,000 μF</td>
<td>1.4 Hz</td>
<td>6.7 Hz</td>
<td>14 Hz</td>
<td>65 Hz</td>
<td>139 Hz</td>
<td>614 Hz</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Note: The high leakage of large value electrolytic capacitors prevents them being used with high value resistors in timing circuits. Instead, use a smaller capacitor and follow the timing circuit with a “divide-by-N” chip to give accurately timed long periods. Not all 555 chips have a manufacturing quality sufficient for them to operate reliably above 20,000 Hz, so for the higher frequencies the chip needs to be selected after testing its actual performance.

We can also wire the 555 to give a variable mark/space ratio while holding the frequency of the oscillation fixed:

The output waveform changes drastically as the variable resistor is adjusted, but the frequency (or pitch of the note) of the output stays unaltered.

A variable-frequency version of this circuit can be produced by changing the 33K resistor to a variable resistor as shown here:
Here, the 33K resistor has been replaced by two variable resistors and one fixed resistor. The main variable resistor is 47K in size (an almost arbitrary choice) and it feeds to a second variable resistor of 4.7K in size. The advantage of this second variable resistor is that it can be set to its mid point and the frequency tuning done with the 47K variable. When the frequency is approximately correct, the 4.7K variable can be used to fine tune the frequency. This is convenient as the small variable will have ten times more knob movement compared to the main variable (being just 10% of its value).

Obviously, it is not necessary to have the fine-tuning variable resistor, and it can be omitted without changing the operation of the circuit. As the 47K variable resistor can be set to zero resistance and the 4.7K variable resistor can also be set to zero resistance, to avoid a complete short-circuit between output pin 3 and the 50K Mark/Space variable resistor, a 3.3K fixed resistor is included. In this circuit, the frequency is set by your choice of the resistor chain 47K + 4.7K + 3.3K (adjustable from 55K to 3.3K) and the 100nF (0.1 microfarad) capacitor between pin 6 and the zero volt rail. Making the capacitor larger, lowers the frequency range. Making the resistors larger, also lowers the frequency range. Naturally, reducing the size of the capacitor and/or reducing the size of the resistor chain, raises the frequency.

One 555 chip can be used to gate a second 555 chip via its pin 4 'Reset' option. You will recall that we have already developed a circuit to do this using two astables and a transistor. We also generated the same effect using four NAND gates. Here, we will create the same output waveform using the more conventional circuitry of two 555 chips:

Both of the NE555 circuits can be bought in a single 14-pin DIL package which is designated 'NE556':

There are many additional circuit types which can be created with the 555 chip. If you wish to explore the possibilities, I suggest that you download the free pdf “50 555 Projects” from the web site:
The 555 chip can also produce a sine wave output:

![Sine Wave Diagram](http://www.talkingelectronics.com/projects/50%20-%20555%20Circuits/50%20-%20555%20Circuits.html)

Or, if you wish, a bi-stable multivibrator:

![Bi-stable Multivibrator Diagram](http://www.talkingelectronics.com/projects/50%20-%20555%20Circuits/50%20-%20555%20Circuits.html)

All right, suppose that we want to design and build a circuit to do the same as Bob Beck’s pulser circuit mentioned in chapter 11. The requirements are to produce a square wave output pulsing four times per second using a 27 volt power supply, the circuit being powered by three small PP3 size batteries. An obvious choice for the circuit seems to be a 555 timer chip which is small, robust and cheap and a suitable circuit would appear to be:

![Pulser Circuit Diagram](http://www.talkingelectronics.com/projects/50%20-%20555%20Circuits/50%20-%20555%20Circuits.html)

This leaves us with choosing a value for the capacitor and the resistor. We need to pay attention to the fact that the circuit will be running on 27 volts and while the capacitor will not charge up to anything like that voltage, we still will pick one which will survive 27V. Looking on the local eBay shows that a pack of ten capacitors of 1 microfarad rated at 50V can be bought for just £1 including postage, so take that as the value for “C”. Looking at the 555 table of frequencies above shows:
Which indicates that to get the circuit switching four times per second (4 Hz) the resistor “R” will need to be somewhere between 100K and 470K. With my capacitor, 120K is about right.

While the switching frequency does not have to be exact, let’s aim at getting it correct. Most reasonably priced components have a tolerance of around 10% so we need to select our resistor/capacitor combination for the exact values of the actual components which we will use. For this, it is worth building the circuit on a solder-less ‘breadboard’, so looking on eBay again we find that a suitable small plug-in board can be bought and delivered for £3. It looks like this:

These type of boards allow ICs to be plugged in spanning the central divide, leaving up to five extra connections on every pin. Short lengths of solid-core wire can be used to connect between any two socket holes. This will allow us to plug in one of our capacitors and find what resistor (or what two resistors) make the circuit switch forty times in ten seconds.

However, if we go to [http://www.alldatasheet.co.kr/](http://www.alldatasheet.co.kr/) and download the data pdf for the NE555 chip, we find that the maximum 555 chip voltage is quite limited:

This means that the chip is liable to burn out instantly if it is fed more than 16 volts. As we need to run our circuit on 27V this is a problem. As the 27V is being provided by three separate batteries, we could supply the 555 chip from just one of the batteries and run it on 9V which would be ok from the point of view of the chip as the table above shows that it can operate correctly with a supply voltage as low as 4.5 volts. The disadvantage of that arrangement is that one of the batteries will run down more quickly than the others and it would be nice to avoid that.

The table also shows that the current draw just to keep the 555 running can be anything from 6 to 15 milliamps. That is not a large current but the PP3 batteries have been chosen for their small size, allowing the whole circuit to be strapped to a person’s wrist. A quick search on the internet shows that cheap PP3 batteries have a capacity of 400 milliamp-hours and the very expensive alkaline types 565 milliamp-hours. These ratings are the “C20” values, based on the battery being discharged at a constant current over a period of twenty hours, which would be ten days of use if Bob Beck’s two hours per day protocol is followed.

This means that the ‘cheap’ batteries should not be discharged at more than one twentieth of their 400 mAh rating, which is 20 mA. The expensive alkaline batteries should be able to be discharged at 28 mA for twenty hours.

Our current draw is made up of two parts. The first part is supplying the circuit with the current which it needs to run. The second part is the current flowing through the body of the user. This second part is limited by the 820
ohm resistor in the output line which limits that part of the current to a maximum of 33 milliamps (Ohm’s Law: 
Amps = Volts /Resistance). This neglects the body resistance and assumes that the output control variable 
resistor is set to minimum resistance, which is unlikely.

Checking these values shows that the 555 chip is liable to draw as much current as the circuit supplies through 
the output electrodes. However, let’s go ahead with the circuit, after all, we might decide to use rechargeable PP3 
batteries which would overcome the need to buy new batteries every few days.

The first essential requirement is to provide the 555 chip with a voltage of, say, 10 volts when it is running in the 
completed circuit. That could be done with one of the voltage-stabiliser integrated circuits:

That is not a particularly expensive option, but those chips draw a current in order to provide the voltage 
stabilisation and an absolutely steady voltage is not needed by the 555 chip. Alternatively, we could use a 
resistor and a 10V zener diode:

But that method does waste some current flowing through the zener in order to provide the wanted voltage. The 
most simple method is to use a resistor and a capacitor:

Considerable care is needed when selecting the resistor value “R”. If the value is too low, then the voltage 
passed to the 555 chip will be too high and the chip will burn out. When selecting the resistor “R”, start with a 
higher value than expected and then substitute slightly lower value resistors while monitoring the voltage across 
the capacitor to make sure that it stays low enough. The resistor value can be assessed using Ohm’s Law. 
Assuming a current of about 6 mA, the voltage drop across the resistor being \((27 - 10) = 17\) volts, then a resistor 
of about 2.83K (as Ohms = Volts / Amps) which suggests that starting with a 4.7K resistor is likely to be ok, and 
then picking each lower standard resistor in turn until a satisfactory voltage across the capacitor is reached.

The capacitor could be 12V or 15V rated, but if one rated at a higher voltage is used, then if it is accidentally 
connected across the full 27V it will not be harmed in any way. The larger the capacitance, the better, say 220 
microfarads which can be got for a few pence on eBay. If you want to play safe, you could connect a 12V zener 
diode across the capacitor. It will not draw any current under normal working conditions, but if anything should 
cause the voltage on the capacitor to rise, then it will fire up and hold the voltage down to a safe 12V level. I 
would be inclined to see the zener as being unnecessary, but the choice is always yours.

So what resistor power rating is needed? Well, if the resistor turns out to be a 2.7K and the capacitor voltage 
ends up as 9.5 volts, then the average voltage across the resistor is 17.5V which makes the current through it 
6.48 mA and as Watts = Volts x Amps, the power rating needs to be 113 milliwatts, so the typical quarter-watt 
(250 mW) resistor should be perfectly ok. If two (nearly equal value) resistors in parallel are used to get some 
intermediate value of “R” then that increases the overall resistor wattage.
The output of the 555 chip is then used to drive the remainder of the circuit which operates at 27V. A BC109C transistor costs only a few pence, can handle the voltage and has a minimum gain of 200 although the gain can be anything up to 800 and a BC109 can handle the current quite easily. If you need to find out any of these things, then download a datasheet for the transistor from the internet.

The output of the 555 timer is on pin 3 and it can easily supply 200 mA which is far, far more current than we would ever need for this circuit. We can feed the 555 square-wave output to the 27V electrodes using a transistor:

As the transistor is made of silicon, the switch-on voltage is when the base voltage is about 0.7 volts above the emitter voltage. That means that when the transistor is switched on, the top of resistor “R1” will be at around 10 volts and the bottom of “R1” will be at about 0.7 volts, which means that the voltage across “R1” will be \((10 - 0.7) = 9.3\) Volts. When that voltage is present across “R1” we want it to feed sufficient current to the transistor to switch it on fully. The transistor supplies a 100K resistor (which will carry 0.27 mA when 27 volts is across it) and the electrodes which will have a minimum resistance of 820 ohms across them (causing a current of 33 mA through them). So, the transistor might have to supply about 33 mA maximum. The BC109C transistor has a minimum gain of 200 so the current flowing into the base needs to be \(\frac{33}{200} = 0.165\) mA and the resistor which will carry that current when it has 9.3 volts across it is 56.3K. A somewhat smaller resistor will suit.

A commonsense check that the resistor calculation is correct is:
A 1K resistor carries 1 mA per volt and so will carry 9.3 mA with 9.3 volts across it.
A 10K resistor will carry one tenth of that amount, or 0.93 mA with 9.3 volts across it.
A 100K resistor will carry one tenth of that again, or 0.093 mA with 9.3 volts across it.
This indicates that for a current of 0.165 mA which is about twice the 100K current, a resistor of about half of 100K should be about the right value, so 56.3K looks correct.

Considering that the gain of 200 is the minimum and three or four times that is typical, we could perhaps choose to use a 47K resistor for “R1”.

As the electrode current is likely to be considerably less than 33 mA and as the BC109C gain is likely to be very high, it could be quite difficult to get the transistor to switch off as it can operate on very tiny amounts of input current. To get it to switch on and off cleanly when the 555 output voltage is say, about 5 volts, (at which point the NE555 voltage will be changing very rapidly), “R2” is included. With it in place, the output voltage of the NE555 is divided between “R1” and “R2” in the ratio of their resistances. The situation we want is:

When The transistor is not switched on, it draws almost no current and so looks like a very high value resistor to the circuit. This allows the “R1” and “R2” resistors to act as a voltage-divider pair. This causes the voltage at point “A” to be determined by the ratio of “R1” to “R2” and the transistor can be ignored provided that the voltage at point “A” is below 0.7 volts. If the voltage at that point rises to 0.7 volts then the situation changes dramatically and Ohm’s Law no longer holds as the transistor is not a passive resistor but instead, is an active semi-conductor device. If the voltage at point “A” tries to rise further it can’t because the transistor base clamps it solidly there by appearing to be an ever lower resistor between the base and the emitter of the transistor. So for higher input voltages, resistor “R2” might as well not be there for all the difference it makes.
So, what value do we need for “R2” in order for the voltage at point “A” to be 0.7V when pin 3 of the NE555 reaches 5V? Well, that part of the circuit is acting in a resistive fashion and so Ohm’s Law can be used. The resistor “R1” is 47K and has 4.3 volts across it, which means that the current through it must be 0.915 mA. That means that “R2” has 0.7V across it and 0.915 mA flowing through it which means that it has a value of 7.65K. A standard 8.2K or 6.8K resistor could be used as there is nothing dramatically important about the 5V switching point. If you were fussed about getting exactly 7.65K (and you shouldn’t be), then you can get that value by combining two standard resistors, either in series or in parallel.

A common sense method of working out the value of “R2” is to use the fact that as the same current flows through them (no matter what that current happens to be), then the ratio of the voltage will be the same as the ratio of the resistors. That is: 0.7V / 4.3V = “R2” / 47K or “R2” = 47K x 0.7 / 4.3 which is 7.65K.

We have now reached the point where we can determine the resistor value needed to provide a reasonable voltage for the NE555 timer chip, the circuit being:

![Circuit Diagram]

The “Rx” value is going to be fairly close to 270K so you can use that value when testing to find a suitable value for “R” (2.2K in my case). The capacitor across the NE555 chip should be as large a capacitance as is convenient, bearing in mind that the entire circuit, batteries, etc. is to fit into a small case to be strapped to a wrist. One way that the components could be positioned on the plug-board is:

![Plug Board Diagram]

Remember that when trying various resistors for “R” you need to start high at about 4.7K and the resulting voltage on the capacitor shows the voltage drop across your first resistor choice and so, the actual current being drawn by your particular NE555 chip. That calculated current will allow you to calculate the resistor value needed to give 10 volts or so, allowing your next resistor to be tested to be almost exact in value.

For checking the frequency produced by the circuit, any ordinary LED can be used as a temporary measure. It can be connected across the 100K ‘load’ resistor between the transistor collector and the +27V positive supply.
A current-limiting resistor is essential to stop the LED burning out instantly. If we allow a current of 5 mA to flow through the LED then since the current-limiting resistor has some 26.3 volts across it, then it’s value will be about 5.4K (1K would give 26 mA, 2K would give 13 mA, 3K would give 9 mA, 4K would give 6.5 mA) and so a 4.7K resistor works well. This LED and resistor are shown in the layout above. Please remember that if your BC109C transistor has a metal case, then that case is normally connected internally to the collector and so, care must be taken that the case does not short-circuit to anything else.

If it is considered important to maximise battery life by reducing the current draw to a minimum, then perhaps using an astable circuit might be a good choice. In common with most electronic circuits, there are many different ways to design a suitable circuit to do the required job. The BC109C transistor can handle the 27V and so we might aim at a current draw for the circuit of just 3 mA. If 2 mA were to flow through the astable transistors when they are switched on, then with 27V across them, the resistors would be 13.5K which is not a standard value. We might select 12K to give a 2.25 mA current, or 15K to give 1.8 mA. Either should be satisfactory. The circuit might then be:

As the voltage swing feeding the output transistor has now risen from 10V to 27V the voltage-divider resistors can now increase in value by 2.7 times, giving around 127K and 22.1K for these resistors. However, the situation is not the same as for the NE555 chip which can supply at least 200 mA at the voltage-high output level. Instead, the transistor becomes such a high resistance that it can be ignored, but the 12K remains in the path which supplies the base current for the output transistor and it will in fact, add to the upper resistor of the voltage-divider pair. So while a 100K resistor is shown, it is effectively 112K due to that extra 12K resistor between it and the +27V supply line. The astable transistors will be switching fast at the point where the output transistor changes state, so the output square wave should be good quality. The BC109C transistor can switch on and off a hundred million times per second, so it’s performance in this circuit should be very good. A test breadboard layout might be:

We now need to choose the timing components. For an even 50% duty cycle where each transistor is ON for half the time and OFF for half the time, the two timing capacitors can be the same size and then the two timing resistors will have the same value, in my case, 330K but it depends on the actual capacitors used.
Bob Beck’s design calls for the LED display to be running when the unit is switched on and then be disconnected when the electrodes are plugged into a 3.5 mm socket mounted on the case containing the circuit. The switched socket looks like this:

When the plug is not inserted into the socket, pin 1 connects to pin 2 and pin 3 is not connected to anything. When the plug is inserted, then pin 1 is isolated, pin 2 is connected to plug pin 4 and pin 3 is connected to plug pin 5.

The Beck circuit is connected to the output socket like this:

This arrangement will give a 27V 4Hz square wave output through the jack socket. But, Bob Beck’s original circuit did not do that. Instead, it was like this:

Here, a relay operates two change-over switch contacts which are used to reverse the battery bank contacts four times per second. That is different from just producing a positive-going square wave voltage between the two output terminals. If you were to consider a resistor connected across the output socket, then with the relay switching, the direction of the current reverses four times per second, but with the square wave, while it starts and stops four times per second, the direction of the current is always the same and there is no reversal of direction.

As Bob wanted to avoid using a relay which clicks four times per second all the way through the two-hour treatment described in chapter 11 and in the “Take Back Your Power” pdf on the http://www.free-energy-info.tuks.nl/ web site, he redesigned the circuit using the very impressive LM358/A integrated circuit:
This chip draws only half of one milliamp, has two very high-gain operational amplifiers and can operate with a wide range of supply voltages. It is also inexpensive.

Bob displays the circuit as:

![Circuit Diagram]

Bob states that the first section acts as a 4Hz square-wave signal generator, the frequency being controlled by the 2.4M resistor “R1” and the 100nF capacitor “C1”. The data sheet for the LM358 states that the output voltage swing is between zero volts and 1.5V less than the supply voltage “Vcc” (which is +27V in this case). That implies that, as would be expected, the pin 1 output voltage from the first stage will switch sharply from 0V to +25.5V and sharply back again, four times per second.

It is difficult to follow the circuit as it is drawn, so it might be a little easier to follow when drawn like this:

![Circuit Diagram]

The output from the first amplifier inside the LM358 package is on pin 1 and it can supply a large amount of current (if a large current is ever needed). That output goes straight to one of the jack socket connections. It also goes the pin 6 input of the second amplifier inside the chip and that causes the high-power output of that amplifier on pin 7 to be the opposite of the pin 1 voltage. When pin 1 goes high to +25.5 volts, then pin 7 goes low, to about zero volts. That output is also fed to the other jack socket connection, placing 25.5 volts across the electrodes when they are plugged in to the jack socket.

When the oscillator circuitry connected to the first amplifier causes the voltage on pin 1 to go low, then the output on pin 7 inverts it and so it goes to +25.5 volts. You will notice that while the overall voltage of 25.5 volts is applied again to the jack socket, the polarity is now reversed, achieving what the relay circuit does (although 1.5 volts is lost in the process). This is a neat solution.

Bob uses a two-colour LED to confirm that the circuit is working correctly before the electrodes are plugged in. He chooses to do it this way:
The two 18V zener diodes drop off 18.7 of the 25.5 volts as one will be forward biased dropping 0.7 volts and the other reversed biased, dropping off 18 volts. That leaves a 7V drop for the LED, which is a bit excessive, so Bob says that he uses a capacitor to limit the current. As there is already an 820 ohm resistor in the LED current path through the socket, the capacitor is not needed. The variable resistor need to be set to it's minimum resistance by rotating it's shaft fully clockwise so that it does not affect the LED brightness as the zeners also show when the battery voltage has dropped as there will no longer be sufficient voltage to light the LED brightly, indicating that the batteries need to be replaced (or recharged if they are rechargeable batteries). When testing the circuit, an alternative to the two zeners is to use a 4.7K resistor and if a bi-colour LED is not to hand, then two ordinary LEDs can be used back to back like this:

With this arrangement, the two LEDs flash alternately. In any circuit, a capacitor with a higher voltage rating can always be used if the capacitance values are the same. The Beck external circuit is completed through the body of the user, so there is just one electrode connected to each side of the output jack socket. A possible plug-board layout is:
The 4.7K resistor and LEDs are only on the board for testing purposes and when the circuit is built in permanent form, then the LED chain connects to pin 1 of the jack socket so that the LEDs are disconnected during the two hours of daily treatment recommended when using the device.

One stripboard layout using the standard 9-strip 25-hole board and incorporating the two 18V zener diodes for voltage sensing is:

When using a Beck device, it is very important to pay attention to the precautions which Bob sets out. These are in his “Take Back Your Power” pdf document: [http://www.free-energy-info.tuks.nl/Beck.pdf](http://www.free-energy-info.tuks.nl/Beck.pdf) which includes the following, which, while it refers to treatment to deal with HIV, presumably applies to all treatments with his device:

**EXPANDED INSTRUCTIONS FOR EXPERIMENTAL / THEORETICAL HIV BLOOD NEUTRALIZATION**

**HYPOTHETICAL PROTOCOLS FOR EXPERIMENTAL SESSIONS**


**PRECAUTIONS:** Do NOT use wrist to wrist current flow with subjects who have cardiac pacemakers. Any applied electrical signals may interfere with 'demand' type heart pacers and cause malfunction. Single wrist locations should be acceptable. Do NOT use on pregnant women, while driving or using hazardous machinery.

Users MUST avoid ingesting anything containing medicinal herbs, foreign or domestic, or potentially toxic medication. nicotine, alcohol, recreational drugs, laxatives, tonics and certain vitamins etc., for one week before starting because blood electrification can cause electroporation which makes cell membranes porous to small quantities of normally harmless-chemicals in plasma. The effect is the same as extreme overdosing which might be lethal. See *Electroporation: a General Phenomenon for Manipulating Cells and Tissues*; J.C. Weaver, Journal of Cellular Biochemistry 51:426-435 (1993). Effects can mimic increasing dosages many fold. Both the magnetic pulsar and blood purifier cause electroporation.

Do NOT place electrode pads over skin lesions, abrasions, new scars, cuts, eruptions, or sunburn. Do NOT advance output amplitude to uncomfortable levels. All subjects will vary. Do NOT fall asleep while using. The magnetic pulsar should be safe to use anywhere on body or head.

Avoid ingesting alcohol 24 hours before using. Drink an 8 oz. glass of distilled water 15 minutes before and immediately following each session end drink at least four additional glasses daily for flushing during 'neutralization' and for one week thereafter. This is imperative. Ignoring this can cause systemic damage from unflushed toxic wastes. When absolutely essential drugs must be ingested, do so a few minutes after electrification then wait 24 hours before next session.

If subject feels sluggish, faint, dizzy, headache or giddy, nauseous, bloated or has flu-like symptoms or rashes after exposures, reduce pulsing per session and/or shorten applications of electrification. Drink more water-preferably ozonized -to speed waste oxidation and disposal. Use extreme caution when treating patients with impaired kidney or liver function. Start slowly at first like about 20 minutes per day to reduce detoxification problems.
To avoid shock liability, use batteries only. Do NOT use any line-connected power supply, transformer, charger, battery eliminator, etc. with blood clearing device. However line supplies are OK with well-insulated magnetic pulse generators (strobe lights).

**Health professionals:** Avoid nicotine addicts, vegans, and other unconsciously motivated death-wishers and their covert agendas of ‘defeat the healer’. Tobacco, the most addictive (42 times more addictive than heroin) and deadly substance of abuse known, disrupts normal cardiovascular function. True vegetarian diets are missing essential amino acids absolutely necessary for the successful rebuilding of AIDS-ravaged tissues. Secondary gains (sympathy / martyrdom, work avoidance, free benefits, financial assistance, etc.) play large roles with many AIDS patients. “Recovery guilt” as friends are dying has even precipitated suicide attempts masked as 'accidents'. Avoid such entanglements, since many have unconscious death wishes.

**SUPERIOR ELECTRODES:** Excellent, convenient and vastly superior electrodes, reusable indefinitely can be made by butt-soldering lead wires to ends of 1” long by 3/32” dia. blanks cut from type 316 stainless steel rods available from welding supply stores (Cameron Welding Supply. 11061 Dale Ave., Stanton, CA 90680). Use 'Stay Clean' flux before soldering (zinc chloride/hydrochloric acid). Shrink-insulate TWO tight layers of tubing over soldered joints to prevent flexing/breaking and lead/copper ions from migrating. Wrap three or four turns of 100% cotton flannel around rods. Spiral-wrap with strong thread starting from wire side to end, tightly pinch cloth over the rod’s end so as to leave no metal exposed by wrapping 6 or 7 turns of thread TIGHTLY just off end of rod, then spiral wrap back to start and tie tightly with four knots then cut off excess cloth at end close to pinch -wraps. Treat end windings and knots with clear fingernail polish or Fray Check®(fabric & sewing supply stores) to prevent ravelling. Soak in a strong solution of sea salt (not table salt) containing a little wetting agent like Kodak Photo Flow, ethylene glycol, or 409 kitchen cleaner. Add a few drops of household bleach, silver colloid, etc., for disinfectant. Store solution for reuse. Tape soaking-wet electrodes tightly over pulse sites with paper masking or Transpore™ tape or with 1 inch wide stretch elastic bands with tabs of Velcro® at ends to fasten. Electrodes should closely conform precisely along blood vessels, not skewing ever so slightly over adjacent flesh. This insures better electrical conductivity paths to circulating blood and insures very low internal impedance. (~2000W). Rinse and blot-dry electrodes and skin after each use. NEVER allow bare metal to touch skin as this will cause burns manifested as small red craters that heal slowly. The objective is to get maximum current into blood vessels, not leak it over to adjacent tissue. Therefore never use any electrode wider that about 1/8 inch (3 mm).

**ELECTRODE PLACEMENTS:** Locate maximum pulse position (NOT to be confused with acupuncture, reflexology, Chapman, etc. points) on feet or wrists by feeling for maximum pulse on inside of ankle about 1 inch below and to rear of ankle bone, then test along top centre of instep. Place electrode on whichever pulse site on that foot that feels strongest. Scrub skin over chosen sites with mild soap and water or alcohol swab. Wipe dry. Position the electrodes lengthwise along each left and right wrists blood vessel. Note: with subjects having perfectly healthy hearts and not wearing pacers, it is convenient to use left wrist to right wrist exactly over ulnar arterial pulse paths instead of on feet. Recent (Dec. 1995) research suggests that placing both electrodes over different arteries on the same wrist works very well (see pg. 7), avoids any current through heart, and is much more convenient and just as effective. An 8” long, 1” wide elastic stretch-band with two 1.5” lengths of 3/4” wide Velcro® sewn to ends of opposite sides makes an excellent wrist band for holding electrodes snugly in place. With electrode cable unplugged, turn switch ON and advance amplitude control to maximum. Push momentary SW. 2 'Test' switch and see that the red and green light emitting diodes flash alternately. This verifies that polarity is reversing about 4 times per second (frequency is NOT critical) and that batteries are still good. When LED’s don’t light replace all three 9V batteries. Zener diodes will extinguish the LEDs when the three 9V battery’s initial 27V drops below 18V after extended use. Never use any electrode larger than 1.125” (28 mm) long by 1/8” wide to avoid wasting current through surrounding tissue. Confine exactly over blood vessels only. Apply drops of salt water to each electrode’s cotton cover ~every 20 minutes to combat evaporation and insure optimum current flow. Later devices are solid-state, use only three batteries and no relays, and are much smaller.

Now rotate amplitude control to minimum (counter-clockwise) and plug In electrode cable. Subject now advances dial slowly until he feels a “thumping” and tingling. Turn as high as tolerable but don’t advance amplitude to where it is ever uncomfortable. Adjust voltage periodically as he adapts or acclimates to current level after several minutes. If subject perspires, skin resistance may decrease because of moisture, so setting to a lower voltage for comfort is indicated. Otherwise it is normal to feel progressively less sensation with time. You may notice little or no sensation at full amplitude immediately, but feeling will begin building up to maximum after several minutes at which time amplitude must be decreased. Typical adapted electrode-to-electrode impedance is on the order of 2000W. Typical comfortable input (to skin) is about 3 mA, and maximum tolerable input (full amplitude) is about 7 mA but this 'reserve' margin although harmless is unnecessary and can be uncomfortable. Current flowing through blood is very much lower than this external input because of series resistance through skin, tissue and blood vessel walls, but 50 to 100 µA through blood is essential.

Apply blood neutralizer for about 2 hours daily for ~2 months. Use judgment here. The limiting factor is...
detoxification. Carefully monitor subject’s reactions (discomfort, catarrh, skin eruptions, weeping exudates, rashes, boils, carbuncles, coated tongue, etc.). With very heavy infections, go slower so as not to overload body’s toxic disposal capability. With circulation-impaired diabetics, etc., you may wish to extend session times. Again, have subject drink lots of water. Recent changes in theoretical protocol being currently tested suggest following up the three weeks of treatments with a 24 hours per day (around the clock) continuous electrification of blood for two days to deal a knockout blow to the remaining HIV’s 1.2 day life cycle. (A. Perelson; Los Alamos Biophysics Group, Mar. 16, 1996 “Science” Journal.) Remember to remoisten electrodes regularly. If you absolutely must ingest prescription drugs, do so immediately after turning off instrument and allow 24 hours before next treatment to let concentrations in blood plasma decay to lower levels.

Remember, if subjects ever feel sleepy, sluggish, listless, nauseous, faint, bloated, or headachy, or have flu-like reactions they may be neglecting sufficient water intake for flushing toxins. We interpret this as detoxification plus endorphin release due to electrification. Let them rest and stabilize for about 45 minutes before driving if indicated. If this detoxing becomes oppressive, treat every second day. Treating at least 21 times should ‘fractionate’ both juvenile and maturing HIV to overlap maximum neutralization sensitivity windows and interrupt ‘budding’ occurring during HIV cells' development cycles. Treatments are claimed to safely neutralize many other viruses, fungi, bacteria, parasites, and microbes in blood. See patents US 5,091,152  US 5,139,684 US 5,188,738 US 5,328,451 and others as well as numerous valid medical studies which are presently little known or suppressed. Also, ingesting a few oz. of about 5 parts per million of silver colloid solution daily can give subjects a 'second intact immune system' and minimise or eliminate opportunistic infections during recovery phase. This miracle substance is pre-1938 technology, and unlike ozone is considered immune from FDA harassment. Silver colloid can easily be made at home electrolytically in minutes and in any desired quantities and parts per million strength for under 14 cents per gallon plus cost of water. It is ridiculous to purchase it for high prices. Colloid has no side effects, and is known to rapidly eliminate or prevent hundreds of diseases. Silver colloids won’t produce drug resistant strains as will all other known antibiotics. No reasonable amount can overdose or injure users either topically, by ingestion, or medical professional injection.

The 741 Operational Amplifier.
An important and very useful group of Integrated Circuits is the “Operational Amplifier” or “op-amp” group. These devices have a very high gain, an ‘inverting’ input and a ‘non-inverting’ input. There are many op-amps but we will look at just one popular type called the “741” which has an ‘open-loop’ gain of 100,000 times. All operational amplifiers work in the same way in theory. The way they operate in a circuit is controlled by the external components attached to them. They can operate as inverting amplifier, a non-inverting amplifier (i.e. a ‘buffer’), a comparator, an astable multivibrator, and various other things. The symbol and connections for a 741 op-amp are:

We can connect the 741 chip to act as an amplifier with any set gain level that we choose:
Here, the gain is set by the ratio of the 220K resistor to the 22K resistor. This circuit has a gain of 10 times, so the input signal at point ‘B’ will generate an output signal at point ‘C’ which is ten times larger, provided that the output signal does not approach the battery voltage. If it does, then clipping will occur with the top and the bottom of the output waveform chopped off at about a volt away from the battery voltage levels, approximately 1 Volt and +11 Volts in this example.

Operational amplifiers are generally designed to operate from a dual power supply. In the above example, the supply would be created by using two 6 Volts batteries instead of one 12 Volt battery. To avoid the inconvenience of this, a mid-point voltage is generated at point ‘A’ by using two equal resistors in series across the battery. This gives a central voltage of +6 Volts which is fed to the IC.

This circuit can be used in many applications. Here is a circuit for a meter to measure sound intensity:

![Circuit diagram](image)

This circuit is two copies of the previous circuit. Each 741 chip has a reference voltage of half the supply voltage created by a voltage-divider pair of 1K resistors. This voltage is fed to pin 3 of the chip, which is the non-inverting input.

At point ‘A’, a microphone or small loudspeaker is used to generate a signal voltage when sound reaches it. This voltage is fed to the 741 op-amp via a 1 microfarad blocking capacitor. This passes the audio signal through while blocking the +4.5 Volts DC on pin 3. The first 741 has a gain of 22, set by the 10K and 220K resistors (220/10 = 22).

Point ‘B’ then receives an audio signal 22 times larger than the signal produced by the microphone. This signal is still quite small, so the second 741 boosts it further. The gain of the second 741 is variable and depends on the resistance set on the 1M variable resistor. If the variable resistor is set to zero ohms, then the gain of the second 741 will be controlled by the 4K7 resistor at point ‘C’ alone and so will be 1 (4.7/4.7 = 1). If the variable resistor is set to its maximum value, then the gain of the second 741 will be some 214 (1,004,700/4,700 = 213.8).

The two op-amps together have a combined gain which ranges from 22 to 4702. The amplified audio signal arrives at point ‘D’ and it can be adjusted to a respectable value. This alternating voltage is now rectified via the diodes at point ‘E’ and it builds up a DC voltage across the 47 microfarad capacitor there. This voltage is displayed on a voltmeter. The result is that the voltmeter shows a reading directly proportional to the sound level reaching the microphone.

The 741 can be wired as a buffer. This is the equivalent of an emitter-follower circuit when using transistors. The set up for the 741 is:
Difficult circuit - huh! Are you sure you can afford all the extra components? This circuit utilises the full gain of the 741 chip. The output follows the input waveform exactly. The input requires almost no current, so the circuit is described as having a 'high input impedance'. The output can drive a serious load such as a relay, so the circuit is described as having a 'low output impedance'.

The 741 chip can be wired to act as a comparator. This is the circuit:

Are you sure you are up to such a difficult circuit? Bit complicated - huh! This is the basic operational form for an operational amplifier.

If the voltage at point ‘A’ is higher than the voltage at point ‘B’ then the output goes as low as it can go, say 1 or 2 volts.

If the voltage at point ‘A’ is lower than the voltage at point ‘B’ then the output goes as high as it can go, say 10 volts or so.

Having seen how transistor circuits work, you should be able to understand why the 741 chip circuitry (which is a transistor circuit inside the 741 package) needs some voltage inside the supply rails to provide an efficient high-current output drive.

Here is a 741 version of the light-operated switch:
This circuit is set up as evening falls. We want the relay to have minimum voltage across it in daylight, so the voltage at point ‘A’ needs to be higher than the voltage at point ‘B’. As the 1K variable resistor is across the supply voltage, its slider can be set to any voltage between 0 Volts and +12 Volts. To make this easy to do, we choose a ‘linear’ variable resistor as the logarithmic variety would be hard to adjust in this application. With the ‘linear’ version, each 1 degree of rotation of the resistor shaft causes the same change in resistance, anywhere along the range. This is not the case for the logarithmic variety.

Anyway, we adjust the variable resistor downwards until the relay voltage drops to a minimum. When the light level has fallen to the level at which we wish the circuit to trigger, we adjust the variable resistor to make the relay click on. The 741 chip has a very rapid output voltage swing when the input voltages swap over, so the relay switching will be decisive. The switching can be made even more positive by adding a resistor between the output and point ‘B’. This acts like a Schmitt trigger when switching occurs by providing some additional positive feedback, lifting the voltage at point ‘B’.

If you wish the circuit to trigger on a rising light level, just swap the positions of the 10K resistor and the ORP12 light-dependent resistor. The same circuit will operate as a temperature sensing circuit by substituting a ‘thermistor’ (which is a temperature-dependent resistor) for the ORP12.

If we would like the circuit to act as a burglar alarm, we could use the same circuit like this:

![Circuit Diagram]

The circuit is still controlled by the voltage at point ‘A’. Under normal circumstances, this voltage will be near +6 Volts (produced by the two 10K resistors and the 100K resistor). The upper switch marked ‘NC’ for ‘Normally Closed’, represents a chain of, say, magnetic switches attached to doors and windows. If any of these are opened, then the voltage at point ‘A’ will be dictated by the lower 10K resistor in series with the 100K resistor. This will cause the voltage at ‘A’ to fall instantly to a low value, triggering the circuit.

The ‘NO’ switch (‘Normally Open’) represents one or more pressure-operated switches under carpets or rugs and/or switches which get brushed when doors are swung open, etc. These switches are wired in parallel across each other and if any of them is closed for even a millionth of a second, the voltage at point ‘A’ will be pulled down by the 1K resistor and the circuit will be triggered.

The circuit can be latched on in any one of a variety of ways. One relay contact can be used to hold the relay on or hold the voltage at ‘A’ low. A transistor can be wired across the relay to hold the circuit on, etc. etc. If this is done, the circuit will remain in its triggered state until the supply voltage is interrupted. You might prefer to use a 555 chip to limit the length of time the alarm sounds to three minutes or so.

The SCR and Triac.

An alternative to using a relay or semiconductor latch is to use a Silicon Controlled Rectifier usually referred to as an ‘SCR’ or ‘Thyristor’. This device is normally “off” with a very high resistance to current flow. If it is switched on by applying a voltage to its Gate connection, it stays continuously on until some external device stops current flowing through it. The following circuit shows how it operates:
When the voltage is first applied to the circuit by closing switch S2, the SCR is in its OFF state so no current is supplied to the load. If the press-button switch S1 is pressed, a current is fed into the Gate of the SCR, turning it ON. When switch S1 is allowed to open, the SCR remains in its ON state and it will stay that way until the current through it is cut off. Opening switch S2 cuts off the current to the load and the SCR returns to its OFF state. A very valid question would be: “Why have an SCR at all and just turn the load on and off with switch S2?” The answer is that switch S1 might be the under-carpet pressure pad of a burglar-alarm and it might be operated some hours after switch S2 was closed to activate the alarm system. Stepping off the pressure pad does not stop the alarm sounding.

While this sort of DC latching action is useful, it is more common for an SCR to be used in an AC circuit. For example, take the circuit shown here:

The 520 volt AC supply coming in from the right hand side, is converted to positive-going sine-wave pulses by the diode bridge. This pulsing voltage is applied to the Load/SCR path. If the voltage at pin 3 of the 555 chip is low, then the SCR will remain OFF and no current will be fed to the load device. If the voltage on pin 3 goes high and the voltage applied to the Load/SCR chain is high, then the SCR will be switched ON, powering the load until the pulsing voltage drops to its zero level again some 1/120 of a second later.

The 555 chip is connected to form a monostable multivibrator and the timing components (the 120K resistor and the 10nF capacitor) cause it to output a 1 millisecond pulse which is long enough to trigger the SCR into its ON state, but short enough to have finished before the mains pulse reaches its zero-voltage level again. The 555 chip is triggered by the rising mains voltage being passed to its pin 2 through the voltage-divider 100K and 120K pair of resistors, and that synchronises it with the AC waveform. Pin 4 of the 555 chip can be used to switch the load power on and off.

In the circuit shown above, the diode bridge is needed to convert the incoming AC waveform to pulsing DC as shown in red in the diagram, as the SCR can only handle current flowing in one direction. The AC load equipment works just as well with the pulsing DC as with a full blown AC waveform. A better semiconductor construction is the ‘Triac’ which acts like two SCR devices back-to-back in a single package. It is shown like this in circuit diagrams:
There are three connections to the device: Main Terminal 1, Main Terminal 2 and the Gate. When switch ‘S’ shown in the diagram is closed, the triac conducts on both positive and negative voltages applied to its MT1 and MT2 terminals. When the switch is open, the device does not conduct at all.

If the external circuit containing switch ‘S’ is placed inside the device as a permanently closed circuit, then the device becomes a 'Diac' which can be used to trigger a Triac and give a very neat circuit for controlling the power to an item of AC mains equipment as shown here:

Here, the variable resistor/capacitor pair controls the point on the AC waveform that the Triac is triggered and so controls how much of each sinewave cycle is passed to the mains equipment, and so it controls the average power passed to the equipment. A very common use for a circuit of this type is the ‘dimmer-switch’ used with household lighting.

To return now to the 741 chip. The 741 can also be used as an **astable multivibrator**. The circuit is:

The rate of oscillation of this circuit is governed by the Resistor marked ‘R’ in the diagram and the capacitor marked ‘C’. The larger the resistor, the lower the rate of oscillation, the larger the capacitor, the lower the rate of oscillation.

When the output goes high, capacitor ‘C’ charges up until the voltage on it exceeds the mid-rail voltage on pin 3, at which time the 741 output goes low. The capacitor now discharges through resistor ‘R’ until the voltage on it drops below the voltage on pin 3, at which time the output goes high again. The 10K resistor connecting the output to pin 3 provides some positive feedback which makes the 741 act quite like a Schmitt trigger, sharpening up the switching.

The same arrangement of resistor and capacitor applied to a Schmitt inverter or Schmitt NAND gate causes exactly the same oscillation:
If you would like to see additional ways of using 741 and 555 chips, I can recommend the excellent book “Elementary Electronics” by Mel Sladden and Alan Johnson ISBN 0 340 51373 X.

**A Hex Inverter Signal Generator.**
Here is a very well tested and highly thought of, low-cost oscillator circuit, using a 74HC14 Schmitt inverter chip (or the 40106B higher voltage CMOS chip). It allows fine tuning control of the frequency and the pulse width produced. Three of the inverters are connected together to give a more powerful output current drive:

**The 4022 Divide-by-Eight Chip.**
One very useful CMOS integrated circuit is the ‘4022’ chip which is a 16-pin ‘divide by 8’ chip with built-in decoding. The connections are:
If pin 14 is provided with the output from some variety of astable multivibrator, on the first pulse, this chip sets the “0” output on pin 2 to High while the other outputs are Low. On the next pulse, the “0” output goes Low and the “1” output on pin 1 goes High. On the next pulse, output “1” goes Low and the “2” output on pin 3, goes High. And so on until on the eighth pulse, output “7” on pin 10 goes Low and output “0” goes high again.

The chip can also divide by lower numbers:

For ‘Divide by 7’ operation, connect pin 10 to pin 15 (this resets the output to ‘0’)
For ‘Divide by 6’ operation, connect pin 5 to pin 15
For ‘Divide by 5’ operation, connect pin 4 to pin 15
For ‘Divide by 4’ operation, connect pin 11 to pin 15
For ‘Divide by 3’ operation, connect pin 7 to pin 15
For ‘Divide by 2’ operation, connect pin 3 to pin 15

If you want a ‘Divide by 1’ circuit, I suggest you cut down on the amount of alcohol you drink.

Here is an illustration of a ‘Divide by 4’ setup:

There are a number of things to notice in the above diagram. Firstly, the practical arrangements for circuitry have not been stressed before. If the circuitry has a pulsing circuit drawing heavy current, as shown by the thick red arrows, then it should be physically connected to the battery and any low-current circuitry should be further away from the battery. The supply from the battery should have a fuse or circuit breaker and a switch in the line before anything else is connected, so that if any component develops a fault and goes short-circuit, the fuse will blow and prevent any significant problems.

Secondly, it is a good idea to provide the other circuitry with a smoothed power supply as shown by the blue components in the diagram. This minimises the effect if the battery voltage gets pulled down by the pulsing of the
high-current circuitry. The diode (silicon, 1 Amp, 50 V) stops the heavy current circuit drawing current from the
large smoothing capacitor. The 100 ohm resistor limits the current into the large capacitor on switch-on and
provides a little more smoothing. This circuitry is called “de-coupling” as it de-couples the low current circuitry
from the high current circuitry.

Thirdly, notice capacitor “C1” which is wired physically as close to the power supply pins of the integrated circuit
as is possible. If a spike is superimposed on the battery supply, then this capacitor soaks it up and prevents it
damaging or triggering the integrated circuit. A spike could be caused by a very strong magnetic pulse nearby as
that can induce an extra voltage in the battery wires.

The lower part of the diagram shows the output voltages produced as the clock pulses reach pin 14 of the chip.
The positive-going part of the clock signal triggers the change in state of the outputs. If necessary, a positive-
going pulse on the reset pin, pin 15, causes output “0” to go high and the other outputs to go low.

The 4017 Divide-by-Ten Chip.
Now, to take this output sequencing a little further. For example, the Charles Flynn magnet motor shown in
Chapter 1 needs coils to be powered up, one after the other and only one should be on at any one time. This
calls for a circuit which has a large number of outputs. The CD4022BC chip gives up to eight outputs one after
the other. The CD4017B chip gives up to ten outputs one after the other but there is no need to be limited by
these numbers as more than one chip can be used. If you find this section difficult to understand, then just skip
past to the next section as it is not important for you to understand these larger circuits.

The pin connections for the divide-by-ten CD4017B chip is shown here:

![CD4017B Pin Connections]

While this shows outputs 1 to 10, the manufacturers and some people who draw circuits, prefer to label the
outputs as “0 to 9” which correspond to digital displays. In our style of operation, it is easier to think of the ten
outputs as being from 1 to 10.

You will notice that there are two pin labels which we have not come across before, namely, the “Carry-out” pin
and the “Clock Enable” pin. These allow us to use several of these chips in a row to give a much larger “divide-
by” number. The “Clock Enable” pin can be used to block the clock input. The operation is like this:
In this example, the sequence is started by the Reset pin being given a high voltage as shown by the green shading. This pushes the output pin 1 to a high voltage and all of the other outputs to a low voltage and holds those voltages as long as the reset voltage is high.

When the Reset voltage drops, the next rising edge of the clock pulse (marked “1” in the diagram) causes the output 1 to go low and output 2 to go high. Each of the successive clock pulses “2” to “9” moves the high voltage steadily along the outputs until output pin 10 is high.

The next clock pulse rising edge (marked “10” in the diagram) starts the sequence again with output 10 going low and output 1 going high again. If nothing changes, then that sequence of output voltage changes will continue indefinitely.

However, in the diagram above, the Clock Enable pin voltage is driven high on clock pulse “11”. Output 2 has just gone high and would have gone low when the rising edge of clock pulse “12” occurred, but in this case, the Clock Enable feature blocks the clock pulse and prevents it reaching the rest of the circuitry. This causes the output 2 voltage to stay high as long as the Clock Enable remains high. In this example, the Clock Enable voltage stays high for just one clock pulse, causing the output 2 voltage to be high for twice its usual length, and then the sequence continues as before.

**A Divide-by-Twenty-five Circuit.**
Here is one way to get a large “divide-by” number. This example is divide-by-25 because there is only one ‘intermediate stage’ but there can be any number and each additional one adds another eight outputs to the total:
At startup, output 10 of the first stage (which is physical pin 11 of the chip) is at a low voltage. This holds the Clock Enable (pin 13) low, allowing the clock pulses to enter the first stage. Because the output 10 voltage is low, one input to the first AND gate is held low, preventing it from letting the clock pulse flow through it, i.e. the “gate” is closed to through traffic.

The first stage chip then operates as normal, producing outputs 1 to 9 in order as you would expect. The next clock pulse sets the first stage output 10 high, allowing the clock pulses through the first AND gate and holding the Clock Enable (pin 13) high, which in turn locks the output 10 high, dropping the first stage chip out of the operation.

As the output 1 of the first stage is connected to the Reset (pin 15) of the second chip, it will have been cleared and it’s output 1 set high, which in turn Resets the third chip and closes the second AND gate. So, when the first pulse gets through to the second chip, it pushes it from state 1 to state 2 where the output 2 goes high. For that reason, output 1 of the second chip is not one of the outputs which can be used by whatever following circuitry you choose to connect to this system. Consequently, only eight of the ten outputs of the second chip are available as counter outputs. That is, outputs 1 and 10 are taken up in passing the switching sequence between the various chips in the chain.

The same applies to all following chips in the chain, each extra chip adding up to eight extra sequential outputs. On the final stage chip, if you connect the red Reset wire (which goes back to fire up the first chip again) to output 9 instead of output 10 of the final chip, then you get a divide-by-24 result.

If the Reset is taken from output 8 of the final chip, then you get a divide-by-23 result, and so on. Using this method, you can have a divide-by-circuit for any number you want. These chips are very popular and so their cost is low, making the entire circuit cheap to make. The pin connections for the AND gates is shown here:

The PIC Revolution.

Over the years, there have been advances in the way that circuitry can be put together, prototypes built and tested. Initially, “valves” or “vacuum tubes” were used and circuits required a good deal of electrical power in order to operate. Mechanical vibrators or “reeds” were used to generate the switching needed to convert DC into AC. Then the transistor became widely available and the transistor replaced the mechanical vibrator reed, the circuit being called an “astable multivibrator” and comprising of two transistors wired back to back (as described in chapter 12). Then came the digital integrated circuit with it’s “NOR gates” which could also be wired back to back to make a multivibrator. This was done so often that a special integrated circuit called the “555 chip” was designed to do the job all on its own. That chip has been a tremendous success and is now found in all sorts of different circuits, being very easy to use, very robust and very cheap. Surprisingly, the dominant position of the “555” chip is being challenged by a completely different type of chip, one which is essentially, a computer on a
single chip, and which is called a "PIC controller".

This new type of chip is not expensive, is easy to use, and can be changed to perform a different task in just a few seconds. It can perform timing tasks. It can act as a multivibrator. It can act as a "Divide-by-N" chip. It is a very impressive chip which is very useful. The reason that I mention it here is because it is at the heart of the fastest working Tesla Switch research forum around (the "energetic forum" group). The chip is something you need to know about as it will certainly take over more and more circuit applications in the coming years.

There is a whole family of these processor chips, but I will select just one for this description, and that will be the one being used by the "energetic forum" members, and I have to thank Jeff Wilson for his help in describing this circuitry, the programming and the methods which he uses.

First, however, some information on this new design of chip and the methods used with it. The one used by Jeff is called the "PICAXE-18X" and it looks like the chip shown here. From which you can see, it looks just like any other chip, although with eighteen pins. The powerful performance comes from the way that it operates. You are probably familiar with the "555" chip and understand that it operates by changing the voltage on just one of its pins (pin 3) the output pin, from a low voltage to a high voltage. The PIC chip can do that as well, but even better still, it has more than one output pin and it can alter the voltage on any of those pins to either a high or a low voltage and it can do that in any order and with any timing that you choose. This makes it a very versatile chip indeed and one which is very well suited to be the central controller for a Tesla Switch test environment.

The chip is used by wiring it into a circuit in the same sort of way that a 555 chip would be used, except that the PIC has its own internal timing clock and can operate in intervals of one thousandth of a second, that is, one millisecond.

The top eight pins are for making the chip work. The next two are for providing the chip with electrical power. The bottom eight pins are separate outputs, any one of which can operate switches, timers, etc., just as the output from a 555 chip can. Having been named by computer people, instead of the eight output pins being numbered from 1 to 8 as any rational person would do, they have numbered them from 0 to 7.

The voltage on those output pins will be either High or Low. PIC switching can be used with a wide range of different free-energy designs. The PIC chip is generally supplied with a socket, a connecting cable and a program for feeding instructions into the chip. The feed is generally from an ordinary PC. The programming instructions are very simple and anyone can learn how to use them in just a few minutes.

So let's look at a circuit which has been used by Jeff when he tests prototype circuitry. The first part of the circuit is for connecting the standard PC socket to the PIC chip and it looks like this:

![PICAXE-18X Circuit Diagram]

**PICAXE-18X**

```
ADC 2 / Input 2  1  18  Input 1 / ADC 1
Serial Out  2  17  Input 0 / ADC 0 / Infrain
Serial In  3  16  Input 7 / keyboard data
Reset  4  15  Input 6 / keyboard clock
5V  5  14  +V
Output 0  6  13  Output 7
i2c sda / Output 1  7  12  Output 6
Output 2  8  11  Output 5
pwm 3 / Output 3  9  10  Output 4 / i2c sol
```

The top eight pins are for making the chip work. The next two are for providing the chip with electrical power. The bottom eight pins are separate outputs, any one of which can operate switches, timers, etc., just as the output from a 555 chip can. Having been named by computer people, instead of the eight output pins being numbered from 1 to 8 as any rational person would do, they have numbered them from 0 to 7.

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So let's look at a circuit which has been used by Jeff when he tests prototype circuitry. The first part of the circuit is for connecting the standard PC socket to the PIC chip and it looks like this:
A standard 9-pin computer socket has its pin 2 connected to the PIC's pin 2, pin 3 connected to the PIC's pin 3 via a 10K / 22K voltage divider resistor pair (which lowers the incoming signal voltage), and pin 5 is connected to the PIC's pin 5. That is all that's needed to feed information into the PIC chip.

The chip is supplied from a 12-volt battery but as it needs a 5-volt supply, the 100 / 150 ohm (2 watt) resistor pair is used to drop the 12 volts down to about 7 volts and then the 5.1-volt zener diode clamps the voltage at 5.1 volts, which is just what the chip needs. The tiny 10 nF (0.01 microfarad) capacitor is there to trap any voltage spikes should any be picked up from some outside influence. Finally, the press-button switch used to short between pins 4 and 5 is used to wipe out the program inside the PIC, ready for a new program to be loaded.

The actual programming is not difficult and the feed into the chip is handled by the program supplied with the chip and which is run on your home computer. Let's take an example. Suppose we want the output on pin 10 to act as a clock signal. The people who made the chip expect that pin to be called "Output 4" in the program. Please don't ask me why it isn't called "10" in the program as I have no answer for you other than "it takes all sorts of people to make a world".

All right, suppose we want to produce an output signal like a 555 chip running at 50 Hz. We choose one of our output pins, say, the physical pin 10, that being the bottom right hand pin on the chip. As you can see from the pin diagram of the chip shown above, pin 10 is called "Output 4" in a set of commands, or just "4" to save typing. The program might be:

```
Main:
    high 4
    pause 10
    low 4
    pause 10
    goto Main
```

Wow - really difficult stuff!! Only a genius could manage to program! Well, we'll see if we can struggle along with this "difficult" stuff.

The "Main:" at the start is a "label" which can be jumped to and that is done by the "goto Main" command which sends the chip back to repeat the commands in the loop indefinitely (or until the chip is powered down).

The second line "high 4" tells the chip to put the maximum possible voltage on the "Output 4" which is the physical pin 10 of the chip. The chip does this immediately, with no time delay.

If we want the output to give a 50 Hz output signal, then the voltage on our chosen output pin will have to go high, pause, go low, pause and go high again, 50 times each second. As there are 1,000 milliseconds in one second, and the chip's clock runs with 1 millisecond ticks, then we need our complete cycle of "up, pause, down, pause" to happen 50 times in those 1,000 clock ticks. That is, once every 20 ticks, so each delay will be 10 clock ticks long.

The third line "pause 10" tells the chip to sit on it's hands and do nothing for the next 10 ticks of it's internal clock (which ticks 1,000 times per second).

The fourth line "low 4" tells the chip to lower the output voltage on it's "Output 4" (pin 10 in real life) to it's minimum value.

The fifth line "pause 10" tells the chip to wait for 10 milliseconds before doing anything else.
The last line "goto Main" tells the computer to go back to the label "Main:" and continue with whatever instructions follow that label. This puts the chip into an 'infinite loop' which will make it generate that output waveform continuously. The output will look like this:

![Waveform Diagram](image)

This gives an even waveform, that is, one with a Mark/Space ratio of 50:50 or a Duty Cycle of 50%. If we want the same rate of pulsing but a Duty Cycle of just 25% then the program would be:

```
Main:
  high 4
  low 4
  pause 5
  pause 15
  goto Main
```

which produces this waveform:

![Waveform Diagram](image)

If you wanted "Output 7" (physical pin 13) to do the reverse of this at the same time - that is, when Output 4 goes high we want Output 7 to go low, and vice versa, then, for a 20% Duty Cycle the program would be:

```
Main:
  high 4
  low 4
  high 7
  low 7
  pause 4
  pause 16
  goto Main
```

These output voltages are then used in exactly the same way as the output voltages on pin 3 of a 555 chip, or any of the outputs of NAND gates, Hall-effect sensor chips, Schmitt triggers, or whatever. If the device to be powered requires very little current, then the easiest method is to connect the load directly to the output pin.

If, as is most often the case, the device to be powered needs a large current to make it work, then the output voltage is used to power a transistor, perhaps like this:
Here, the resistor "R1" limits the current fed into the base of the transistor when pin 10 goes high, but allowing enough current for the transistor to switch on fully, powering the load. The resistor "R" makes sure that the transistor switches off fully when the output on pin 10 goes low. The circuit as shown restricts the load to some piece of equipment which can operate on just five volts, so an alternative circuit could be:

This allows whatever voltage the load needs to be applied to the load, while the PIC chip remains running on its normal 5-volt supply. However, the equipment to be powered may not be able to have a common zero voltage connection with the PIC. To deal with this, an optical isolation chip can be used like this:

Here a high output voltage on pin 10 of the PIC chip lights up the LED inside the opto-isolator chip, causing a major drop in the resistance between the other two pins. This causes a current controlled by the resistor "R" to be fed into the base of the transistor, switching it on and powering the load.

Recently, a very popular programmable chip has been introduced. It is called the "Arduino" and it is fast and versatile and very popular with experimenters. There is an extensive set of English-language Video tutorials on the Arduino chip, the first in the series by Jeremy Blum is [http://www.youtube.com/watch?v=fCxzA9_kg6s](http://www.youtube.com/watch?v=fCxzA9_kg6s). The board looks like this:
Capacitors.
We have avoided mentioning capacitors in any detail as it has not been necessary for understanding the circuitry covered so far. Capacitors come in many sizes, types and makes. Their size is stated in ‘Farads’ but as the Farad is a very large unit, you are unlikely to encounter a capacitor marked in anything larger than a microfarad, which is a millionth of a Farad. The symbol for a microfarad is μF where ‘μ’ is the letter of the Greek alphabet. This is a pain for normal text production as Greek letters do not occur in your average font. Some circuit diagrams give up on ‘μ’ and just write it as uF which looks like μu-F slightly mis-printed where the descender of the μ has not printed.

Anyway, very large capacitors which you may encounter range from 5,000 microfarads to maybe as much as 20,000 microfarads. Large capacitors range from 10 microfarads to 5000 microfarads. Medium sized capacitors run from 0.1 microfarad to about 5 microfarads and small capacitors are those below 0.1 microfarad.

1000 nanofarads (‘nF’) = 1 microfarad.
1000 picofarads (‘pF’) = 1 nanofarad

So:
0.01 microfarad can be written as 10nF
0.1 microfarad can be written as 100nF
0.1nF can be written as 100pF

Capacitors larger than 1 microfarad tend to be ‘polarised’. In other words, the capacitor has a ‘+’ connector and a ‘−’ connector, and it does matter which way round you connect it. The larger capacitors have a voltage rating and this should not be exceeded as the capacitor can be damaged and possibly even totally destroyed. Capacitors can be added together, but surprisingly, they add in the reverse way to resistors:

Example 1

Example 2
If two capacitors are wired in series, as shown in Example 1 above, the overall capacity is reduced while the voltage rating increases. The reduction in capacitance is given by:

\[
1/C_t = 1/C_1 + 1/C_2 + 1/C_3 + \ldots
\]

In Example 1, then, \(1/\text{total capacitance} = 1/100 + 1/100\) or \(1/C_t = 2/100\) or \(1/C_t = 1/50\) so the overall capacitance reduces from 100 microfarads to 50 microfarads. The advantage in wiring the capacitors like this is that the voltage rating has now increased to 32V (16V across each of the capacitors).

In Example 2, the overall capacitance has reduced to a third of 100 microfarads but the voltage rating has tripled.

In Example 3, the capacitors are wired in parallel. The voltage rating is unchanged but the overall capacitance is now the sum of the three capacitors, namely 300 microfarads.

There is no need for the capacitors to have similar values, they are merely shown that way in the examples to make the arithmetic easier and not distract you from the ways in which the capacitors interact together.

 Occasionally, a circuit needs a large capacitor which is not polarised. This can be provided by placing two polarised capacitors back-to-back like this:

\[
\begin{align*}
\text{Example 4} \\
100 \mu F 16V & = 50 \mu F 32V \\
100 \mu F 16V
\end{align*}
\]

When the capacitors are connected this way, it does not matter which end of the pair is connected to the positive side of the circuit and which to the negative side.

Large capacitors usually have their capacitance and voltage printed on the outside of the capacitor, but small capacitors are usually far too tiny for that to be an option. So, a code very much like that used for resistors is used for small capacitors. The code is a 2-digit code for capacitors up to 100 picofarads and for higher values it is a 3-digit code where the first two digits are the numerical value of the capacitor in picofarads and the third digit is the number of zeros following the two digits. One thousand picofarads (pF) is one nanofarad (nF) and one thousand nanofarads is one microfarad. These are some common values:

<table>
<thead>
<tr>
<th>Value</th>
<th>Code</th>
<th>Value</th>
<th>Code</th>
</tr>
</thead>
<tbody>
<tr>
<td>10 pF</td>
<td>10</td>
<td>2.2 nF</td>
<td>222</td>
</tr>
<tr>
<td>22 pF</td>
<td>22</td>
<td>4.7 nF</td>
<td>472</td>
</tr>
<tr>
<td>47 pF</td>
<td>47</td>
<td>10 nF</td>
<td>103</td>
</tr>
<tr>
<td>100 pF</td>
<td>101</td>
<td>22 nF</td>
<td>223</td>
</tr>
<tr>
<td>220 pF</td>
<td>221</td>
<td>47 nF</td>
<td>473</td>
</tr>
<tr>
<td>470 pF</td>
<td>471</td>
<td>100 nF</td>
<td>104</td>
</tr>
<tr>
<td>1 nF</td>
<td>102</td>
<td>220 nF</td>
<td>224</td>
</tr>
</tbody>
</table>
The time has come for a serious warning: High voltages are very, very dangerous. Do not become so familiar with them that you treat them casually. High voltages can kill you. Capacitors are capable of building up high voltages and some good makes can hold the charge for several days.

In particular, do not try to make adjustments to, or take parts from, the inside of a TV set. A black and white TV set uses 18,000 Volts on the magnetic coils used to create the moving picture on the tube. A capacitor inside the set may well have that voltage on it three days after the set was last used. Don't fool around inside a TV set, it could kill you quick, or if you are really unlucky, it could injure you for life. A colour TV set uses 27,000 Volts to operate the coils inside it and that will fry you in jig time if you touch it.

Also, please don't think that you are safe if you don't quite touch it; 27,000 volts can jump across a gap to your hand. If you try to discharge a TV capacitor using a metal screwdriver with a wooden handle, please ensure that you medical insurance is up to date before you do it. You can receive a hefty shock through the screwdriver handle.

Voltages up to 24 Volts should be quite safe. However, some circuits will generate very high voltages even though the battery driving the circuit is low voltage. A standard off-the-shelf inverter circuit produces 240 Volts AC from a 12 Volt battery. Just because the battery is only 12 Volts does not mean that the circuit is not dangerous. Circuits which have inductors in them can produce high voltages, especially if they contain large capacitors. The voltage which produces the spark in your car engine is very high and it comes from the 12-volt car battery. You know enough about this by now, so pay attention!

The more advanced stuff:
You do not need to bother with this section if you are just starting out with some basic switching circuits of the type already described in this tutorial, so please feel free to skip this section and move on to the “Prototype Construction” section which you will find immediately useful.

This section is a lightweight introduction to Alternating Current circuits and pulsed DC circuits. Let me stress again that I am mainly self-taught and so this is just a general introduction based on my present understanding.

AC Power Factors.
Alternating Current, generally called “AC” is called that because the voltage of this type of power supply is not a constant value. A car battery, for instance, is DC and has a fairly constant voltage usually about 12.8 volts when in it’s fully charged state. If you connect a voltmeter across a car battery and watch it, the voltage reading will not change. Minute after minute it says exactly the same because it is a DC source.

If you connect an AC voltmeter across an AC power supply, it too will give a steady reading, but it is telling a lie. The voltage is changing all the time in spite of that steady meter reading. What the meter is doing is assuming that the AC waveform is a sine wave like this:

and based on that assumption, it displays a voltage reading which is called the “Root Mean Square” or “RMS” value. The main difficulty with a sine wave is that the voltage is below zero volts for exactly the same length of time as it is above zero volts, so if you average it, the result is zero volts, which is not a satisfactory result because you can get a shock from it and so it can’t be zero volts, no matter what the arithmetical average is.

To get over this problem, the voltage is measured thousands of times per second and the results squared (that is, the value is multiplied by itself) and then those values are averaged. This has the advantage that when the voltage is say, minus 10 volts and you square it, the answer is plus 100 volts. In fact, all of the answers will be positive, which means that you can add them together, average them and get a sensible result. However, you end up with a value which is far too high because you squared every measurement, and so you need to take the square root of that average (or “mean”) value, and that is where the fancy sounding “Root Mean Square” name comes from – you are taking the (square) root of the (average or) mean value of the squared measurements.

With a sine wave like this, the voltage peaks are 41.4% higher than the RMS value which everyone talks about. This means that if you feed 100 volts AC through a rectifier bridge of four diodes and feed it into a capacitor the
capacitor voltage will not be 100 volts DC but instead it will be 141.4 volts DC and you need to remember that when choosing the voltage rating of the capacitor. In that instance I would suggest a capacitor which is made to operate with voltages up to 200 volts.

You probably already knew all of that, but it may not have occurred to you that if you use a standard AC voltmeter on a waveform which is not a sine wave, that the reading on the meter is most unlikely to be correct or anywhere near correct. So, please don’t merrily connect an AC voltmeter across a circuit which is producing sharp voltage spikes like, for instance, one of John Bedini’s battery pulsing circuits, and think that the meter reading means anything (other than meaning that you don’t understand what you are doing).

You will, hopefully, have learned that power in watts is determined by multiplying the current in amps by the voltage in volts. For example, 10 amps of current flowing out of a 12 volt power supply, represents 120 watts of power. Unfortunately, that only holds true for circuits which are operating on DC, or AC circuits which have only resistors in them. The situation changes for AC circuits which have non-resistive components in them.

The circuits of this type which you are likely to come across are circuits which have coils in them, and you need to think about what you are doing when you deal with these types of circuit. For example, consider this circuit:

![Circuit Diagram](image1)

This is the output section of a prototype which you have just built. The input to the prototype is DC and measures at 12 volts, 2 amps (which is 24 watts). Your AC voltmeter on the output reads 15 volts and your AC ammeter reads 2.5 amps and you are delighted because 15 x 2.5 = 37.5 which looks much bigger than the 24 watts of input power. But, just before you go rushing off to announce on YouTube that you have made a prototype with COP = 1.56 or 156% efficient, you need to consider the real facts.

This is an AC circuit and unless your prototype is producing a perfect sine wave, then the AC voltmeter reading will be meaningless. It is just possible that your AC ammeter is one of the few types that can accurately measure the current no matter what sort of waveform is fed to it, but it is distinctly possible that it will be a digital meter which assesses current by measuring the AC voltage across a resistor in series with the output, and if that is the case, it will probably be assuming a sine wave. The odds are that both readings are wrong, but let’s take the case where we have great meters which are reading the values perfectly correctly. Then the output will be 37.5 watts, won’t it? Well, actually, no it won’t. The reason for this is that the circuit is feeding the transformer winding which is a coil and coils don’t work like that.

The problem is that, unlike a resistor, when you apply a voltage across a coil the coil starts absorbing energy and feeding it into the magnetic field around the coil, so there is a delay before the current reaches it’s maximum value. With DC, this generally doesn’t matter very much, but with AC where the voltage is changing continuously, it matters a great deal. The situation can be as shown in this graph of both voltage and current:

![Graph Diagram](image2)

At first, this does not look like any great problem, but it has a very significant effect on the actual power in watts. To get the 37.5 watts output which we were talking about earlier, we multiplied the average voltage level by the average current level. But these two values do not occur at the same time and that has a major effect.

As this can be a little difficult to see, let’s take the peak values rather than the averages as they are easier to see. Let’s say that in our example graph that the voltage peak is 10 volts and the current peak is 3 amps. If this were
DC we would multiply them together and say that the power was 30 watts. But with AC, this does not work due to the timing difference:

When the voltage is peaking, the current is nowhere near its peak value of 3 amps:

As a result of this, instead of getting our expected peak power at the top of the voltage peak, the actual power in watts is very much lower – less than half of what we were expecting. Not so good, but it gets worse when you look at the situation more closely. Take a look at what the voltage is when the current crosses the zero line, that is, when the current is zero. The output power is zero when the current is zero but this occurs when the voltage is at a very high value:

The same goes for when the voltage is zero. When the voltage is zero, then the power is also zero, and you will notice that this occurs when the current is at a high value:

The power is **not** the average current multiplied by the average voltage if there is a coil involved in the circuit – it will be less than that by an amount known as the “power factor” and I’ll leave you to work out why it is called that.

So, how do you determine what the power is? It is done by sampling the voltage and current many times per second and averaging those combined results:
Both the voltage and the current are sampled at the times indicated by the vertical red lines and those figures are used to calculate the actual power level. In this example, only a few samplings are shown, but in practice, a very large number of samples will be taken. The piece of equipment which does this is known as a wattmeter as it measures watts of power. The sampling can be done by windings inside the instrument, resulting in an instrument which can be damaged by overloading without the needle being anywhere near full deflection, or it can be done by digital sampling and mathematical integration. Most digital sampling versions of these meters only operate at high frequencies, typically over 400,000 cycles per second. Both varieties of wattmeter can handle any waveform and not just sine waves.

The power company supplying your home measures the current and assumes that the full voltage is present all of the time that the current is being drawn. If you are powering a powerful electric motor from the mains, then this current lag will cost you money as the power company does not take it into account. It is possible to correct the situation by connecting one or more suitable capacitors across the motor to minimise the power loss.

With a coil (fancy name “inductor” symbol “L”), AC operation is very different to DC operation. The coil has a DC resistance which can be measured with the ohms range of a multimeter, but that resistance does not apply when AC is being used as the AC current flow is **not** determined by the DC resistance of the coil alone. Because of this, a second term has to be used for the current-controlling factor of the coil, and the term chosen is “impedance”. The wire in any coil has a resistance and that opposes current flow through the coil irrespective of whether the voltage applied to the coil is DC or AC. The capacitance between the neighbouring turns of wire in a coil, introduces a feature of the coil which “impedes” AC current flow through the coil and the amount of that impedance depends on the frequency of the AC voltage being applied to the coil.

The impedance of a coil depends on it’s size, shape, method of winding, number of turns and core material. If the core is made up of iron or steel, (usually thin layers of iron which are insulated from each other), then it can only handle low frequencies. You can forget about trying to pass 10,000 cycles per second (“Hz”) through the coil as the core just can’t change it’s magnetisation fast enough to cope with that frequency. A core of that type is ok for the very low 50 Hz or 60 Hz frequencies used for mains power, which are kept that low so that electric motors can use it directly.

For higher frequencies, ferrite can be used for a core and that is why some portable radios use ferrite-rod aerials, which are a bar of ferrite with a coil wound on it. For higher frequencies (or higher efficiencies) iron dust encapsulated in epoxy resin is used. An alternative is to not use any core material and that is referred to as an air-core coil. These are not limited in frequency by the core but they have a very much lower inductance for any given number of turns. The efficiency of the coil is called it’s “Q” (for “Quality”) and the higher the Q factor, the better. The resistance of the wire lowers the Q factor.

A coil has inductance, and resistance caused by the wire, and capacitance caused by the turns being near each other. However, having said that, the inductance is normally so much bigger than the other two components that we tend to ignore the other two. Something which may not be immediately obvious is that the impedance to AC current flow through the coil depends on how fast the voltage is changing. If the AC voltage applied to a coil completes one cycle every ten seconds, then the impedance will be much lower than if the voltage cycles a million times per second.

If you had to guess, you would think that the impedance would increase steadily as the AC frequency increased. In other words, a straight-line graph type of change. That is not the case. Due to a feature called resonance, there is one particular frequency at which the impedance of the coil increases massively. This is used in the tuning method for AM radio receivers. In the very early days when electronic components were hard to come by, variable coils were sometimes used for tuning. We still have variable coils today, generally for handling large currents rather than radio signals, and we call them “rheostats” and some look like this:
These have a coil of wire wound around a hollow former and a slider can be pushed along a bar, connecting the slider to different winds in the coil depending on its position along the supporting bar. The coil connections are then to the slider and to one end of the coil. The position of the slider effectively changes the number of turns of wire in the part of the coil which is in the circuit. Changing the number of turns in the coil, changes the resonant frequency of that coil. AC current finds it very, very hard to get through a coil which has the same resonant frequency as the AC current frequency. Because of this, it can be used as a radio signal tuner:

![Diagram of radio signal tuner](image)

If the coil’s resonant frequency is changed to match that of a local radio station by sliding the contact along the coil, then that particular AC signal frequency from the radio transmitter finds it almost impossible to get through the coil and so it (and only it) diverts through the diode and headphones as it flows from the aerial wire to the earth wire and the radio station is heard in the headphones. If there are other radio signals coming down the aerial wire, then, because they are not at the resonant frequency of the coil, they flow freely through the coil and don’t go through the headphones.

This system was soon changed when variable capacitors became available as they are cheaper and more compact. So, instead of using a variable coil for tuning the radio signal, a variable capacitor connected across the tuning coil did the same job:

![Diagram of radio signal tuner with variable capacitor](image)

**Resonance.**

While the circuit diagram above is marked “Tuning capacitor” that is actually quite misleading. Yes, you tune the radio receiver by adjusting the setting of the variable capacitor, but, what the capacitor is doing is altering the resonant frequency of the coil/capacitor combination and it is the resonant frequency of that combination which is doing exactly the same job as the variable coil did on its own.

This draws attention to two very important facts concerning coil/capacitor combinations. When a capacitor is placed across a coil “in parallel” as shown in this radio receiver circuit, then the combination has a very high impedance (resistance to AC current flow) at the resonant frequency. But if the capacitor is placed “in series” with the coil, then there is nearly zero impedance at the resonant frequency of the combination:

![Diagram showing high and low resistance at resonance](image)

This may seem like something which practical people would not bother with, after all, who really cares? However, it is a very practical point indeed. In Chapter 3, some of the very high-power devices produced by Don Smith are described. Typically, he uses an off-the-shelf neon-tube driver module as an easy way to provide a high-voltage, high-frequency AC current source, typically, 6,000 volts at 30,000 Hz. He then feeds that power into a Tesla Coil which is itself, a power amplifier. The arrangement is like this:
People who try to replicate Don’s designs tend to say “I get great sparks at the spark gap until I connect the L1 coil and then the sparks stop. This circuit can never work because the resistance of the coil is too low”.

If the resonant frequency of the L1 coil does not match the frequency being produced by the neon-tube driver circuit, then the low impedance of the L1 coil will definitely pull the voltage of the neon-tube driver down to a very low value. But if the L1 coil has the same resonant frequency as the driver circuit, then the L1 coil (or the L1 coil/capacitor combination shown on the right, will have a very high resistance to current flow through it and it will work well with the driver circuit. So, no sparks, means that the coil tuning is off. It is the same as tuning a radio receiver, get the tuning wrong and you don’t hear the radio station.

Choosing components which are not specified.

Some people find it difficult to select a suitable component where the exact component is not specified or where an alternative has to be selected, so perhaps a few general pointers might be helpful. The reason why component values are omitted may well be because a very wide range of alternative values can be used and if one particular is specified, the newcomers to electronics feel that they have to use that one value or the circuit will not work, (which is almost never the case). For example, I have been asked if a capacitor rated at 25V could be used instead of the same value capacitor rated at 16V shown in the circuit, to which the answer is ‘yes, most definitely’. The lower voltage rating is adequate and the component cheaper to buy, but if one of a higher voltage rating is available, then it can be used.

With capacitors, you need to consider the physical size and wire connections, the capacitance, the voltage rating, and the leakage. The cost and size of a capacitor is directly related to it’s voltage rating, and once the voltage rating exceeds that normally used, the price shoots up rapidly as the sales volume reduces rapidly, which in turn, discourages further sales. This sometimes causes circuit builders to connect chains of cheaper capacitors together to make a smaller-capacity high-voltage capacitor. In the case of Tesla Coil builders, they then may connect several of these chains in parallel to boost the capacitance.

If the voltage rating is exceeded (usually by a very large amount), the capacitor will be damaged and become either a short-circuit, or more likely, an open circuit. Either way, it will never work as a capacitor again. In a household circuit, where the capacitor is being used as part of the power supply to the circuit, the voltage rating does not need to be much higher than the supply voltage, with say, 16V being used for a 12V circuit. You could use a capacitor rated at 25V, 40V, 63V, 100V or 400V and it would work perfectly well, but it will be much larger and have cost much more. But, if you have one sitting around and not being used, there is no reason why you should not use it rather than paying to buy another one.

If the capacitor is being used in a timing circuit where a high-value resistor is feeding current to it, then the leakage current of the capacitor becomes very important. Electrolytic capacitors are seldom suitable for such an application as they have a small, unpredictable leakage current which will vary with the age of the capacitor. For accurate timing with a capacitor, ceramic, polypropylene, mylar or tantalum should be used.

The voltage rating for an electrolytic capacitor is for DC, so if you use it for limiting current in an AC power supply, that is, where the current flows through the capacitor rather than the capacitor being placed across the supply and is acting to combat ripple, then great care is needed. The capacitor will heat up due to the power flowing through it, and it is possible for an electrolytic capacitor used in that way to rupture or ‘explode’ due to the electrolyte boiling. Instead, you need to use the very much more expensive oil-filled can capacitors (as shown near the end of chapter 10). That style of usage is unusual for home constructors.

With bi-polar transistors, you need to use commonsense. Suppose a 555 timer chip is required to power a transistor which controls a relay:
For the moment, we will ignore the fact that the 555 could drive the relay directly without the need for a transistor. Let’s say that the relay draws a current of 30 mA when connected to a 12V supply. Therefore, the transistor needs to be able to handle a current of 30 mA. Any small switching transistor such as the BC109 or 2N2222 can easily handle that current. The transistor also needs to be able to handle 12 volts. If in doubt, look up the characteristics of your choice of transistor at http://www.alldatasheet.co.kr/ by entering the transistor name ‘BC109’ or whatever in the entry box at the top of the screen and clicking on the button to the right of it. Eventually, it will let you download a pdf document specifying the transistor, and that will show you the voltages which the transistor can handle. Both of the above transistors can handle far more than 12V.

The next question is, ‘can the transistor switch fast enough to work in this circuit?’ and the data sheet will show that they can switch on and off a million times per second. As the relay can only switch on and off a few times per second, the transistor can easily operate fast enough to handle the switching.

Next, we need to know what size of resistor would be suitable. The data sheet will also show the DC current gain of the transistor. This is usually marked as “hfe” and for these transistors is likely to be a minimum of, say, 200. This means that the current flowing into the base of the transistor needs to be one two-hundredth of the relay’s 30 mA which is 0.15 mA. The resistor will have about +11 volts at pin 3 of the 555 timer and around +0.7 volts at the base of the transistor when it is switched fully on. That means that the resistor will have about 10.3 volts across it when the relay is switched on:

So, what size of resistor will have 0.15 mA flowing through it when there is a 10.3-volt drop across it? We know that a 1K resistor passes 1mA per volt and so would pass 10.3 mA with 10.3 volts across it. That is far more than we need. A 10K resistor would pass 1.03 mA which is still far too much but certainly could be used. As it is a resistor, we can use Ohm’s Law: \( R = \frac{V}{A} \) (Ohms equals Volts over Amps), or \( R = \frac{10.3}{0.00015} \) which is 68K. So, any resistor between 68K and perhaps 15K should work well.

The diode is there to protect the transistor from excessive voltage caused by the coil of the relay. When a coil is switched off suddenly, it generates a reverse voltage which can be hundreds of volts, pulling the collector of the transistor far above the +12V power supply line. When that starts to happen, it effectively reverses the diode direction, allowing it to conduct and short-circuit that big voltage spike:
Due to the short-circuiting, the voltage can’t get any higher and the current through the diode is not large, so most diodes such as the popular and cheap 1N4001 or 1N4007 types can be used.

When a transistor is connected like that and switched on, it is effectively a short-circuit between it’s collector and emitter, and that places the full 12 volts across the relay, powering it very solidly. This connection method is called a “common-emitter” circuit because all of the transistors used have their emitters all wired in common to the 0V line. An alternative arrangement is the “emitter-follower” circuit:

With this circuit arrangement, the emitter of the transistor “follows” the voltage on pin 3 of the 555 timer. It is always a constant voltage below it, typically about 0.7 volts. The output of the 555 timer has a maximum of about 0.7V below the supply voltage, and so it’s maximum value is about 11.3V in this circuit. The transistor drops that by a further 0.7V, which means that the relay only gets about 10.6V across it instead of the full 12V of the supply, which means that it should be a 10-volt relay rather than a 12-volts relay.

Those are the easy cases because the 555 timer can supply at least 200 mA through it’s output pin, while keeping the output voltage steady. That is not the case with simple transistor circuits. Take a situation like this:

For audio work - microphone pre-amplifiers and the like – the rule of thumb is that the current flowing through the first transistor should be at least ten times the current required by the base of the second transistor in order not to drag down and distort the audio waveform.
Relay switching is not so critical but the same general principle applies and attention needs to paid to the collector resistor of the preceding transistor. For example, if the current flowing through the preceding transistor is small, say, 0.5 mA and the output transistor needs 1.5 mA flowing into it’s base, then there can be a problem. In this circuit, for example:

Here, the voltage at point “A” goes high because the first transistor switches off and so becomes the same as a resistor of 1Meg or more. Normally, that resistance is so much greater than the 27K of it’s resistor, that the voltage at point “A” would be nearly +12V, but if you were to connect the resistor “R” of just 1K in value, then the situation is changed completely. The base of “Tr” can’t rise above 0.7V. The first transistor can be ignored due to its very high resistance. That leaves a voltage-divider pair of resistors, the 27K and the 1K, with 11.3 volts across them, slopping the voltage at point “A” from rising above 1.13V instead of the original 12V and transistor “Tr” will only get 0.43 mA instead of the 1.5 mA which was wanted. The transistor “Tr” has effectively a 28K resistor feeding it current from the +12V rail.

One solution would be to raise the current through the first transistor by using a resistor a good deal smaller than the present 27K. Another option is to lower the input current requirement of transistor “Tr” by making it a Darlington pair or by using a transistor with a much higher gain.

**Constructing Prototypes.**
The main options for building a prototype circuit are:

1. A (plug-in) breadboard
2. Electrical screw connector strips.
3. Stripboard
4. A printed circuit board.

1. The typical breadboard unit consists of a matrix of clip holes wired in strips, into which component leads can be pushed to make a circuit. In my opinion, they are best avoided as it takes quite some effort to implement any significant circuit using them, some components do not fit well in the sockets which are small enough to take DIL IC packages, and when you do get a circuit working well on the breadboard, there is no guarantee that it will work well when you attempt to move it to a permanent soldered board:

While a plastic board of this type looks as if it should be quick and easy to use, I have never found it to be so ever since the boards were scaled down in size to take the closely-spaced pins of integrated circuits (“chips”). It is generally difficult to lay the components out in the same pattern as the circuit diagram, and if they are not, then it becomes slow to follow the circuit through on the breadboard layout.
2. The local hardware shop has cheap screw connectors which can be very effective. These come in several sizes and the smaller ones are very convenient for constructing transistor circuitry. They look like this:

Circuits can be assembled very easily, using these connectors and an example might be one of the John Bedini battery pulsing circuits which might have a layout like this:

I have built this circuit using this style of construction and it was very successful indeed, being very quick and easy to construct and it proved to be very tough and effective over a long period of use. The plastic strip has a hole between each connector strip and that allows you to bolt the strip to a base board on which you mount other components, in this case, the pulsing coil and the rotor with the magnets attached. Each connection block can take two or three wires. The wires need to have the insulation removed and the wires scraped clean and shiny if they are not already in that state. If more than one multi-strand wire is being put into one side of a connector, then it is usually best to twist the wires together before tightening the clamping screw. If you want, you can give the twisted wires a thin coat of solder, but this has to be done neatly to avoid producing a joint which is too large to fit into the connector. One connector can be cut out of the strip quite easily, using a pair of scissors or a craft knife. Single connectors can join two wires very effectively without the need to solder them.

While the wire trigger switch is shown as a thin line in the diagram above, it is suggested that it is more convenient to use wires of identical diameter, and if it is not clear which is the beginning and end of a single wire, then an ohmmeter may be used to identify the ends. It is suggested that the cables are stretched out in a long length and then twisted together using an electric drill. I have found that doing that is not very good because the cable near the drill is twisted much more strongly than the rest of the wire. Also, it needs a considerable distance outside to lay out a sufficient length of wire. If you really want to twist the wires together (it is not immediately obvious why you would want to do that), then use two coils of wire and twist them together for a short length by turning the reels over as a pair, then wind the twisted length on to a third spool or temporary holder. This method does not need for you to set long cables (which tangle and catch on things very easily) and it gives uniformly twisted wires which can be prepared when sitting in a small workspace. The coil of 850 turns is wound like this:
The first strand of the coil starts at point “C” at the base of the coil and finishes at point “A” at the top of the coil. This is the coil which drives the motor with point “A” connected to the Plus of the drive battery. The second strand starts at point “D” at the base of the coil and finishes at point “B” with point “B” connected to the transistor’s base resistor. This arrangement generates a magnetic North field at the top of the coil and that pushes against the rotor’s permanent magnet’s North pole which is the one facing the coil. With the implementation which I used to charge a car battery, the wheel rotation was gentle, giving perhaps 200 to 300 pulses per minute to the battery. The speed of the wheel reduced as the battery charge increased and so a glance at the wheel showed the charge state of the battery. It is recommended that the coil core be made up from lengths of 1.5 millimetre diameter copper coated welding rod, but as copper is highly conductive electrically, I prefer to coat each rod with enamel paint to block sideways eddy currents which waste power.

3. Stripboard, usually called ‘Veroboard’ even if it is not made by Vero, is a quick and satisfactory method, although you have to make very tiny solder joints. Please be aware that the fumes from the burning resin when soldering are most definitely not good for your health and should be avoided by making sure that the ventilation is adequate.

4. A printed circuit board is feasible for a one-off prototype and making one will increase your production skills, so it is also a reasonable option if you have the etching and drilling equipment to hand. Buying all of the necessary equipment if you do not have any, can cost a fair amount, but the skills gained are significant and the finished boards looks very professional.

There are several other methods of construction, and many varieties of construction board and stripboard. Simple stripboard will be used in the following descriptions, although the method does apply to many different styles of construction.

The first step is to produce a layout for the components on the board. When designing the layout provision should be made for drilling holes to allow the completed board to be bolted to its case using bolts and insulating pillars to keep the soldered joints clear of all other surfaces.

The circuit diagram of the circuit to be built is the starting point. You might wish to draw a light grid of lines to represent the matrix of holes in the strip board. This helps to visualise the run of the copper strips and the sketch can be made to show the exact number of holes available on the piece of strip board to be used. The strip board looks like this:

So you might wish to produce a layout sketch re-usable drawing like this:
where the horizontal strips are numbered and the vertical lines of holes are also numbered. In this sketch, where
the lines cross, represents a hole in the board. The sketch of a possible physical layout can then be prepared and
it might look like this when seen from the top although the copper strips on the underside of the board are shown
in the sketch:

It is very important when producing a sketch like this, that the copper strips making up the circuit are not
accidentally used to connect components further along the board, without breaking the copper strip between
the two sections of the board. It helps to mark a copy of the circuit diagram when you are sketching a possible
physical layout on the strip board. It might be done like this:

Here, the components just below the diode are ringed to show that they have been marked on the layout sketch
and, if necessary, the copper strip broken to isolate the components. A component worth mentioning in passing,
is the capacitor marked with red in the circuit diagram. This is a decoupling capacitor, fed from the 12V battery
via a resistor and a diode (a diode is not normally used in this part of the circuit).
The decoupling is to provide the 555 chip and drivers with a supply which is reasonably isolated from the heavy current-draw circuit not shown in this small section of the circuit diagram. The pulsating heavy current draw of the rest of the circuit is capable of pulling the battery voltage down slightly many times per second. This creates a voltage ripple on the positive supply line from the battery and to smother the ripple, the resistor and diode are used to feed a large reservoir capacitor which smoothes out the ripple.

The circuit itself is not beyond criticism. Transistor ‘TR2’ and its associated components are redundant since pin 3 of the 555 chip already supplies the required signal (and with higher drive capacity) so the second output line should be taken directly from pin 3 of the 555 chip. This snippet of circuit is only shown here as an example of marking up a circuit diagram when making a components layout sketch.

As the layout sketch is produced, the circuit diagram should be marked off with a highlighting pen to make sure that every part of the circuit diagram has been successfully copied to the sketch. In the example below, not all of the highlighted strip is shown, since it runs off the small section of the board being shown here:

```
= Copper strip break
```

Many electronic components can be damaged by the high temperatures they are subjected to when being soldered in place. I personally prefer to use a pair of long-nosed pliers to grip the component leads on the upper side of the board while making the solder joint on the underside of the board. The heat running up the component lead then gets diverted into the large volume of metal in the pair of pliers and the component is protected from excessive heat. On the same principle, I always use a DIL socket when soldering a circuit board, that way, the heat has dissipated fully before the IC is plugged into the socket. It also has the advantage that the IC can be replaced without any difficulty should it become damaged.

If you are using CMOS integrated circuits in any construction, you need to avoid static electricity. Very high levels of voltage build up on your clothes through brushing against objects. This voltage is in the thousands of volts range. It can supply so little current that it does not bother you and you probably do not notice it. CMOS devices operate on such low amounts of current that they can very easily be damaged by your static electricity. Computer hardware professionals wear an earthing lead strapped to their wrists when handling CMOS circuitry. There is no need for you to go that far. CMOS devices are supplied with their leads embedded in a conducting material. Leave them in the material until you are ready to plug them into the circuit and then only hold the plastic body of the case and do not touch any of the pins. Once in place in the circuit, the circuit components will prevent the build up of static charges on the chip.

Soldering is an easily-acquired skill. Multi-cored solder is used for electronic circuit soldering. This solder wire has flux resin contained within it and when melted on a metal surface, the flux removes the oxide layer on the metal, allowing a proper electrical joint to be made. Consequently, it is important that the solder is placed on the joint area and the soldering iron placed on it when it is already in position. If this is done, the flux can clean the joint area and the joint will be good. If the solder is placed on the soldering iron and then the iron moved to the joint, the flux will have burnt away before the joint area is reached and the resulting joint will not be good.

A good solder joint will have a smooth shiny surface and pulling any wire going into the joint will have no effect as the wire is now solidly incorporated into the joint. Making a good solder joint takes about half a second and certainly not more than one second. You want to remove the soldering iron from the joint before an excessive
amount of heat is run into the joint. It is recommended that a good mechanical joint be made before soldering when connecting a wire to some form of terminal (this is often not possible).

The technique which I use is to stand the solder up on the workbench and bend the end so that it is sloping downwards towards me. The lead of the component to be soldered is placed in the hole in the strip board and gripped just above the board with long-nosed pliers. The board is turned upside down and the left thumb used to clamp the board against the pliers. The board and pliers are then moved underneath the solder and positioned so that the solder lies on the copper strip, touching the component lead. The right hand is now used to place the soldering iron briefly on the solder. This melts the solder on the joint, allowing the flux to clean the area and producing a good joint. After the joint is made, the board is still held with the pliers until the joint has cooled down.

**Test Equipment.**

When developing new circuitry, it may be convenient to try different values of resistor in some position in the circuit (the resistor value may be dependent on the gain of a transistor or the actual resistance of an ORP12, or some such other situation). For this, it is very convenient to have a resistor-substitution box which allows you to select any standard resistor at the turn of a switch.

These are not readily available on the market. In years gone by, it was possible to buy custom wafer switches, where the number of wafers could be built up to whatever switch size was required, but these do not seem to be available any more. A slightly less convenient method of construction is to use four of these, selected by a second wafer switch:

![Resistor Substitution Box Diagram](image)

In the above diagram, all of the resistors in one range (100 ohms to 820 ohms, 1K to 8K2, 10K to 82K or 100K to 820K) are wired to a single 12-way switch. The output wires then have any of these standard resistors across them, depending on the setting of the switch. A second switch can then be used to select several of these groups, while still using the same output wires. When boxed, it might look like this:
It can also be useful to have a versatile signal generator. You can easily construct your own with variable frequency, variable mark/space ratio and optional variable gating. If you do, you might as well make it with a low output impedance so that it can drive devices under test directly rather than having to provide additional buffering. It might look like this:

The really essential item of equipment is a multimeter. These come in many shapes, sizes and varieties and the cost varies enormously. The reliability also varies a great deal. The most reliable and the cheapest is the analogue type which does not use a battery (other than for the occasional measurement of resistance). Although these types are looked down upon nowadays, they are 100% reliable:

The meter shown above is rated at 2,000 ohms per volt, so connecting it to a circuit to make a measurement on the 10V range is the same as connecting a 20K resistor to the circuit. The big brother of this style of equipment is about five times larger and has 30,000 ohms per volt performance, so connecting it on a 10V range is the same as connecting a 300K resistor to the circuit being measured. This one is battery driven, so if you get one of these, may I suggest that you check its accuracy on a regular basis:
The really excellent non-battery (ex-professional) Avo meter multimeters are still available through eBay at affordable prices. These have 30,000 ohms per volt performance and are robust and accurate, having been built to very high standards.

A multimeter uses a 1.5V battery to measure resistance. Ohm’s Law is used as the working principle and the operation is:

The meter shown in the diagram has a small resistance of its own. This has a small variable resistor added to it. This variable resistor will have a small knob mounted on the face of the multimeter, or it will be a thumbwheel knob projecting slightly from the right hand side of the multimeter case. The 1.5V battery will be positioned inside the multimeter case as is the 1K resistor. To use the resistance ranges, the multimeter probes are touched firmly together to form a short-circuit and the variable resistor adjusted so that the meter points to zero.

For the purpose of this discussion, let us assume that the internal resistance of the meter, when correctly adjusted, is exactly 1K. If the resistor under test is exactly 1K in value, then the current through the meter will be halved and the meter will show a needle deflection half way across the scale. If the resistor under test is 2K, then the current will be one third and the scale marking will be at the 1/3 position from the left. If the resistor is 4K, then there will be one fifth (1K + 4K = 5K) of the full-scale current and the 4K mark will be 20% from the left hand side of the scale.

Two things to notice: firstly, the scale has to read from right to left which can take some getting used to, and secondly, the scale is not linear, with the markings getting closer and closer together and consequently, more difficult to mark and read, the higher the value of the resistor being measured. The bunching up of the scale markings is why the more expensive multimeters tend to have more than one range.

A mains-operated oscilloscope is an excellent piece of equipment to own but they are expensive when new. It is possible to pick one up at a reasonable price second-hand via eBay. An oscilloscope is by no means an essential item of equipment. One of its most useful features is the ability to measure the frequency, and display the shape of a waveform. Most waveforms are of known shape so the frequency is the major unknown. The following meter is not expensive and it displays the frequency of a signal on a digital readout:
So, when you are deciding what multimeter to buy, consider the following points:

1. How reliable is it? If you are opting for a battery driven unit, what happens to the accuracy if the battery starts to run down. Does it display a warning that the battery needs to be replaced? Mains-operated digital multimeters are brilliant but are a problem if you want to make measurements away from the mains.

2. What DC voltage ranges does it have? If you are intending to work mainly with 12V circuits, it is inconvenient for the ranges to be 9V and 30V as successive ranges. Digital meters do not have this problem but the question then is, how accurate are they going to be in day to day use?

3. Transistor testing options you can ignore - you are better off making your own dedicated unit to check transistors if you think you will ever need to do this - you probably won't.

4. Measuring current can be very useful so see what ranges are offered.

5. Measuring capacitance is very useful, especially since many capacitors are not well marked to indicate their value.

6. Measuring the frequency of a waveform could be a significant bonus but the question is; are you every likely to need it?

7. Measuring resistance is very useful. Every meter has it. There is no need to be over fancy on measurement ranges as you usually only need to know the approximate answer - is it a 1K resistor or a 10K resistor?

Look around and see what is available, how much it costs and what appeals to you. It might not be a bad idea to buy a really cheap multimeter and use it for a while to see if it has any shortcomings which are a nuisance, and if so, what improvements you personally want from a more expensive meter.

The ‘Bench’ Power Supply.
It might be worth getting a fancy bench power supply which allows you to set any voltage you want and which displays the current being drawn by your development circuit:
However, there is no need to spend money on a fancy unit when you can build an excellent unit of your own with voltage stabilisation, adjustable output, metered current, etc. etc. Personally, if developing a circuit to be used with a battery, I believe you are better off powering the development from a battery, that way the characteristics of the battery are included in any tests which you carry out.

If you wish, you can construct a very convenient development test bed power supply system. This has the advantage that you can make it in the most convenient style for your own use. You can also make the protection ultra-sensitive and build in additional circuitry such as transistor tester and resistor substitution box to produce an integrated test bed. You could perhaps use a circuit like this:

Here, the power is supplied by a pack of re-chargeable Ni-Cad batteries or possibly, a mains unit with voltage stabilisation. As in all actual circuits, the next thing in the circuit is always an on/off switch so that the power source can be disconnected immediately should any problem arise. Next, as always, comes a fuse or circuit breaker, so that should the problem be serious, it can disconnect the circuit faster than you can react. If you wish, you can build your own super-accurate adjustable circuit breaker to use in this position.

The two transistors and three resistors form an adjustable, stabilised output. The FET transistor has a high output power handling capacity and a very low input power requirement and so is good for controlling the output voltage.
Resistor ‘VR1’ is padded with the 4K7 resistor solely to reduce the voltage across the variable resistor. VR1 is adjusted to control the output voltage. If the current draw is increased and the output voltage is pulled down slightly, then the voltage on the base of the BC109 transistor is reduced. This starts to turn the transistor off, raising the voltage at point ‘A’, which in turn, raises the output voltage, opposing the variation caused by the load.

The output is monitored, firstly by a large milliammeter to show the current draw and secondly, on the output side of the milliammeter, a voltmeter. This allows very close monitoring of the power supplied to the prototype, especially if the milliammeter is placed alongside the prototype. You can build this circuit into a wide flat box which provides a working surface beside the milliammeter.

At point ‘B’ in the above diagram, a method for altering the current range of the milliammeter by placing a ‘shunt’ resistor across it. When the switch is closed, some current flows through the resistor and some through the milliammeter. This resistor has a very low value, so you are better off making it yourself. Let’s say we wish to double the range of the meter. Solder the switch across the meter and for the resistor use a length of enamelled copper wire wound around a small former. Put a load on the output so that the meter shows a full-scale deflection. Close the switch. If the current displayed is exactly half of what it was, if not, switch off, remove some wire to lower the reading or add some wire to raise the reading and repeat the test until exactly half the current is displayed. The lower the value of the shunt resistor, the more current flows through it and the less through the meter, which then gives a lower reading.

Please note: it is very important to have a fuse or circuit breaker in the power being delivered to your test circuit. Any error in building the prototype can cause a major current to be drawn from the supply and this can be dangerous. Remember, you can’t see the current. Even if you have a meter on the current being delivered, you may not notice the high reading. The first sign of trouble may be smoke! You can easily fry the circuit you are building if you do not have a safety cut-off, so use a fuse or other device which limits the current to twice what you are expecting the circuit to draw.

So, after all that, what equipment do you really need? You need a small soldering iron and multicore solder, a pair of long-nosed pliers and a multimeter. One other thing is some tool to cut wires and remove the insulation prior to soldering. Personal preferences vary. Some people prefer one of the many custom tools, some people use a knife, I personally use a pair of straight nail scissors. You pick whatever you are comfortable with.

Not exactly a vast array of essential equipment. The other items mentioned are not by any means essential so I suggest that you start by keeping things simple and use a minimum of gear.

If you are not familiar with electronics, I suggest that you get a copy of the Maplin catalogue, either from one of their shops or via the http://www.maplin.co.uk web site. Go through it carefully as it will show you what components are available, how much they cost and often, how they are used. The specifications of almost any semiconductor can be found free at http://www.alldatasheet.co.kr in the form of an Adobe Acrobat document.

Finally, because it is not important, all of the circuitry shown so far has indicated current flowing from the + of a battery to the - terminal. The discovery of voltage was made by Volta but he had no way of knowing which way the current was flowing, so he guessed. He had a 50 - 50 chance of getting it right but he was not lucky and got it wrong. Electrical current is actually a flow of electrons, and these flow from the battery minus to the battery plus. So, who cares? Almost nobody, as it has no practical effect on any of the circuitry. Some useful websites:

http://www.esr.co.uk for components
http://www.maplin.co.uk for components
http://www.alldatasheet.co.kr for semiconductor specifications
http://www.cricklewoodelectronics.com for components
http://www.greenweld.co.uk for components

The Oscilloscope.
If you do decide that you are going to research new equipment, design and possibly invent new devices, then an oscilloscope is useful. Let me stress again that this is not an essential item of equipment and most certainly is not needed until you are quite familiar with constructing prototypes. It is quite easy to misread the settings of an oscilloscope and the methods of operation take some getting used to. The low-cost book “How to Use Oscilloscopes and Other Test Equipment” by R.A. Penfold, ISBN 0 85934 212 3 might well be helpful when starting to use a ‘scope.

It is possible to get an oscilloscope at reasonable cost by buying second-hand through eBay. The best scopes are ‘dual trace’ which means that they can display the input waveform and the output waveform on screen at the same time. This is a very useful feature, but because it is, the scope which have that facility sell at higher prices.
The higher the frequency which the scope can handle, the more useful it is, but again, the higher the selling price. Not all scopes are supplied with (the essential) ‘test probes’, so it might be necessary to buy them separately if the seller wants to keep his. Getting the manual for the scope is also a decided plus. A low-cost scope might look like this:

Measuring Magnetic Field Strength.
People who experiment with permanent magnets, can make use of an instrument which displays the strength of a magnetic field. Professionally made devices to do this tend to be well outside the purchasing power of the average experimenter who will already have spent money on materials for his prototypes. Here is a design for a simple and cheap circuit, powered by four AA dry cell batteries, and utilising a Hall-effect semiconductor as the sensor:

This design uses an OP77GP operational amplifier chip to boost the output signal from the A1302 chip which is a Hall-effect device. The gain of the DC-connected operational amplifier is set by the ratio of the 1K and 1M fixed resistors shown shaded in the circuit diagram, giving a gain of 1,000.

The circuit operation is simple. The six-volt battery charges the 10 microfarad capacitor which helps iron out any supply line fluctuations caused by varying current draw by the circuit. The 10K variable resistor is used to set the
output meter display to zero when the Hall-effect device is not near any magnet. The 1K variable resistor is there to make fine tuning adjustments easier.

When the A1302 chip encounters a magnetic field, the voltage on its output pin 3 changes. This change is magnified a thousand times by the OP77GP amplifier. Its output on pin 6 is connected to one side of the display meter and the other side of the meter is connected to point “A”. The voltage on point “A” is about half the battery voltage. It would be exactly half the voltage if the two 4.7K resistors were exactly the same value. This is rather unlikely as there is a manufacturing tolerance, typically around 10% of the nominal value of the resistor. The exact value of the voltage on point “A” is matched by the OP77GP tuning and so the meter reads zero until a magnetic field is encountered. When that happens, the meter deflection is directly proportional to the strength of the magnetic field.

The Weird Stuff.
You don’t need to know the following information, so please feel free to skip it and move on to something else.

The presentation shown above is based on the conventional view of electronics and electrical power as taught in schools and colleges. This information and concepts works well for designing and building circuits, but that does not mean that it is wholly correct. Unfortunately, the world is not as simple as is generally made out.

For example, it is said that current is a flow of electrons passing through the wires of a circuit at the speed of light. While it is true that some electrons do actually flow through the metal of the wires, the small percentage of electrons which actually do that, do it quite slowly as they have to negotiate their way through the lattice of the molecules of metal making up the body of the wires.

In spite of this, when the On/Off switch of a circuit is flipped on, the circuit powers up immediately, no matter how long the wires are. The reason for this is that electrical current flows along the wires at very high speed indeed, but it flows rapidly along the outside of the wires, not rapidly through the wires. One thousandth of a second after switching on a circuit, the electrons flowing through the wires have hardly got started, while the current flowing along the outside of the wires has gone all around the circuit and back:

The above sketch does not show the proportions correctly, as the current flow spiralling along the outside of the wire should be hundreds of thousands of times longer than shown, which is not practical in a diagram.

The actual path taken by current flow makes the surface of the wire of particular importance, and the insulation material is also of great importance. In years gone by, wire manufacturers used to anneal (cool down) copper wires in air. This created a layer of cupric oxide on the outer surface of copper wires, and that layer gave the wire different characteristics than copper wire has today. William Barbat in his patent application claims that the cupric oxide layer can be utilised in making devices with greater power output than the power input from the user.

Unfortunately, the world is not quite as simple as that, as power flowing in a circuit has at least two components. The electrical current which we measure with ammeters is as described above and is sometimes referred to as “hot” electricity as when it flows through components, it tends to heat them up. But there is another component referred to as “cold” electricity, so named because it tends to cool components down when it flows through them. For example, if the output wires of Floyd Sweet’s VTA device were short circuited together, frost would form on the device due to the heavy flow of “cold” electricity, and getting a “shock” from it could give you frostbite instead of a burn.

“Cold” electricity is not something new, it has always been there as it is just one aspect of “electricity”. It has not been investigated much by conventional science because none of the instruments used to measure “hot” electricity, react to “cold” electricity at all. (Actually, “hot” electricity, “cold” electricity and magnetism are all features of a single entity which should really be called “electromagnetism”).

Now the spooky bit: “cold” electricity does not flow along or through the wire at all. Instead, it flows in the space around the wire, possibly riding on the magnetic field caused by the “hot” current. Thomas Henry Moray is famous for building a device which captured “cold” electricity and produced a massive power output capable of powering a whole host of ordinary electrical pieces of equipment. In his many public demonstrations before he was intimidated into silence and his equipment smashed, he invited members of the audience to bring a piece of ordinary glass with them. Then, when his circuit was powering a row of lights, he would cut one of the wires and
insert the piece of glass between the cut ends of the wires. This had no noticeable effect on his circuit, with the power flowing happily through the glass and on through his circuit, powering the lights just as before. That does not happen with “hot” electricity, but as the “cold” electricity is not flowing through or along the surface of the wire, a break in the wire is not a major obstacle to it.

We still do not know very much about “cold” electricity. Edwin Gray snr. demonstrated light bulbs powered by “cold” electricity being submerged in water. Not only did the bulbs continue to operate unaffected by the water, but Edwin often put his hand in the water along with the lit bulb, suffering no ill effects from doing so. Neither of those two effects are possible with conventional electricity, so please don’t try them to check it out.

Another interesting item is the water-powered car system produced by an American man Nathren Armour. His system, (among other things) involves feeding extra electrical power to the spark plugs. One thing which has always puzzled him is that the engine will not run with just one wire going to the spark plug cap. He has to have a second wire running from his extra power supply to the body of the plug where it screws into the engine block. Take that wire away and the engine stops. Put it back again and the engine runs. But according to conventional electrics, that wire cannot possibly be needed, because the engine block is grounded and the power supply output is grounded, so in theory, there is no voltage difference between the ends of the wire, therefore no current can flow along the wire, hence the wire is not needed and has no function. Well, that is true for “hot” electricity, but it seems possible that the Nathren Armour system is using “cold” electricity as well as “hot” electricity and the “cold” electricity needs the extra wire as a flow guide to the spark plug.

Enough about that for now. Let’s go one step further into the “weirdness” of the actual world. If, three hundred years ago, you had described X-rays, gamma rays, nuclear energy and TV signals to the average well-educated person, you would have run a considerable risk of being locked up as being mad. If you do it today, your listener would probably just be bored as he already knows all this and accepts it as a matter of fact (which it is). Please bear that in mind when you read the following information. If it seems strange and far-fetched, that is only because conventional science today is lagging badly behind and still teaching things which have been conclusively proven to be wrong decades ago.

If you lived in a desert and every day a company drove in with a lorry-load of sand and sold it to you for a large amount of money, what would you think about that? Not a very good deal for you, is it? What’s that you say, you would never do that? But you already do, because you don’t realise that the sand is all around you ready for the taking at next to no cost at all. Several people have tried to publicise the fact, but the sand company has immediately silenced them by one means or another. The company does not want to lose the business of selling you the sand and definitely doesn’t want you to start picking it up for yourself for free.

Well... to be perfectly fair, it is not actually sand, it is energy, and it is all around us, free for the taking. Sound a bit like X-rays did three hundred years ago? Doesn’t mean that it is not true. It is perfectly true. The design of all computers made today is based on the equations of Quantum Mechanics, and while those equations are not yet perfect, they are easily good enough for practical purposes. The snag is that the world seen at the level of the quantum is not much like the world we think that we see around us and which we think that we understand fully. Examining the world at the quantum level shows that we live in a seething mass of incredible energy. Einstein is famous for stating that Mass equals a very large amount of Energy, a fact that is shown clearly when an atomic bomb is detonated. Put in different words, a small amount of matter is the equivalent of a very large amount of energy. Actually, Energy and Matter are two different aspects of a single thing (which could reasonably be called “Mass-Energy”).

At the quantum level, it can be seen that particles of matter pop into existence and drop out again into energy on a continuous basis, everywhere in the whole of the universe. The whole universe is seething with energy. That energy doesn’t bother us any more than water bothers a fish, as we evolved in this sea of energy and we just don’t notice it. It doesn’t harm us, but if we wanted, and knew how, we could use as much of that energy as we wanted for ever and ever. The amount of that energy is unbelievable. It has been calculated that one cubic centimetre anywhere in the universe contains enough energy to create all of the matter we can see in the whole of the universe. Think how many cubic centimetres there are in the Earth ... the Solar System ... our Galaxy ... If every person on Earth were to run their vehicles, power their homes, fly their planes, etc. etc. for the next million years, it would not make the slightest dent in the energy contained in one cubic millimetre of the universe. This is not a theory, it is a fact. (Would you like to buy a big pile of sand? - I’ve got a load just over here...). This big energy field has gone under different names over the years. A popular name at the present time is the “Zero-Point Energy Field” and it is responsible for everything that happens in the universe. It powers life itself. It balances out in equilibrium everywhere, which is one reason which makes it hard to realise that it is all around us.

Tom Bearden is an American man with very considerable abilities and considerable in-depth knowledge of how the world actually operates. His statements are generally based on laboratory-proven criteria backed up by his high level of mathematical skills which give him an additional grasp of things. He explains how electricity actually works in circuits, and it is nothing like the system taught in schools and colleges. We think that when we attach a
battery to an electrical circuit, the battery forces a current through the wires of the circuit. Sorry Chief - it is actually nothing like that at all. The power in the circuit comes directly from the Zero-Point Energy Field and has very little to do with the battery at all. We tend to think of “using up” power, but that is just not possible. Energy cannot be destroyed or “used up” the most you can do to it is to change it from one form to another. It will perform “work” (power equipment, generate heat, generate cold...) when it changes from one form to another, but if you reverse the process and convert it back to it’s original form, it will perform another lot of “work” during the conversion and end up back in exactly the same state as it started out from, in spite of having performed two lots of “work” during the operation.

A battery does not provide energy to power a circuit. Instead, what happens is that the chemical action inside the battery causes negative charges to gather at the “minus” terminal of the battery and positive charges to gather together at the “plus” terminal of the battery. These two close-together “poles” of the battery are called a “dipole” (two opposite poles near each other) and they have an effect on the Zero-Point Energy Field which is everywhere. The “Plus” pole of the battery causes a massive cluster of Zero-Point Energy Field negative charges to cluster around it. In the same way, the “Minus” pole of the battery causes a massive gathering of ZPE (“Zero-Point Energy”) positive charges to gather around it. Not only do these charges gather around the poles of the battery, but an imbalance in the energy field is created and the ZPE charges continue to arrive at the poles and they radiate out in every direction in a continuous stream of incredible energy.

So, there is your shiny new battery sitting there, not connected to anything and yet it causes massive energy streams to radiate out from its terminals in every direction. We don’t notice it, because the energy flows freely through us and we can’t feel it and none of our conventional instruments, such as voltmeters, ammeters, oscilloscopes, etc. react to it at all.

The situation changes immediately if we connect a circuit to the battery. The circuit provides a flow path for the ZPE energy to flow along, and a significant amount of energy flows near the wires of the circuit, actually powering the circuit for a split second until it reaches the battery “pole” at the far end of the circuit. When it gets there it promptly wipes out the pole, destroying it completely. The ZPE field calms down and the energy flow ceases. But our trusty battery immediately does it all again, using it's chemical energy to create the “dipole” once more, and the imbalance of the ZPE field starts again. It is because the battery has to use it’s chemical energy all the time, creating and re-creating, and re-creating it’s “dipole” that it runs down and eventually ceases to be able to create the dipole any more - result: no more power in the circuit.

Sorry to spoil the illusion, but the battery never did power the circuit itself, it merely acted as channelling device for the Zero-Point Energy Field. In passing, Direct Current (“DC”) is actually not a continuous current at all, but instead it is a stream of DC pulses at an incredibly high frequency - way higher than we can measure at present. The speed of the pulses is so great that it looks continuous to us, a bit like the individual still pictures which are the frames of a movie, appear to be a moving image to us if they are played one after the other at a rate of 25 per second - it looks like continuous movement to us, but in reality, it is a rapid series of still pictures.

The way that a battery “dipole” works on the Zero-Point Energy Field is rather like the way that a magnifying glass acts on sunlight. The rays of the sun get concentrated into a point, focused by the lens. You can start a fire with the lens, and it would be easy to think that the lens started the fire, when in actual fact, it is the rays of the sun that started the fire and the lens just influenced a local area of the large “field” of sunlight, raising the temperature at just one point.

While we tend to think of a “dipole” being generated by a battery, the same effect is also created by a magnet, whether an electromagnet or a permanent magnet - remember that electricity and magnetism are two faces of the same entity. It is possible, but not easy, to capture the energy streaming out from the interference with the ZPE field caused by the poles of a magnet. For example, Hans Coler managed to do this with a completely passive device which, when set up correctly, could produce electrical power, hour after hour from apparently “nothing” (well, actually, the ZPE field). Roy Meyers also did it with his patented array of magnets and zinc plates - completely passive, with no moving parts at all, no battery and no circuitry.
### TABLE OF WIRE SIZES:

The wire sizes specified for use in some designs are American Wire Gauge so a comparison table showing the UK 'Standard Wire Gauge' (with lengths on a 500 gram reel of enamelled copper wire), and the 'American Wire Gauge' is given here:

<table>
<thead>
<tr>
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Patrick Kelly